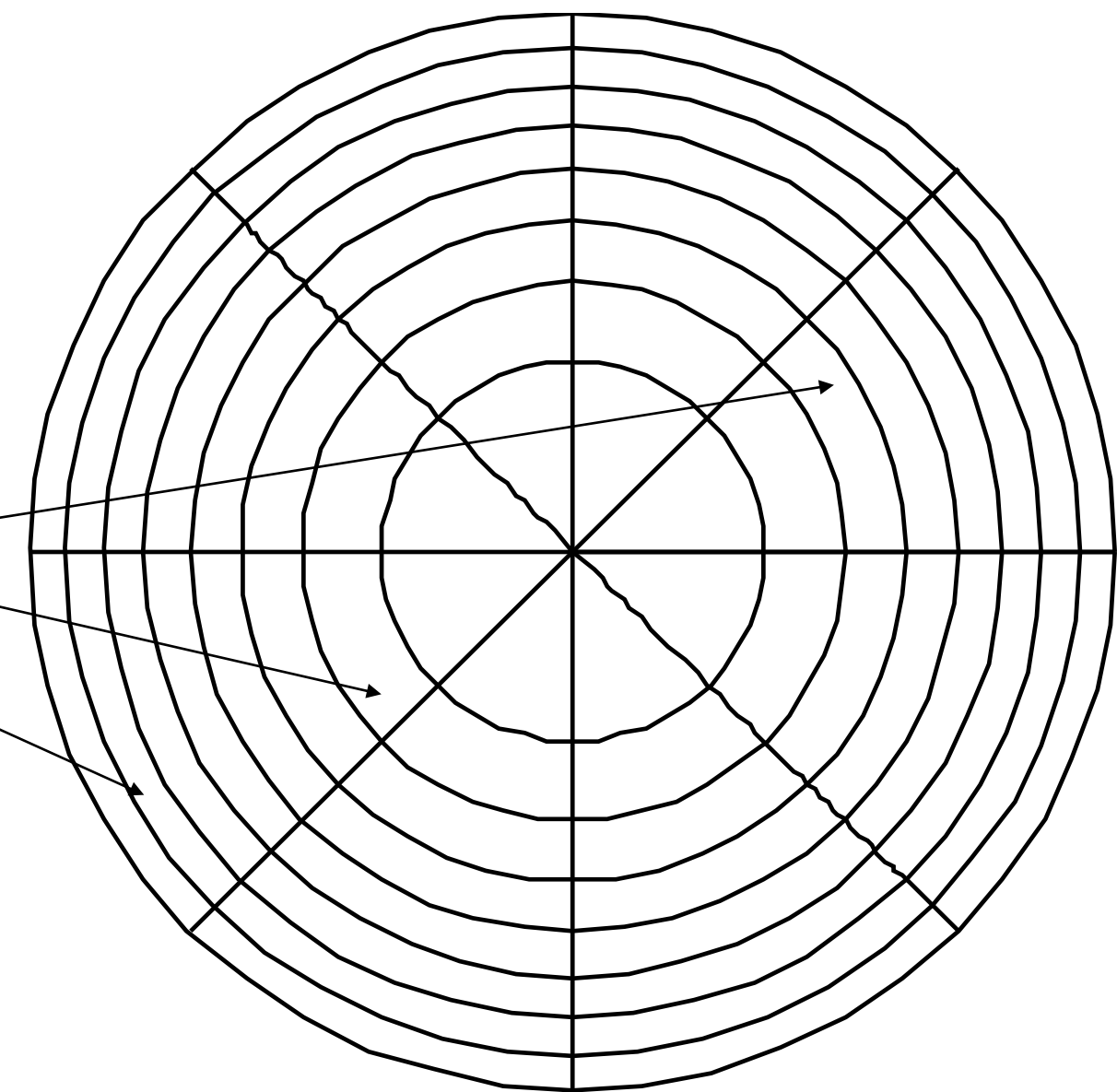
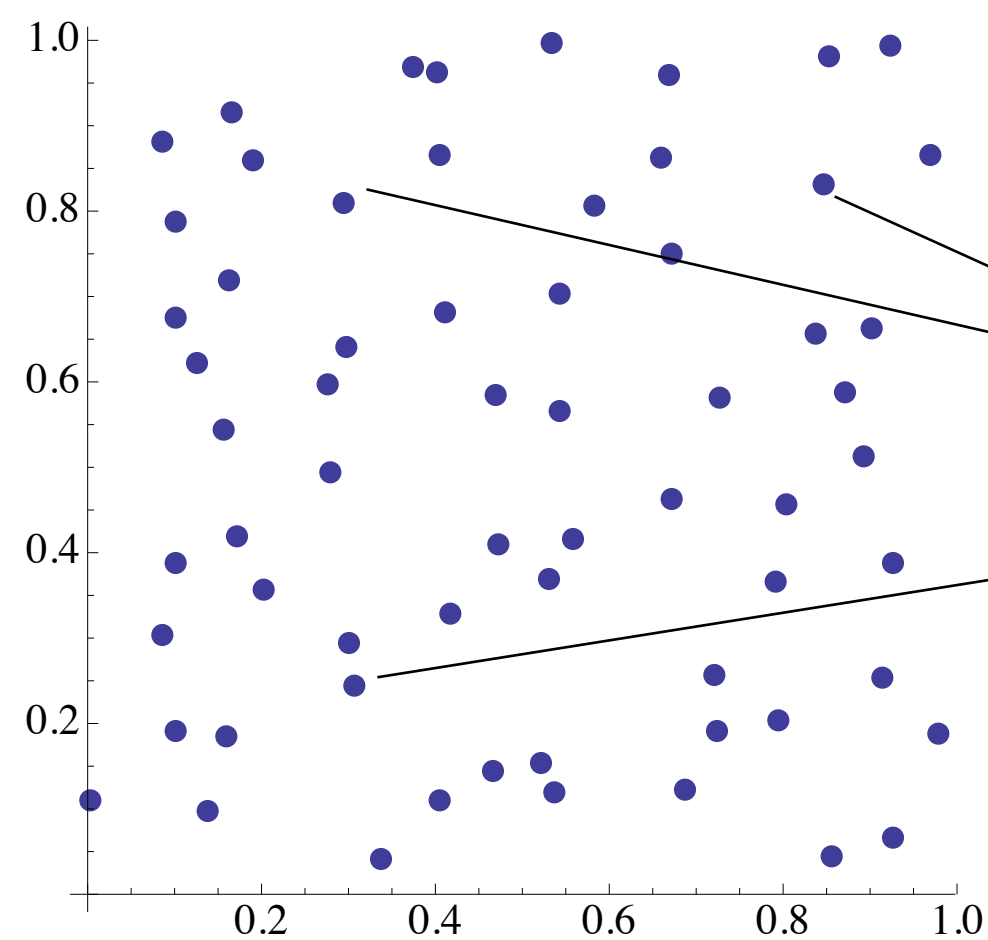


Monte Carlo 2

Today

- **Discrepancy and Quasi-Monte Carlo (QMC)**
- **Low-discrepancy constructions**
 - **Halton, Hammersley, Sobol'**
- **Randomized low-discrepancy: RQMC**
- **Spectral analysis of sampling patterns and MC**

Warping Samples For MC Integration



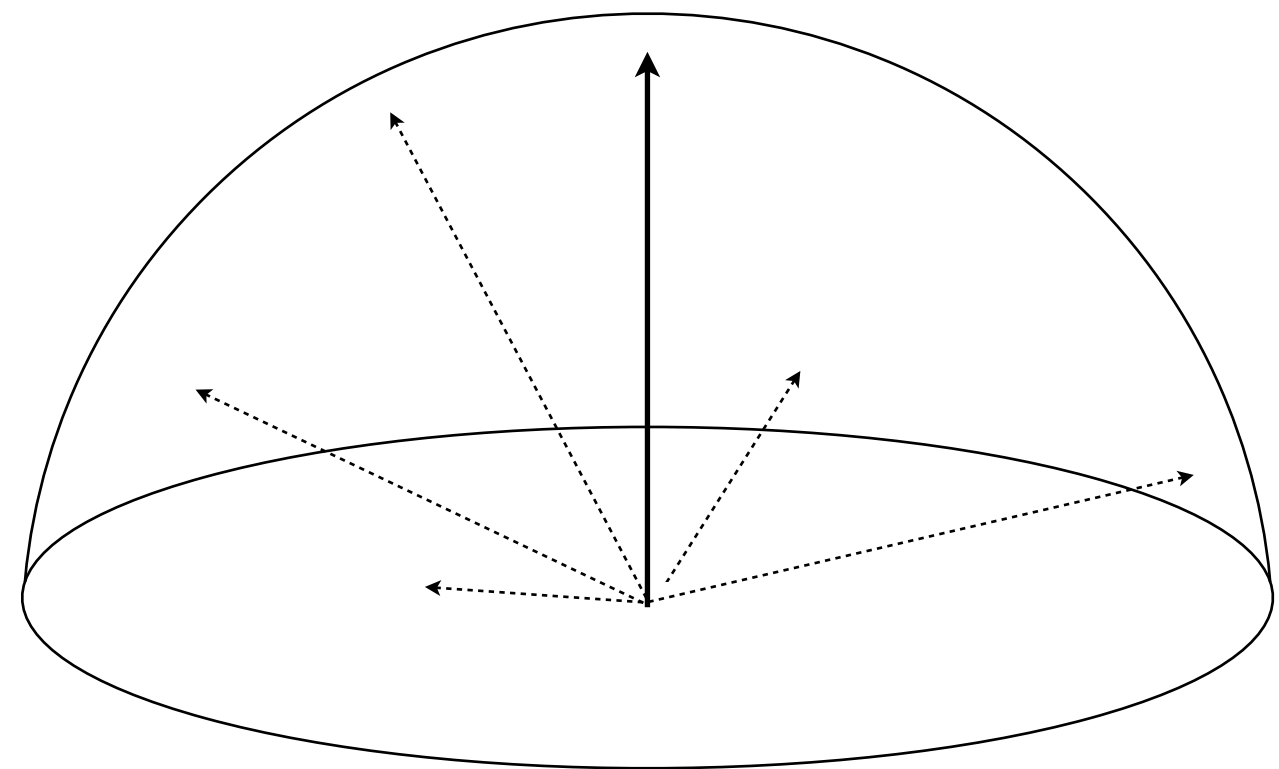
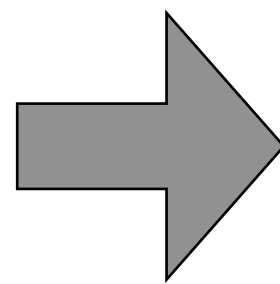
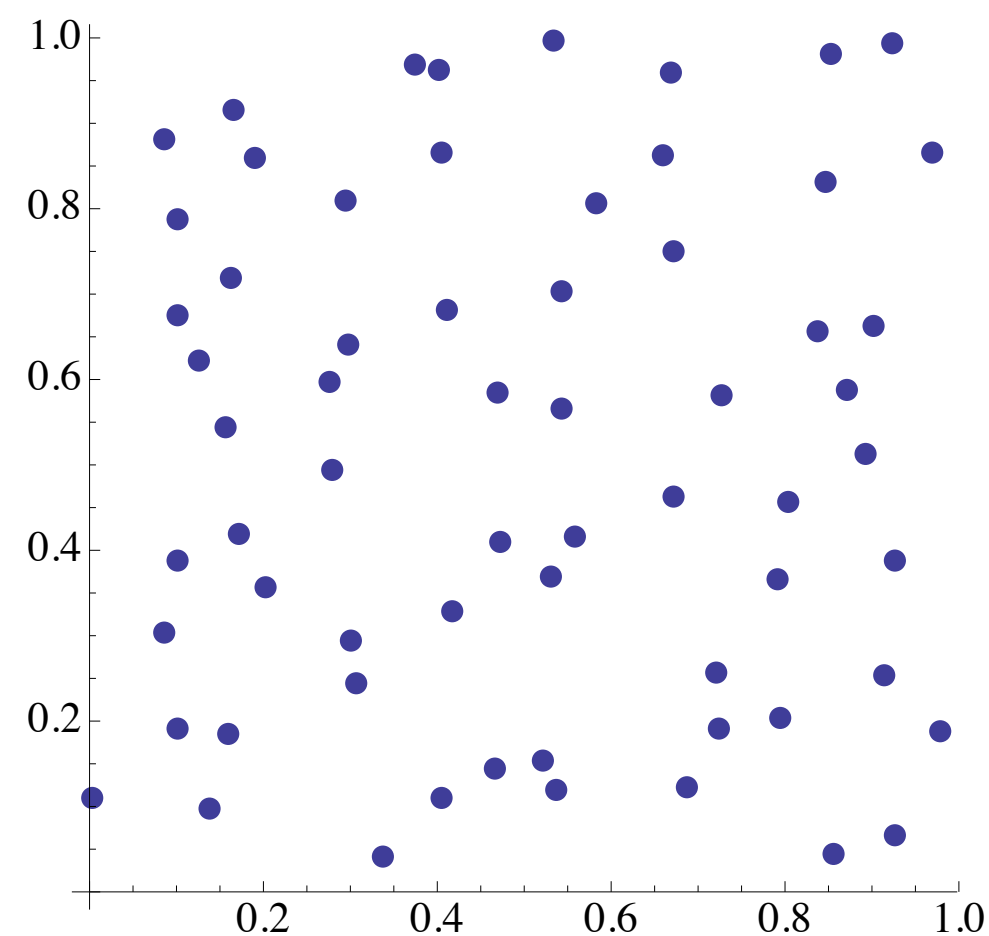
$$\xi_i \in [0, 1)^2$$

$$\theta = 2\pi\xi_1$$

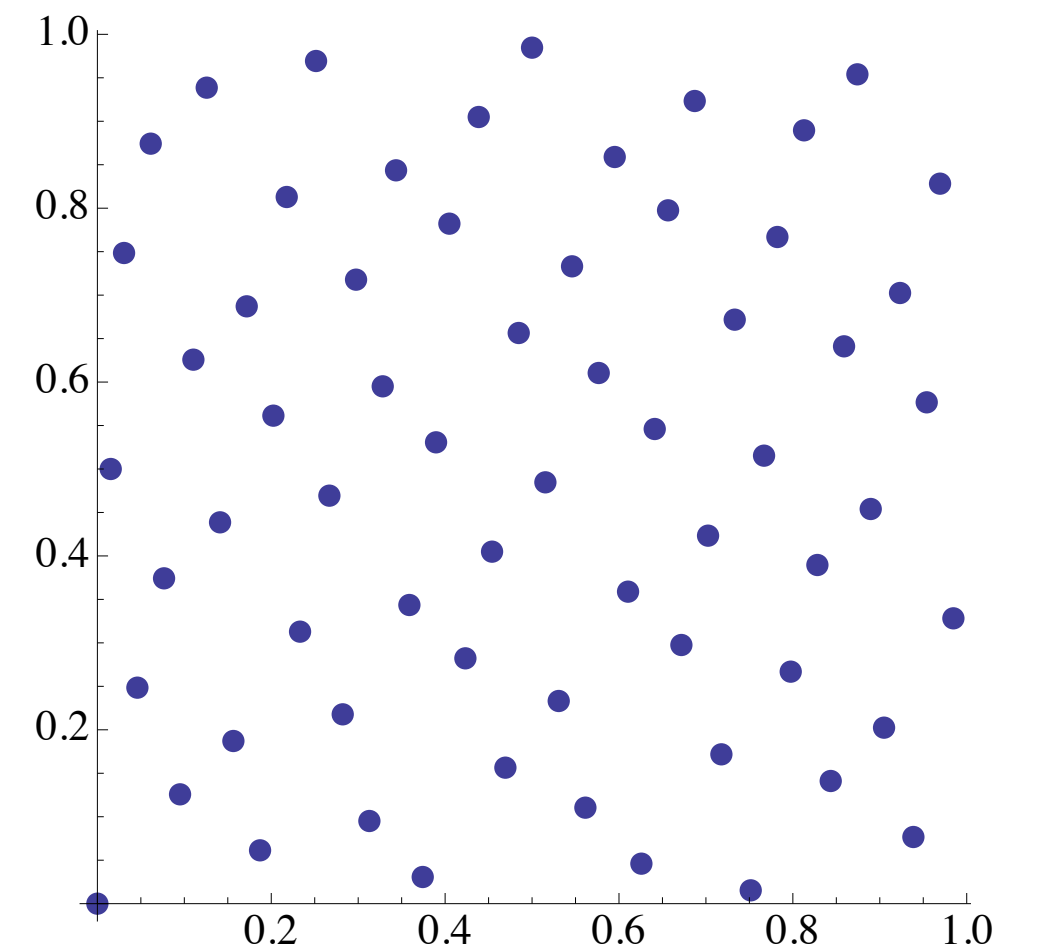
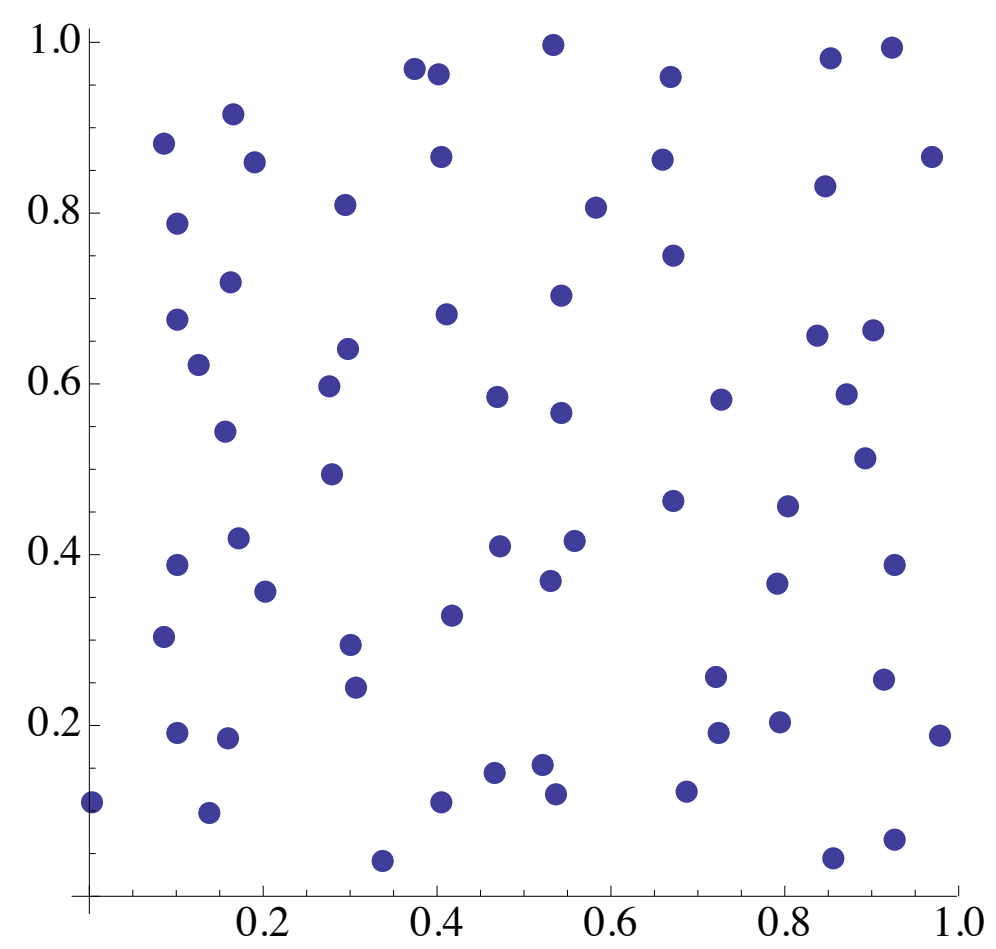
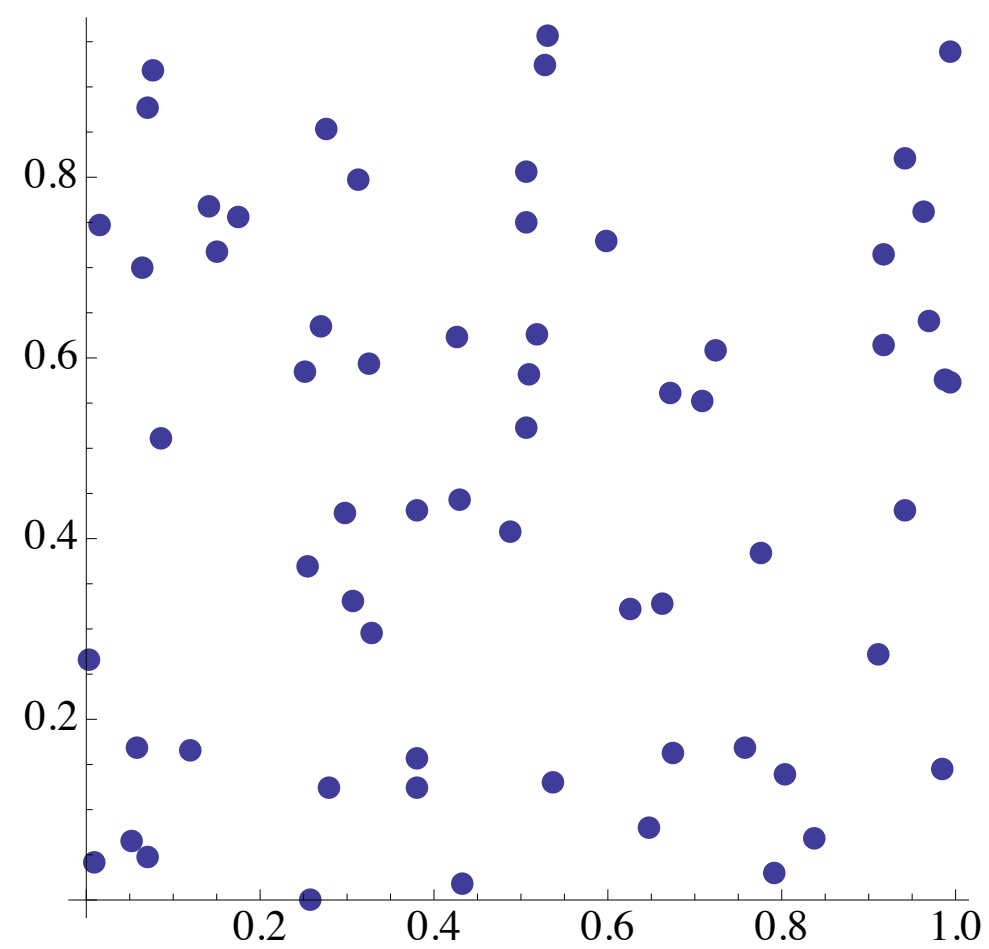
$$r = \sqrt{\xi_2}$$

Warping Samples For MC Integration

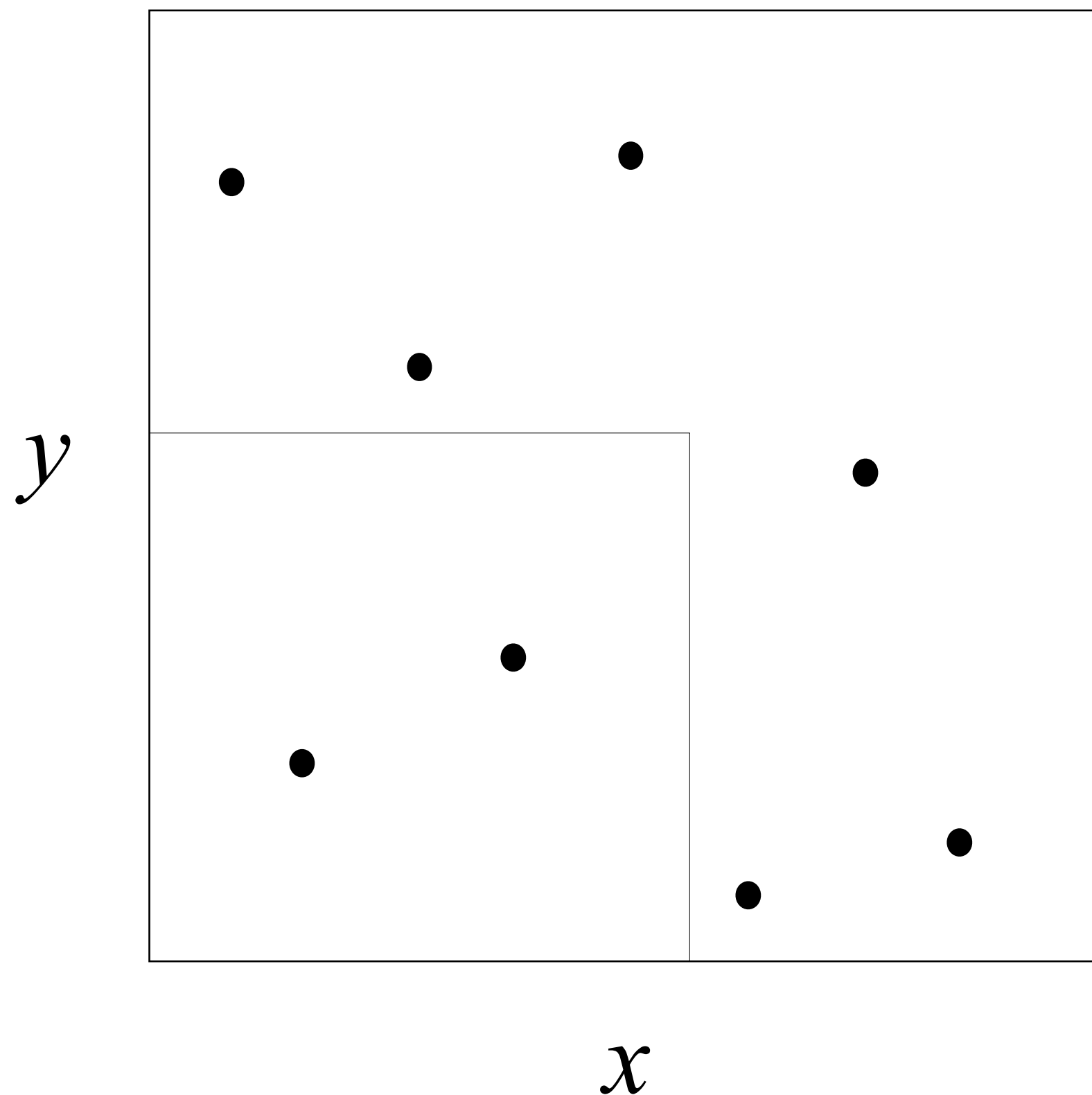
Cosine-weighted hemisphere sampling:



Three 2D Point Sets



Point Set Evaluation: Discrepancy



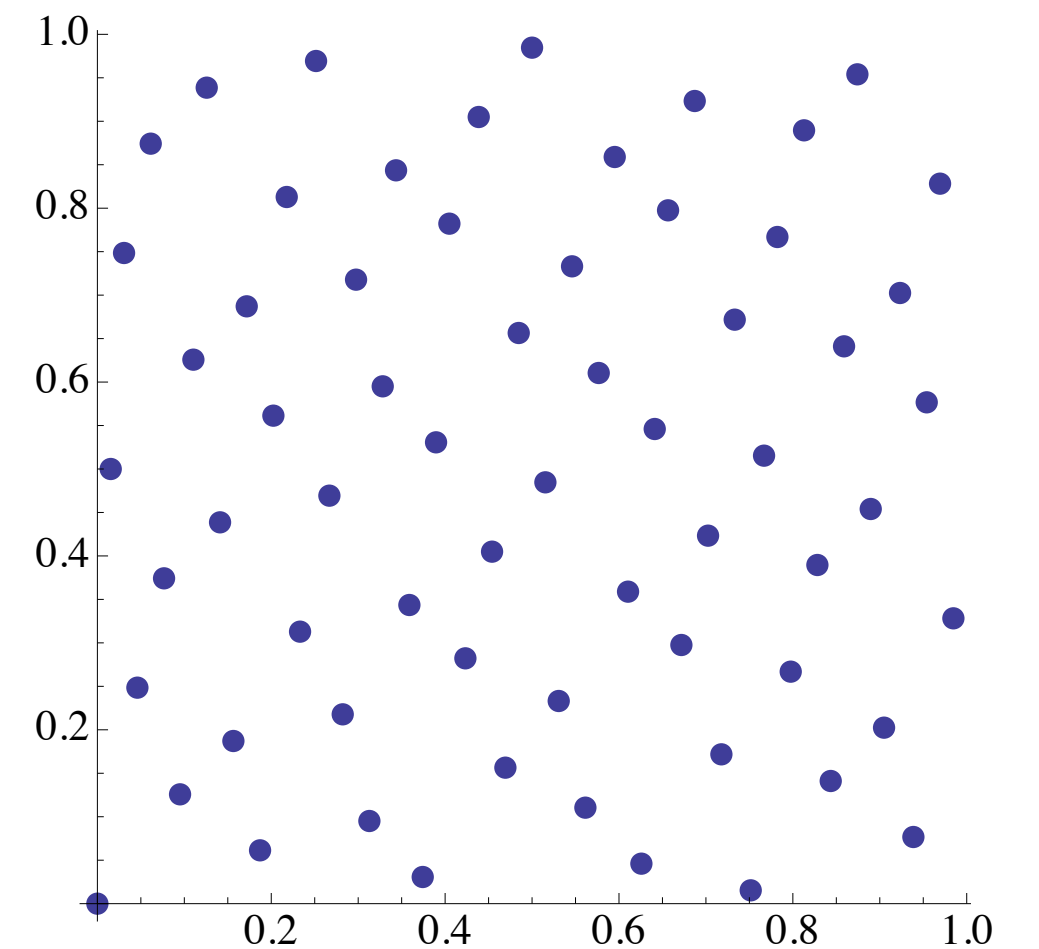
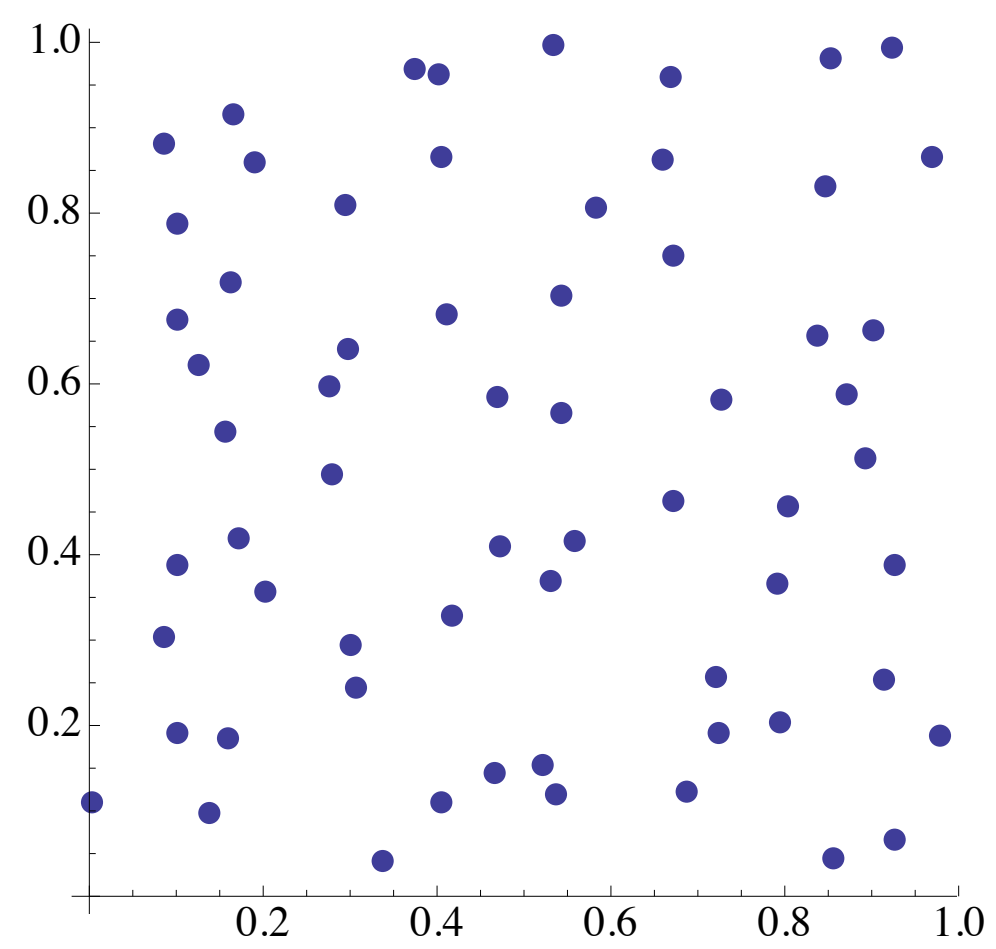
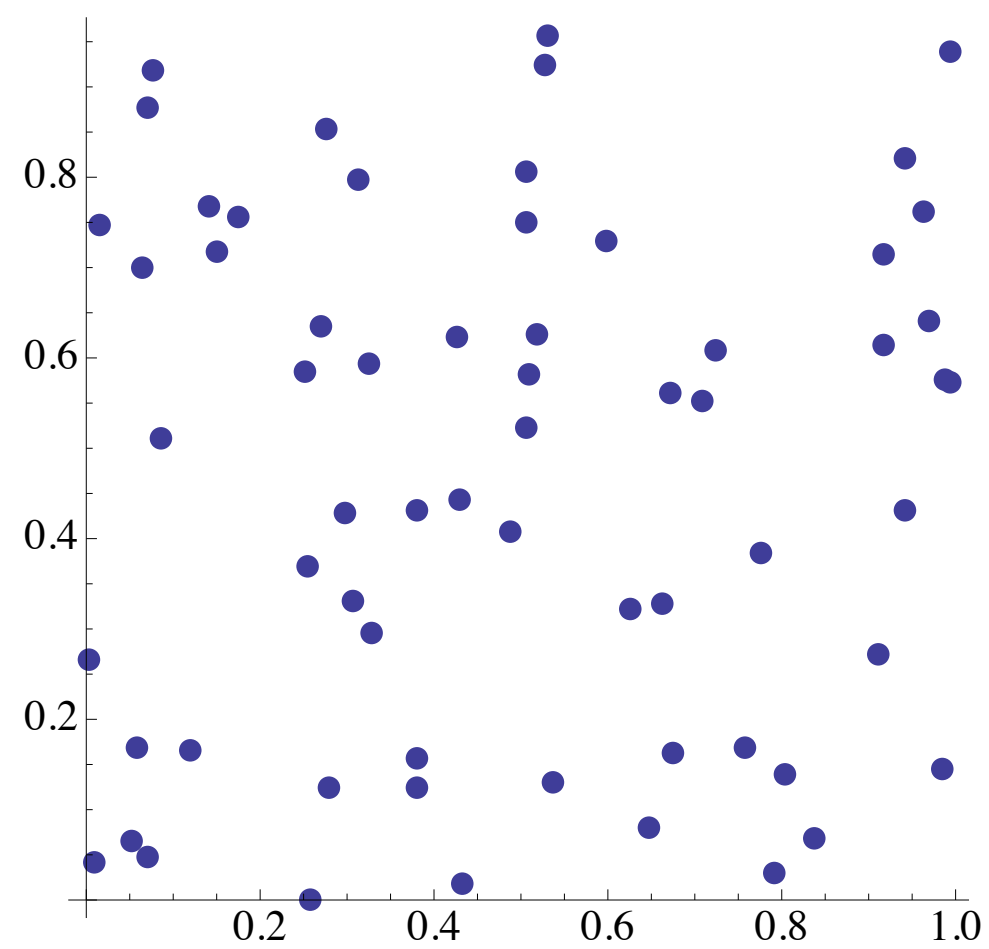
$$\Delta(x, y) = \frac{n(x, y)}{N} - xy$$

$$A = xy$$

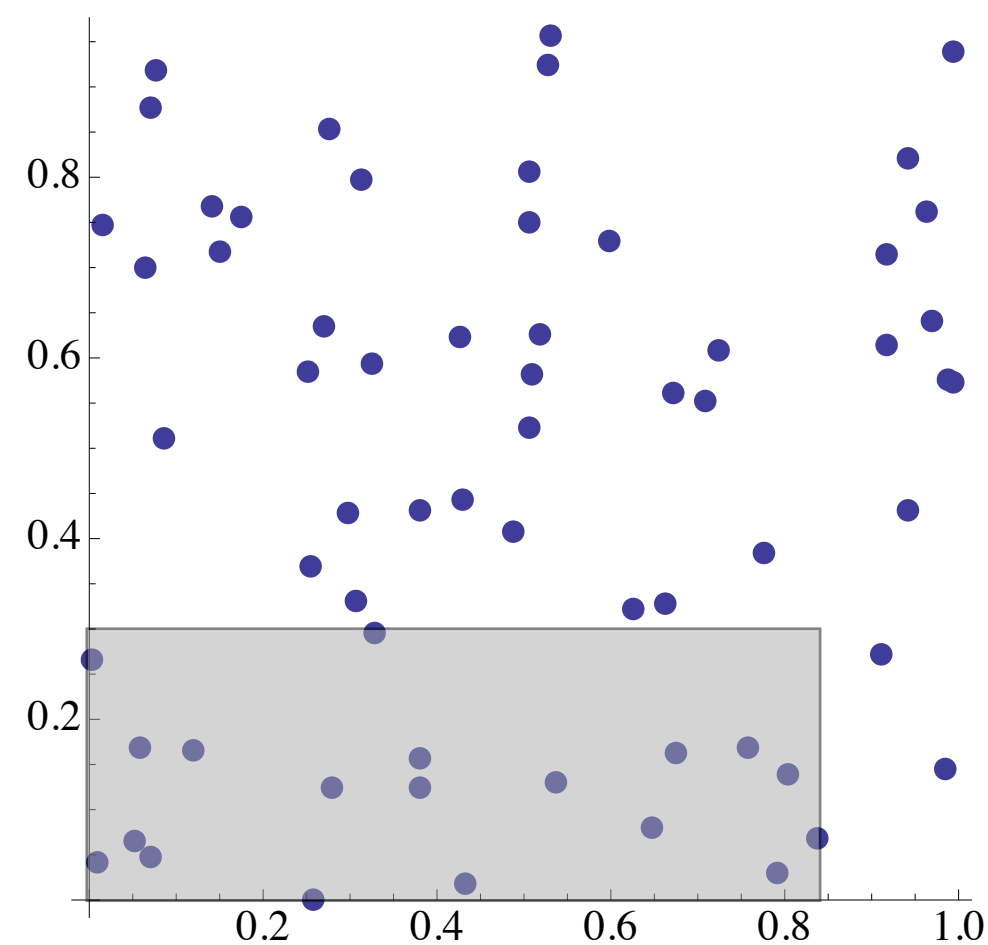
$n(x, y)$ number of samples in A

$$D_N = \max_{x, y} |\Delta(x, y)|$$

Three 2D Point Sets

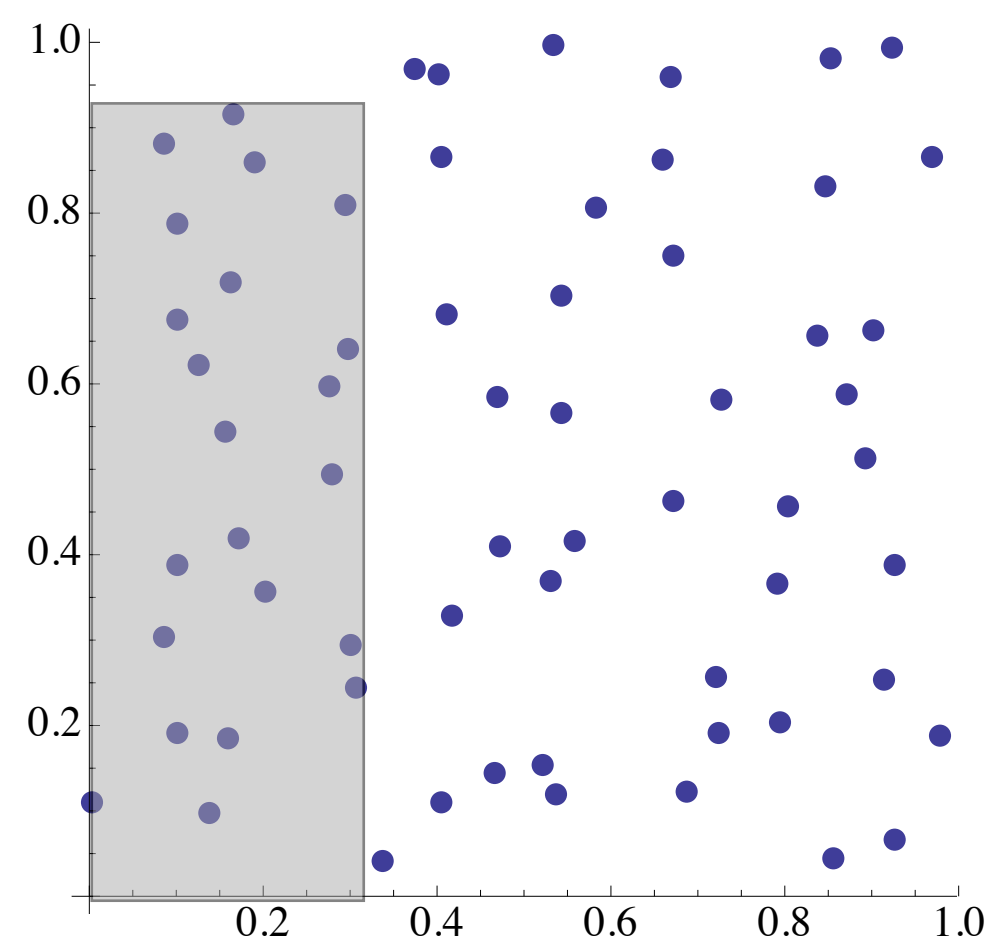


Discrepancy (Empirical)



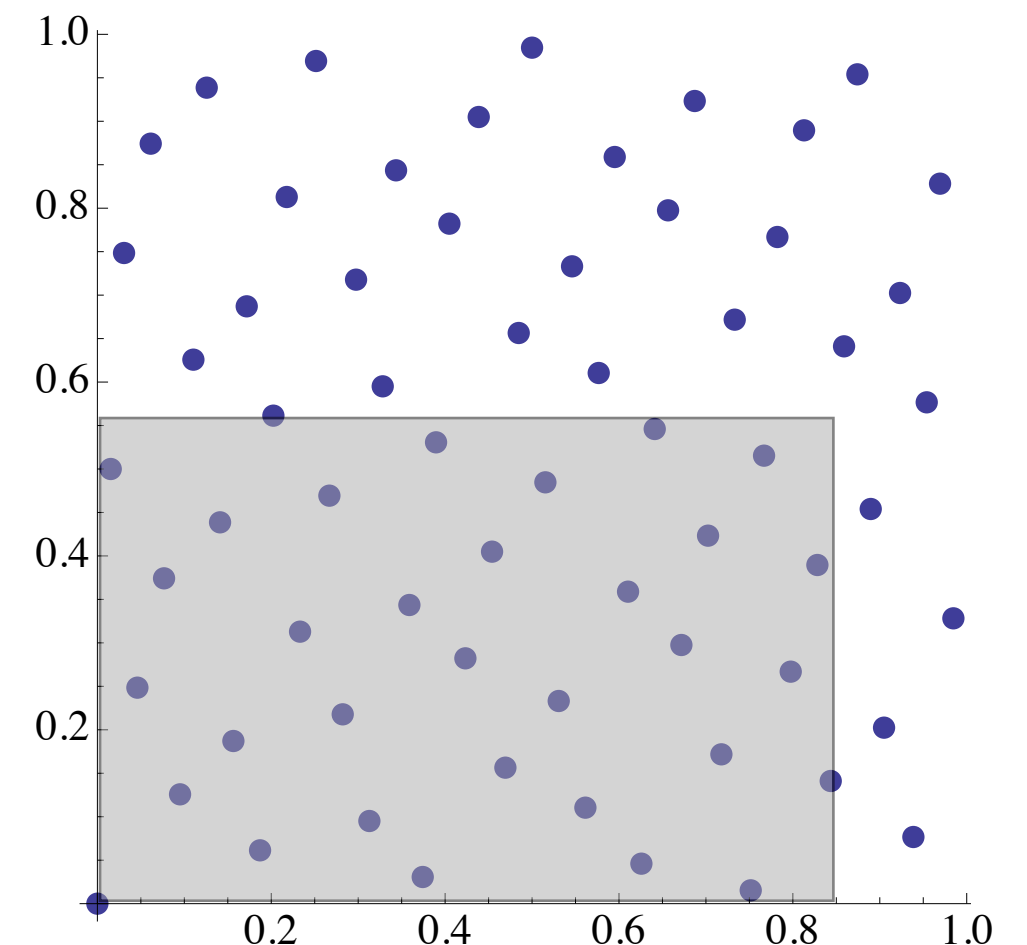
Independent

0.148



Stratified

0.081



**Larscher-
Pillichshammer**

0.041

Low-Discrepancy Definition

An (infinite) *sequence* of n samples in dimension d is low discrepancy if:

$$D_n = O\left(\frac{(\log n)^d}{n}\right)$$

A (finite) *set* of n samples in dimension d is low discrepancy if:

$$D_n = O\left(\frac{(\log n)^{d-1}}{n}\right)$$

Theorem on Total Variation

Koksma-Hlawka inequality:

$$\left| \frac{1}{N} \sum_{i=0}^{N-1} f(X_i) - \int f(x) dx \right| \leq V(f) D_N$$

$$V(f) = \int \left| \frac{\delta f}{\delta x} \right| dx$$

Quasi-Monte Carlo Error Bounds

**Although error is bounded as $|e| \leq V(f)D_N$
not a tight bound!**

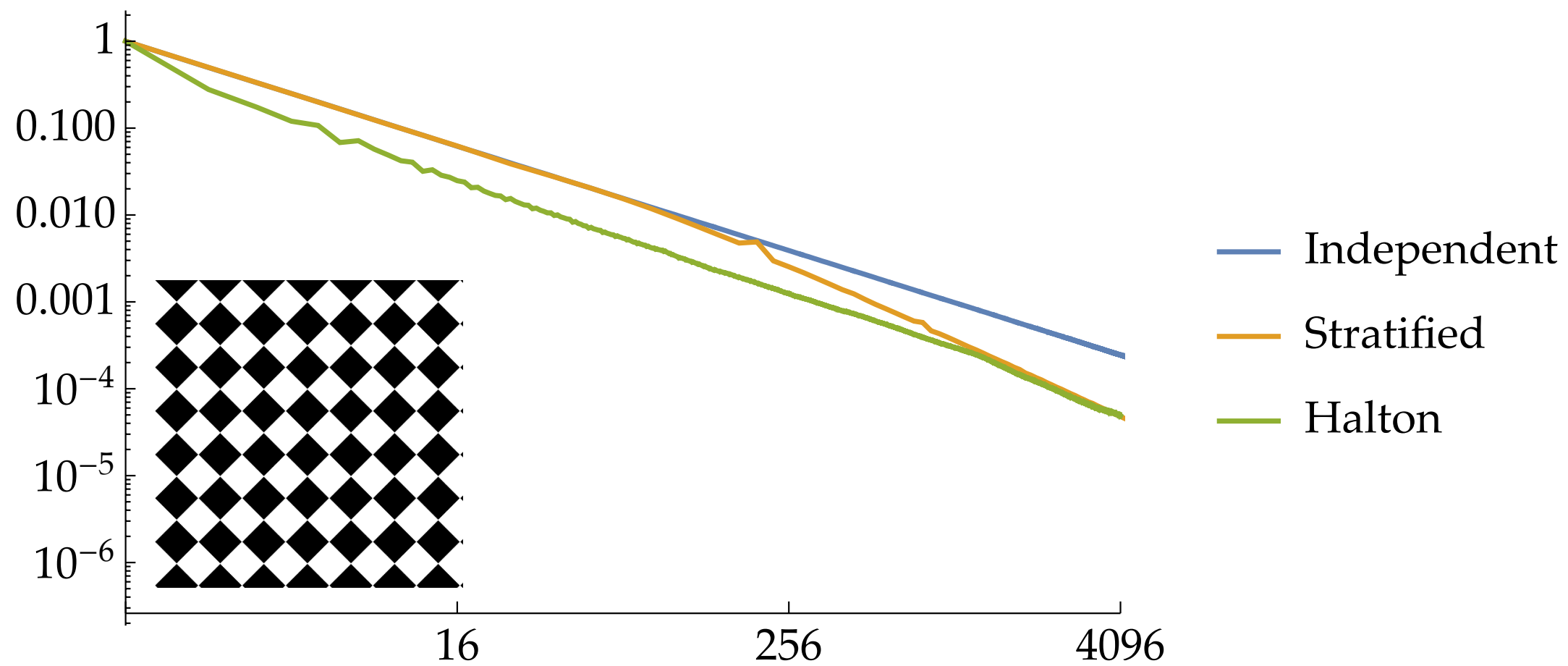
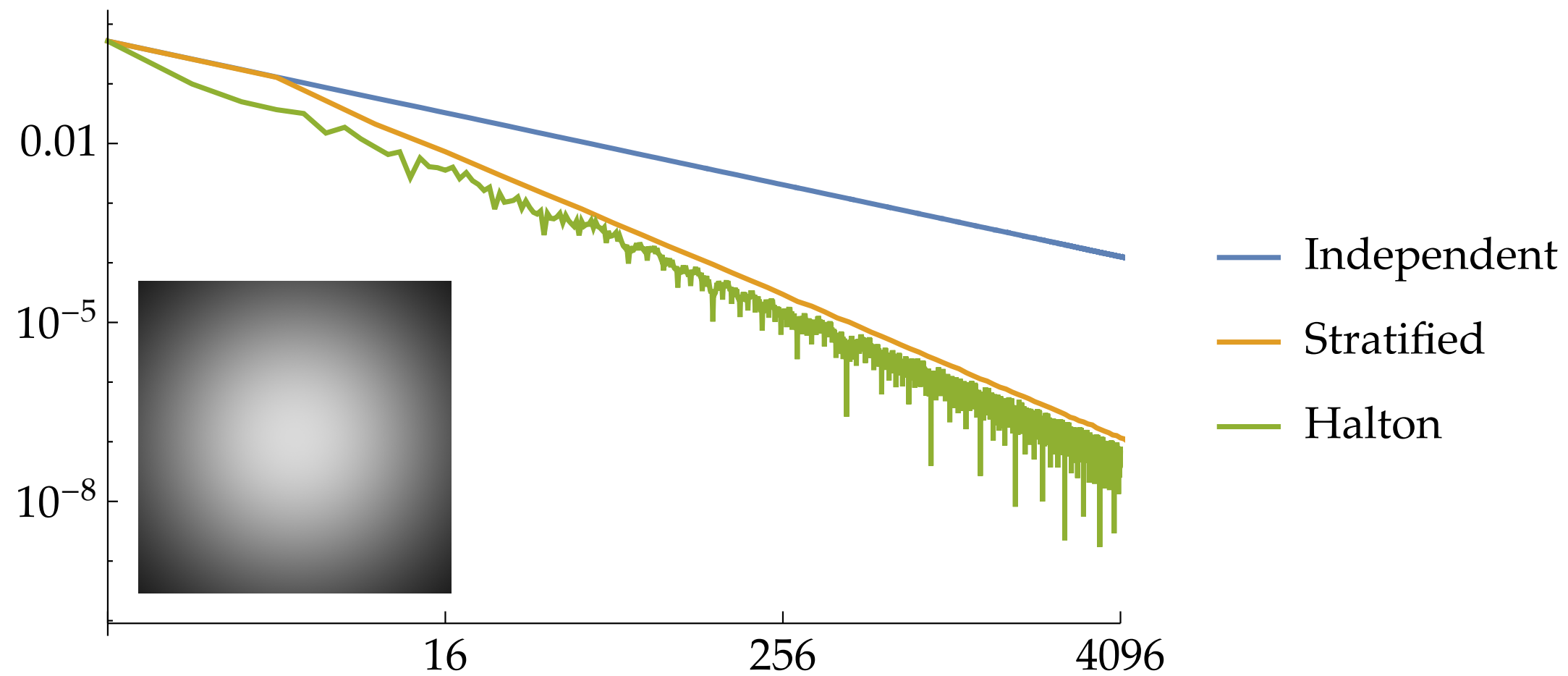
Even worse, $V(f)$ is sometimes unbounded

**We can use this inequality to show that QMC
error converges as:**

$$\sim \frac{(\log N)^d}{N}$$

**(recall that MC variance goes at $O(1/N)$,
so error goes at $O(1/\sqrt{N})$.)**

Measured Quasi-Monte Carlo Error



Low-Discrepancy Point Sequences

The Radical Inverse

Consider the digits of a number n , expressed in base b

$$n = \sum_{i=1}^{\infty} d_i b^{(i-1)}$$

e.g. for $n = 6$ in base 2 , $n=110_2$, and

$$d_1 = 0, d_2 = 1, d_3 = 1, d_i = 0$$

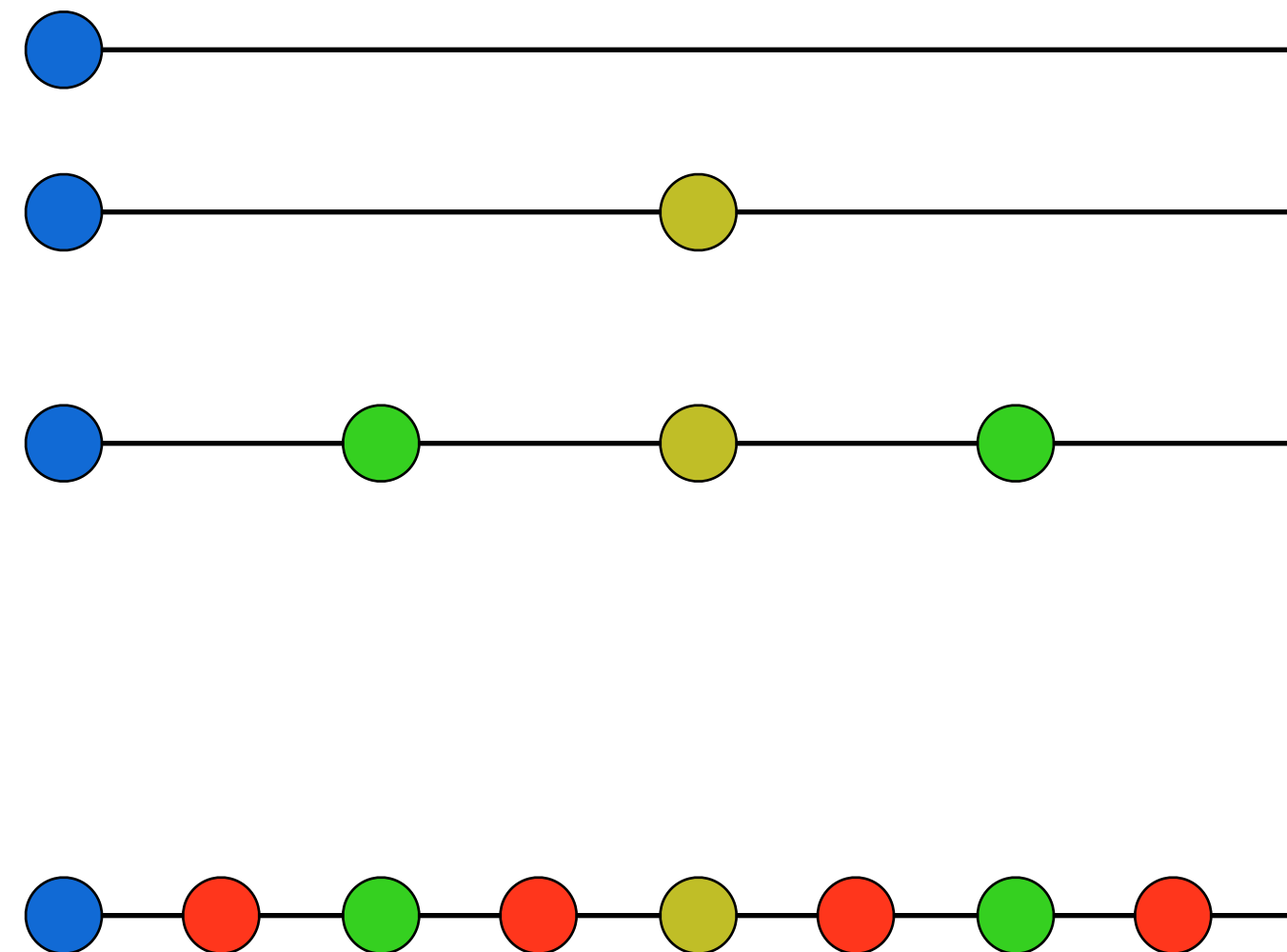
The radical inverse mirrors the digits around the decimal:

$$\Phi_2(6) = 0.011_2 = 0 \cdot 2^{-1} + 1 \cdot 2^{-2} + 1 \cdot 2^{-3} = 0.375$$

$$\Phi_b(n) = \sum_{i=1}^{\infty} d_i b^{-i}$$

1D Low Discrepancy: van der Corput

n	$\Phi_2(n)$
0	0
1	0.5
2	0.25
3	0.75
4	0.125
5	0.625
6	0.375
7	0.875
...	...



Efficient Base 2 Radical Inverse

Assume a fixed number of bits (say 32):

$$\Phi_b(n) = \sum_{i=1}^{32} d_i b^{-i}$$

We have the sum: $d_1 2^{-1} + d_2 2^{-2} + \dots + d_{32} 2^{-32}$

Pull out a factor of 2^{-32} : $2^{-32}(d_1 2^{31} + d_2 2^{30} + \dots + d_{32})$

Can also express in terms of bit shifts:

$$2^{-32}((d_1 \ll 31) + (d_2 \ll 30) + \dots + d_{32})$$

Efficient Base 2 Radical Inverse

$$2^{-32}((d_1 \ll 31) + (d_2 \ll 30) + \cdots + d_{32})$$

We already have the digits in the bits of n

$$n = \sum_{i=1}^{\infty} d_i b^{(i-1)}$$

32	31	30	29	28	27	26	25	24	23	22	21	20	19	18	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1
----	----	----	----	----	----	----	----	----	----	----	----	----	----	----	----	----	----	----	----	----	----	----	---	---	---	---	---	---	---	---	---

So

- **Reverse the bits**
- **Multiply by 2^{-32}**

Reversing Bits in Parallel

```
uint32_t ReverseBits(uint32_t n) {
```

```
    n = (n << 16) | (n >> 16);
```

```
    n = ((n & 0x00ff00ff) << 8) | ((n & 0xff00ff00) >> 8);
```

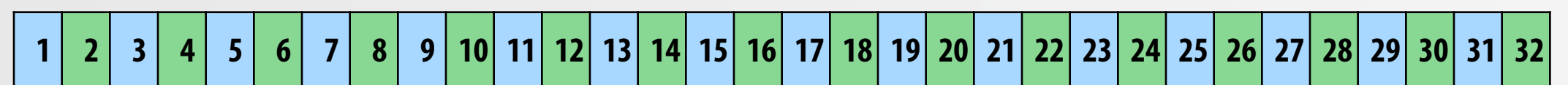
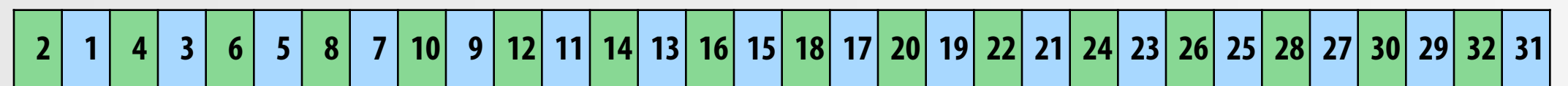
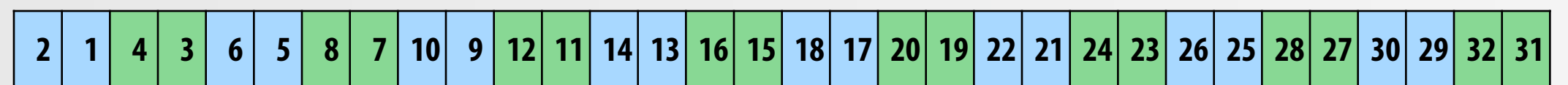
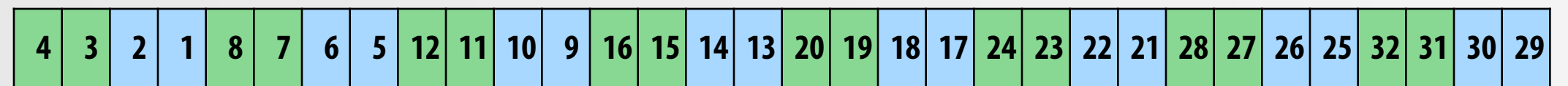
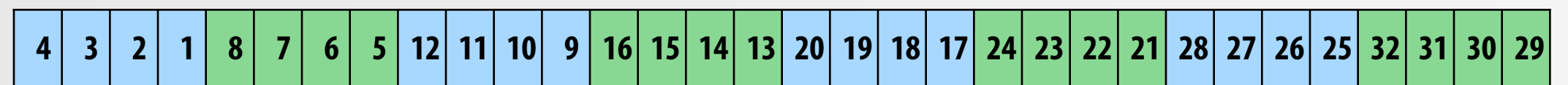
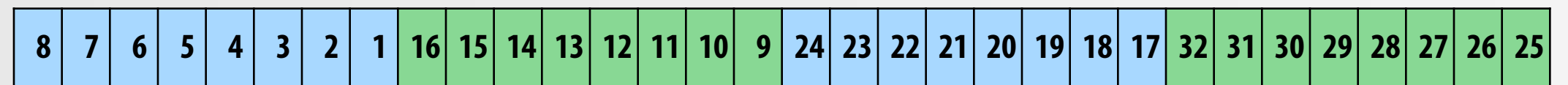
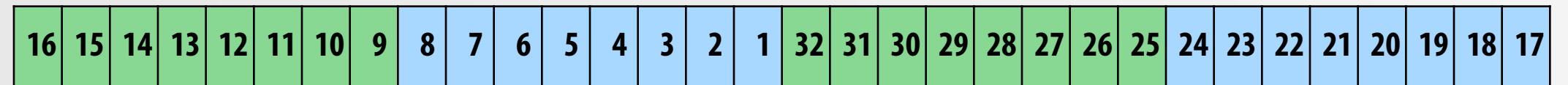
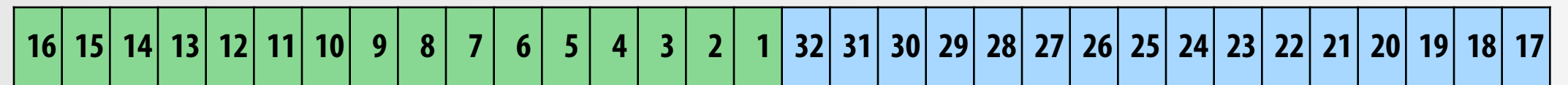
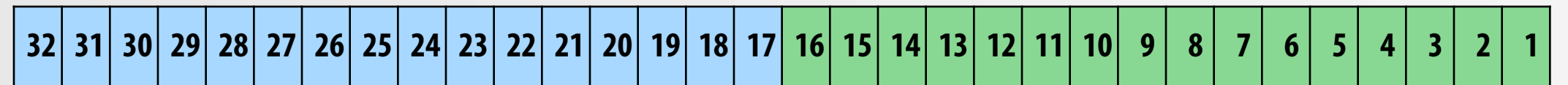
```
    n = ((n & 0x0f0f0f0f) << 4) | ((n & 0xf0f0f0f0) >> 4);
```

```
    n = ((n & 0x33333333) << 2) | ((n & 0xcccccccc) >> 2);
```

```
    n = ((n & 0x55555555) << 1) | ((n & 0xaaaaaaaa) >> 1);
```

```
    return n;
```

```
}
```



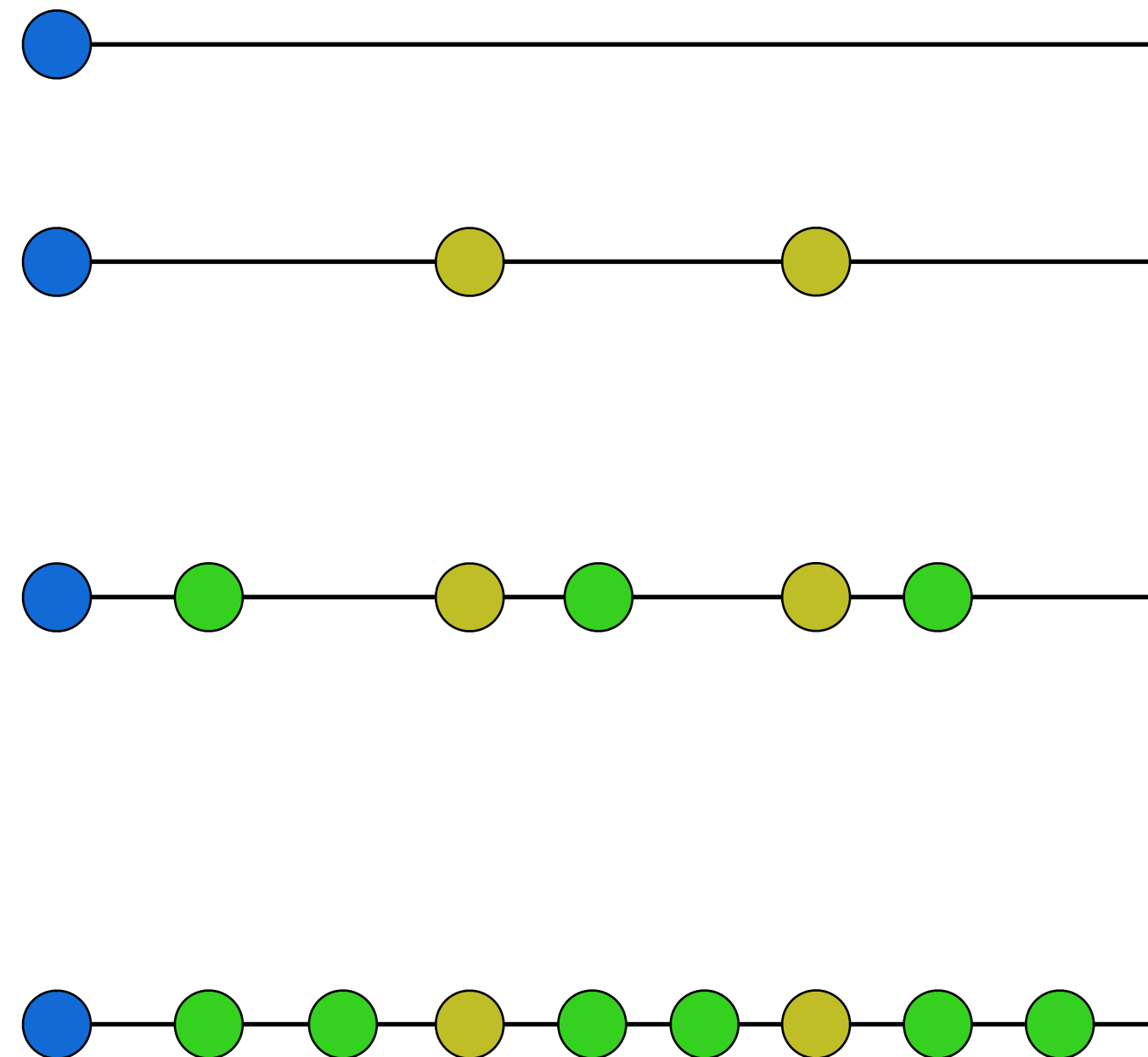
Efficient van Der Corput

```
uint32_t ReverseBits(uint32_t n) {  
    n = (n << 16) | (n >> 16);  
    n = ((n & 0x00ff00ff) << 8) | ((n & 0xff00ff00) >> 8);  
    n = ((n & 0x0f0f0f0f) << 4) | ((n & 0xf0f0f0f0) >> 4);  
    n = ((n & 0x33333333) << 2) | ((n & 0xcccccccc) >> 2);  
    n = ((n & 0x55555555) << 1) | ((n & 0xaaaaaaaa) >> 1);  
    return n;  
}
```

```
float RadicalInverse2(uint32_t v) {  
    v = ReverseBits(v);  
    const float Inv2To32 = 1.f / (1ull << 32);  
    return v * Inv2To32;  
}
```

Radical Inverse Base 3

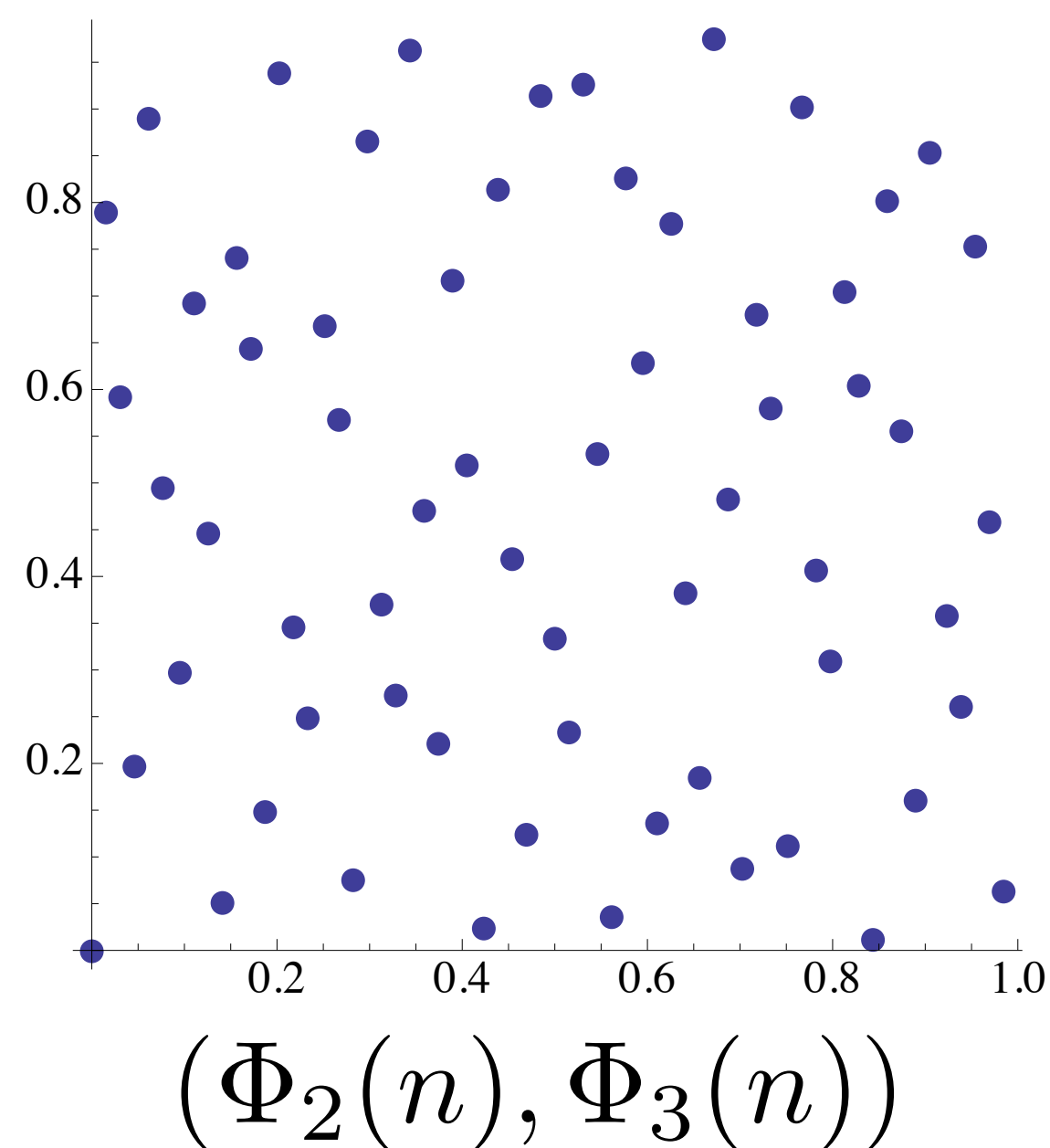
n	$\Phi_3(n)$
0	0
1	0.333..
2	0.666...
3	0.111...
4	0.444...
5	0.777...
6	0.222...
7	0.555...
8	0.888...



The Halton Sequence

Low discrepancy sequence $(\Phi_{b_1}(n), \Phi_{b_2}(n), \Phi_{b_3}(n), \dots)$

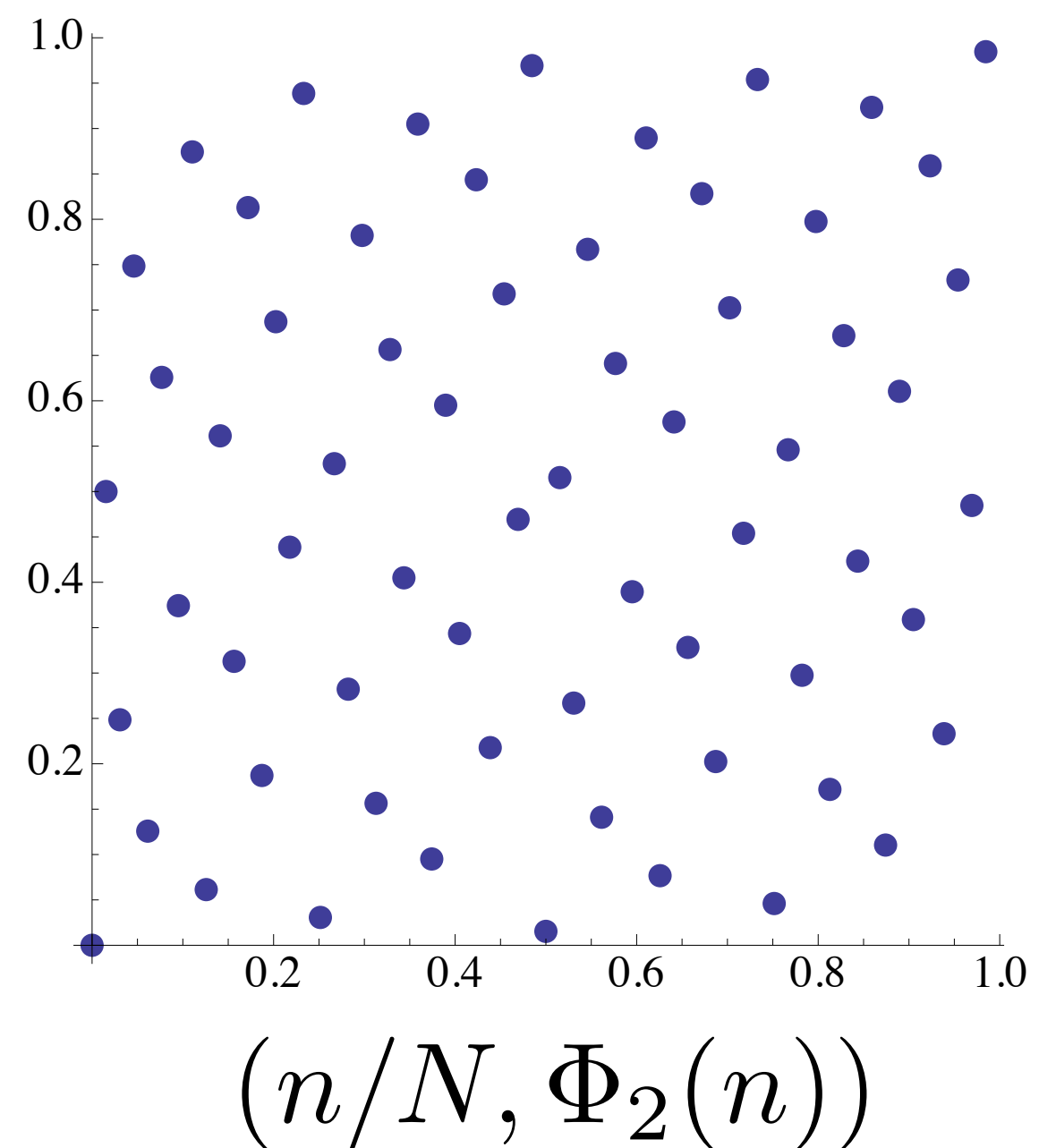
- **The dimensions' bases are relatively prime**
- **Arbitrary number of dimensions**
- **Arbitrary number of points**



The Hammersley Point Set

**If the number of points N is known in advance,
set one dimension to n/N**

$$(n/N, \Phi_{b_1}(n), \Phi_{b_2}(n), \dots)$$

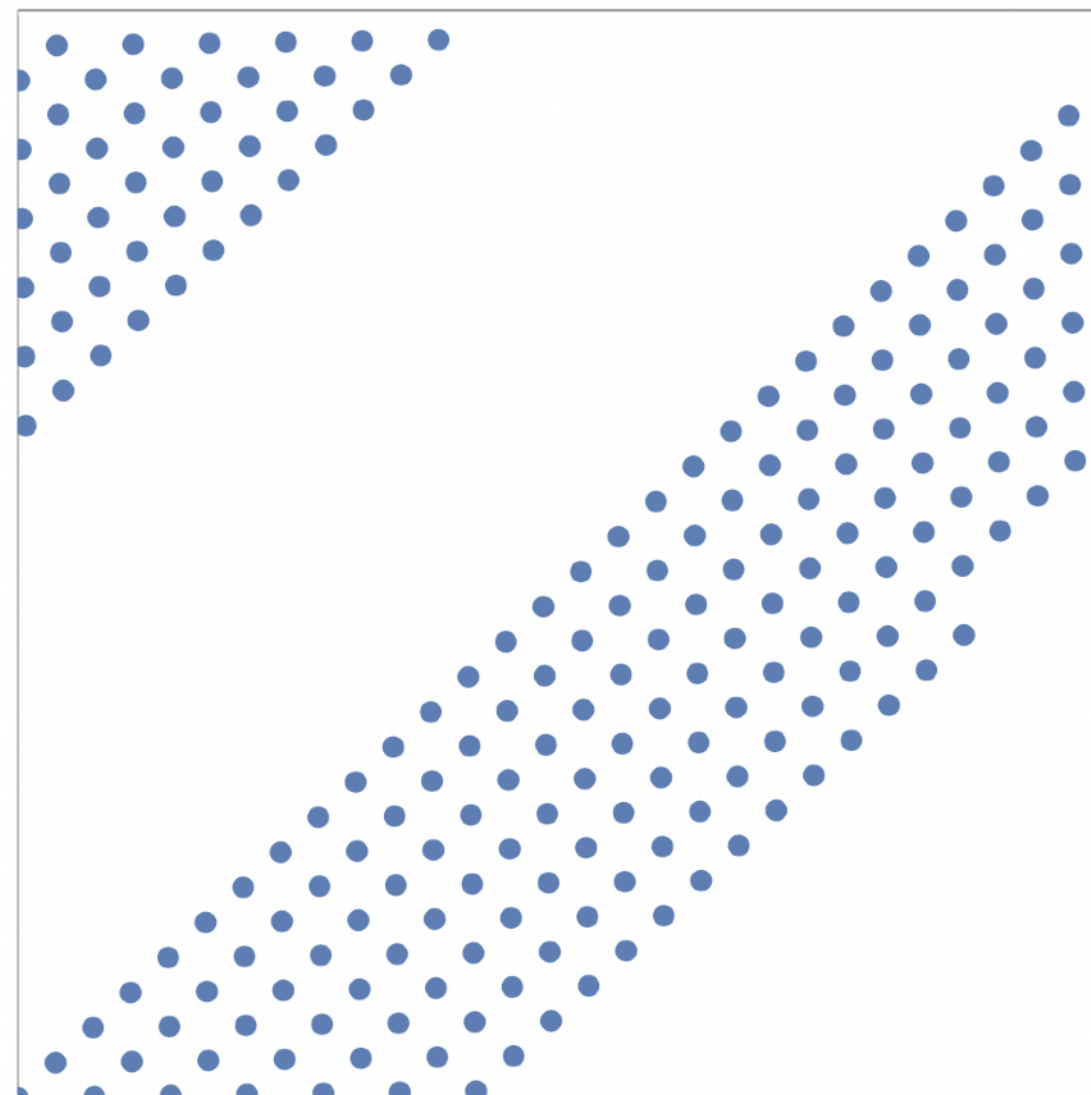


Slightly lower discrepancy than Halton

Low-Dimensional Projections...

Caution: 2D projections of higher bases may not be great

- **The overall pattern remains low-discrepancy over all dimensions, though**



$$(\Phi_{29}(n), \Phi_{31}(n))$$

Randomized Low Discrepancy

$$\Phi_b(n) = \sum_{i=1}^{\infty} d_i b^{-i}$$

Radical Inverse

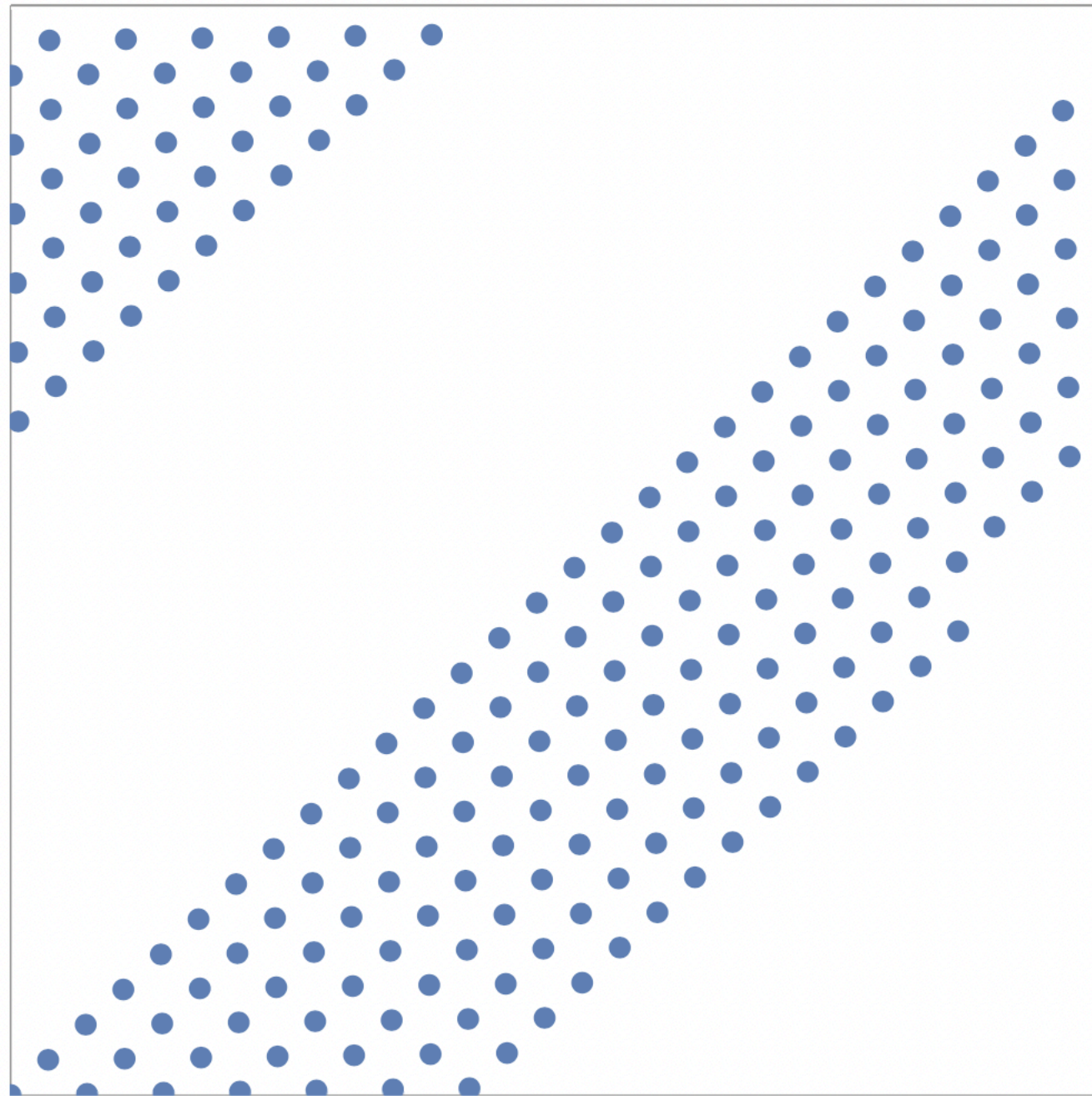
$$\Phi'_b(n) = \sum_{i=1}^{\infty} \sigma_i(d_i) b^{-i}$$

Permuted Radical Inverse

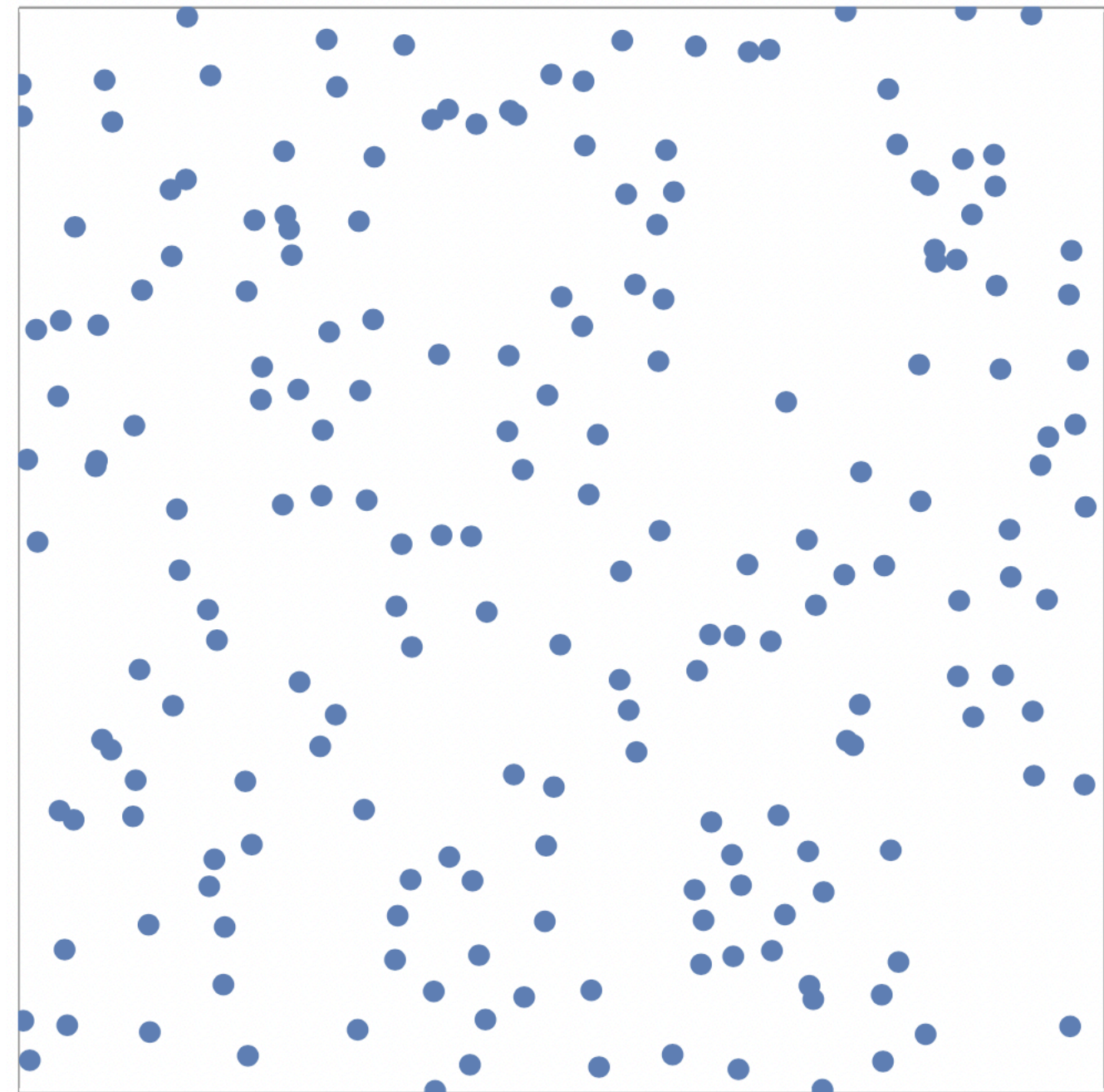
σ_i are (random) permutations of the digits

■ Random permutations maintain LD

Halton + Random Digit Permutations



Unscrambled



Scrambled

$$(\Phi_{29}(n), \Phi_{31}(n))$$

Owen Scrambling

Apply random digit permutations that depend on previous digits

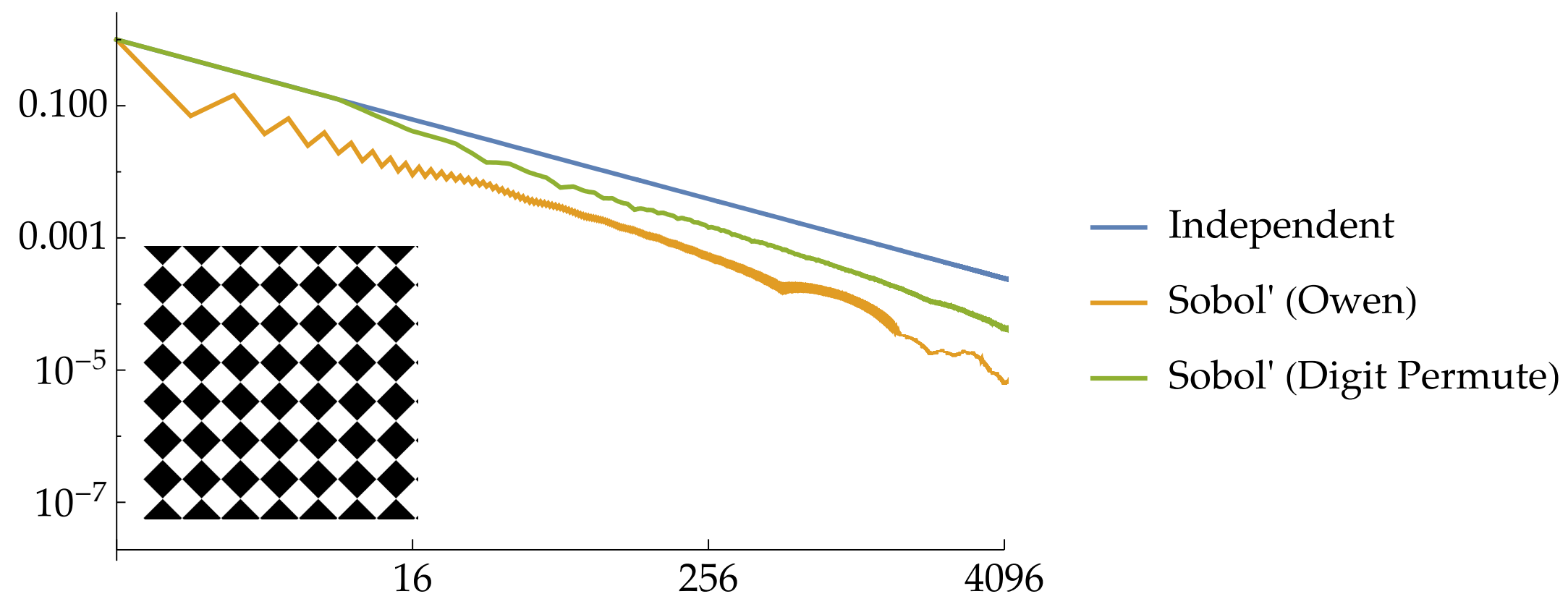
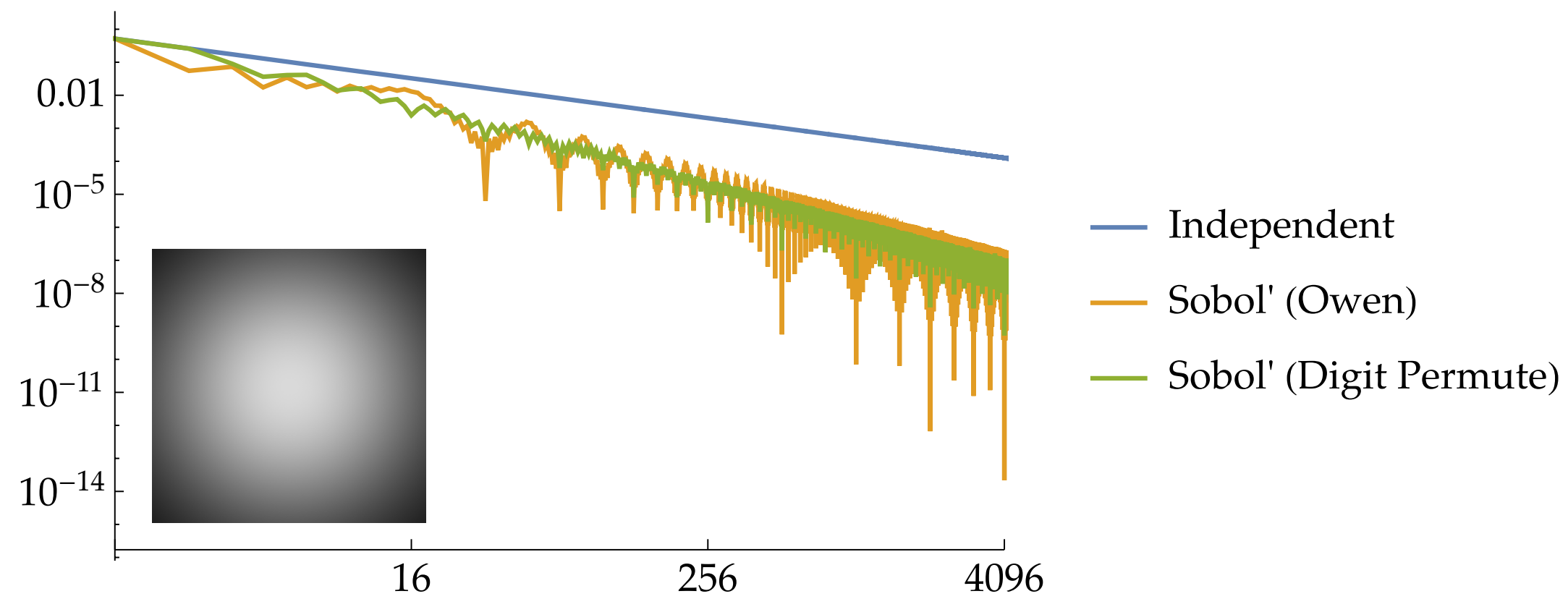
$$\Phi'_b(n) = \sum_{i=1}^{\infty} \sigma_i(d_i) b^{-i}$$

Permuted Radical Inverse

$$\Phi'_b(n) = \sum_{i=1}^{\infty} \sigma_{\{d_1, \dots, d_{i-1}\}}(d_i) b^{-i}$$

Owen Scrambled Radical Inverse

Error With Owen Scrambling



Sobol' Point Sets

Generator Matrices

Given a base b and a matrix C , define:

$$c(n) = (b^{-1}, b^{-2}, \dots, b^{-m}) C \begin{pmatrix} d_1 \\ d_2 \\ \vdots \\ d_m \end{pmatrix}$$

- where d_i are the base- b digits of n
- and arithmetic is done over the ring \mathbb{Z}_b
 - For our purposes, just do everything “mod b ”

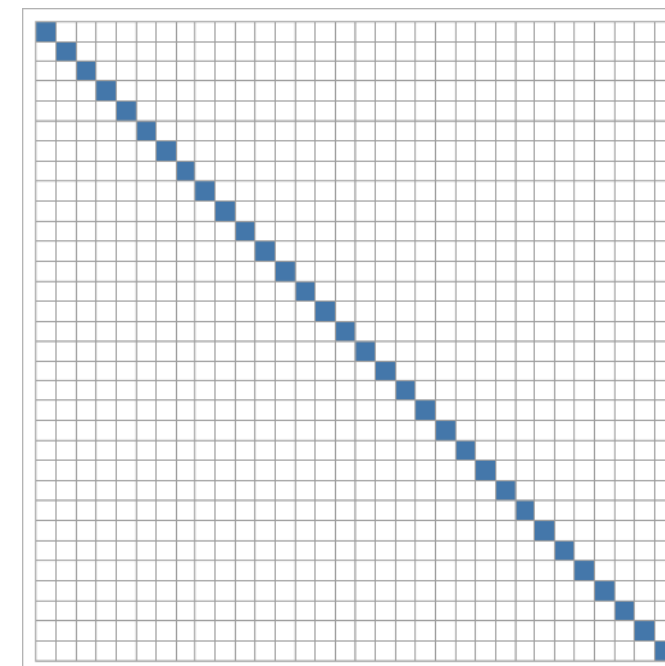
This generates a set of b^m points

Generator Matrices

We'll focus only on $b=2$, which allows particularly efficient implementation

$$c(n) = (2^{-1}, 2^{-2}, \dots, 2^{-m}) C \begin{pmatrix} d_1 \\ d_2 \\ \vdots \\ d_m \end{pmatrix}$$

Quiz: what happens if $C =$

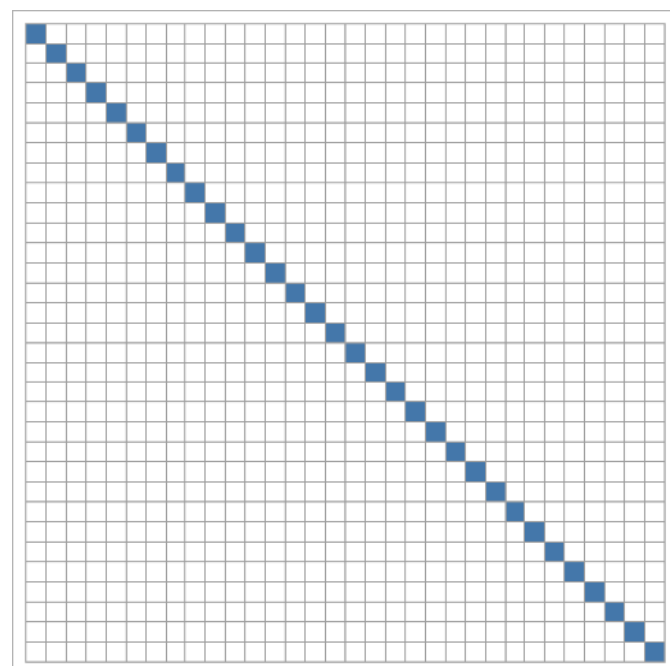


?

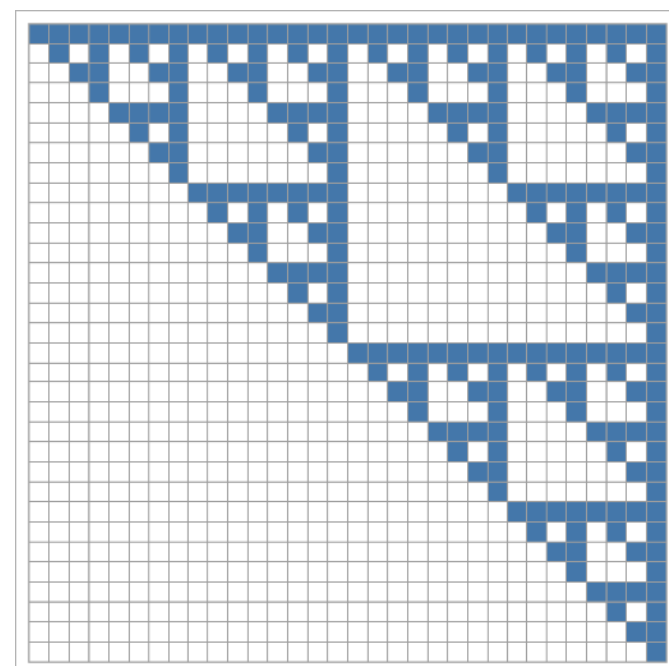
Sobol' Point Sets

Sobol' showed how to find generator matrices for LD point sets in base 2

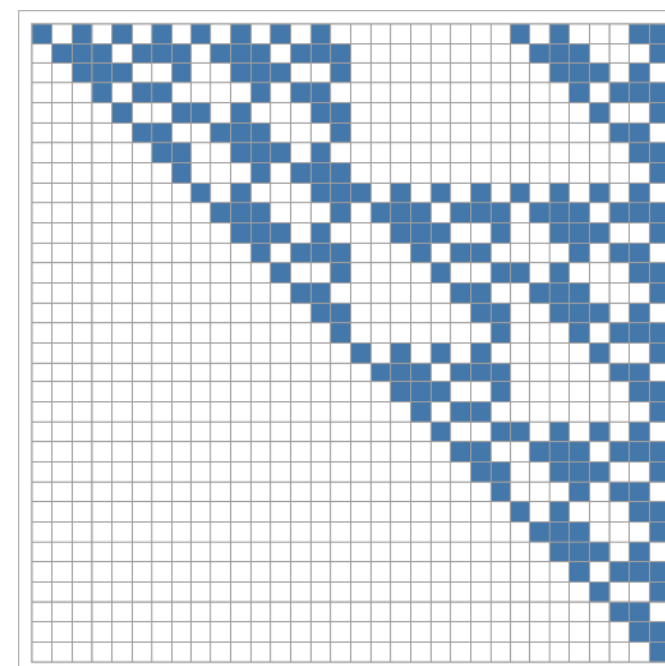
- **Can scale low-discrepancy samples in 1000s of dimensions**



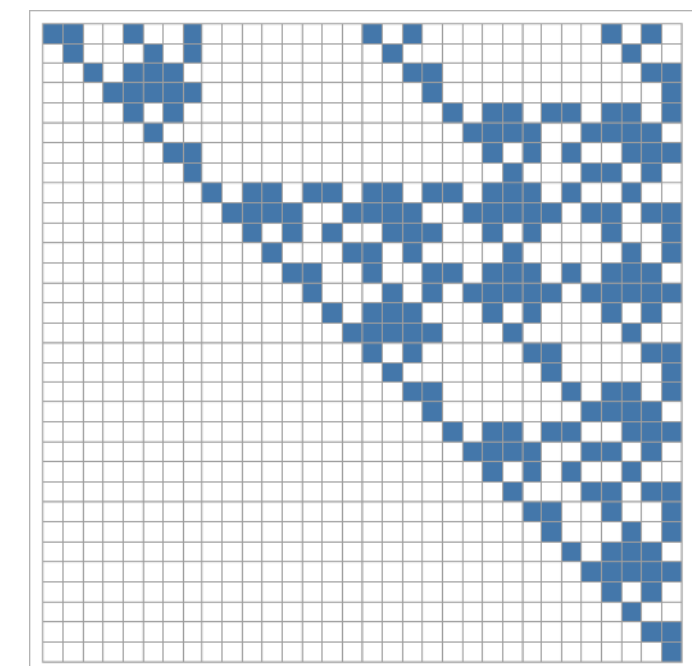
C_0



C_1



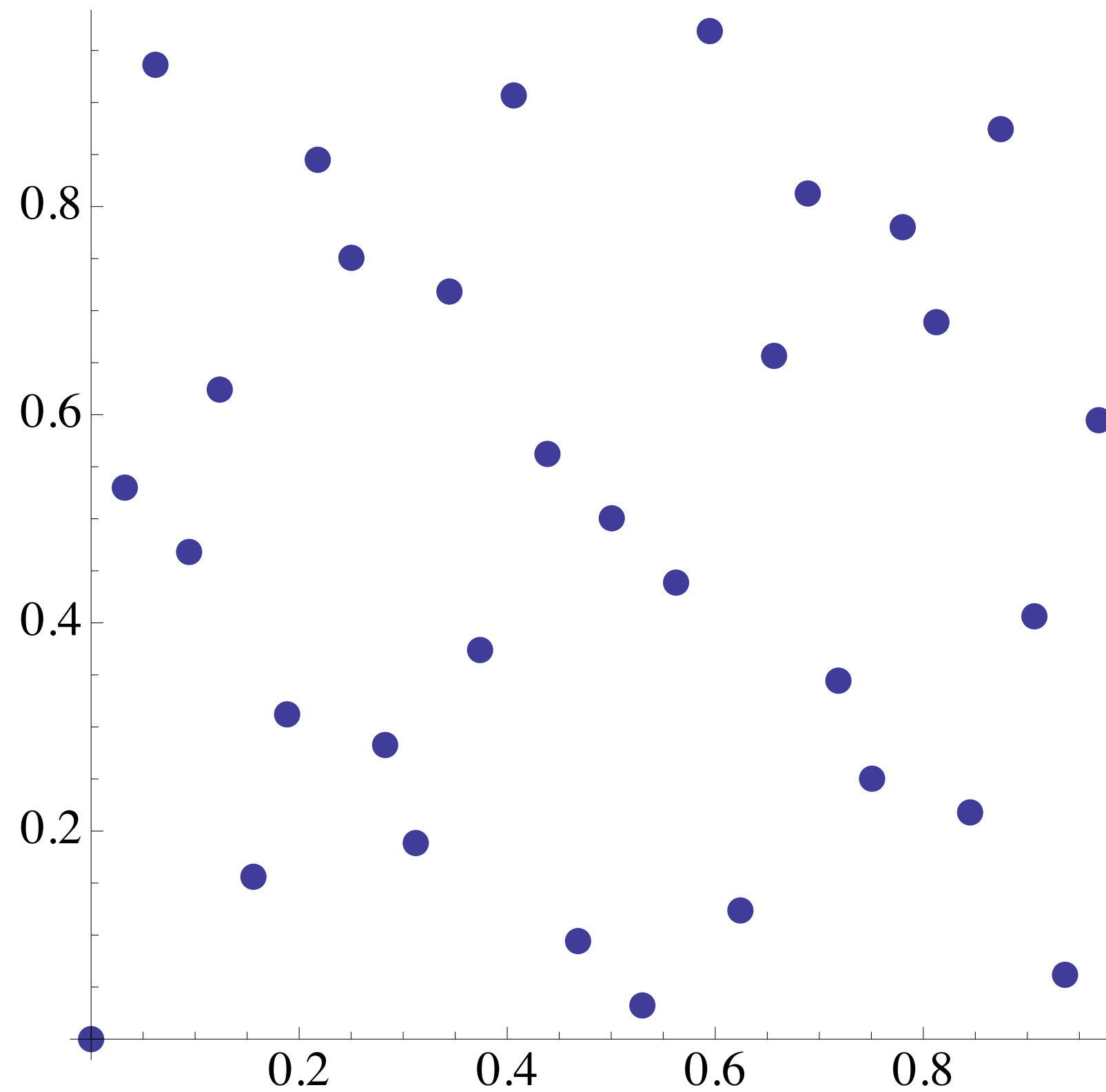
C_2



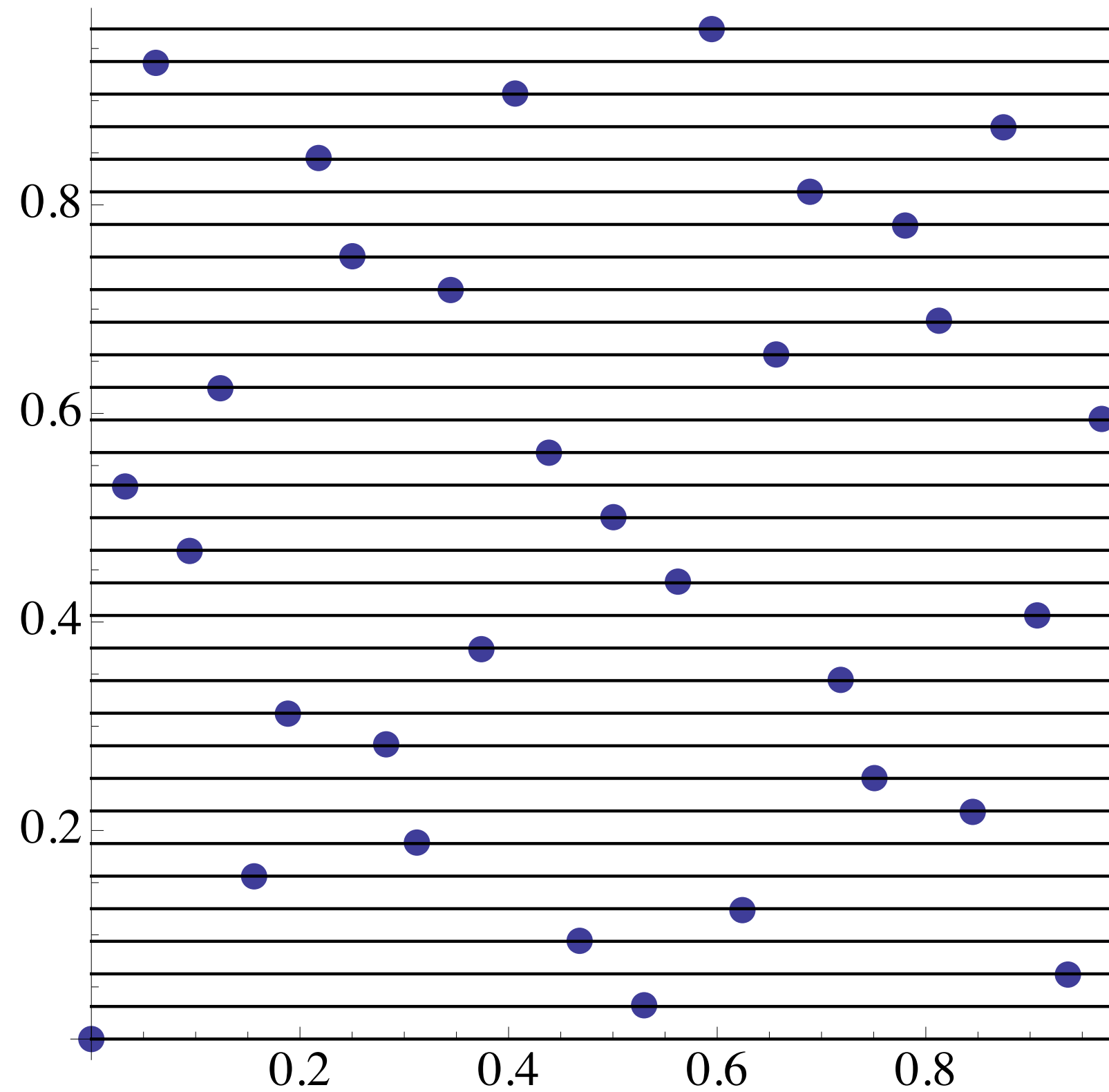
C_3

...

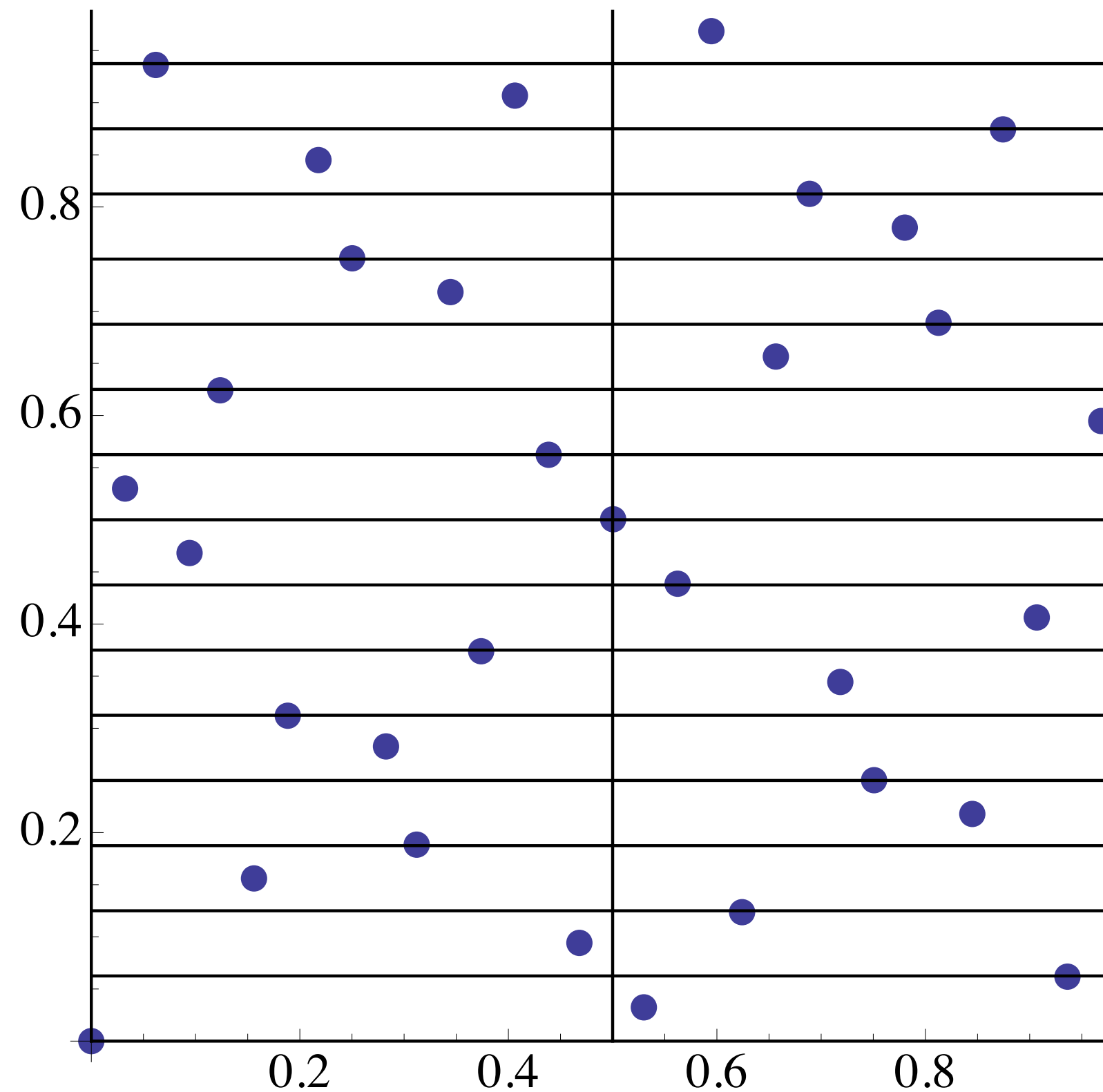
32 2D Sobol' Points



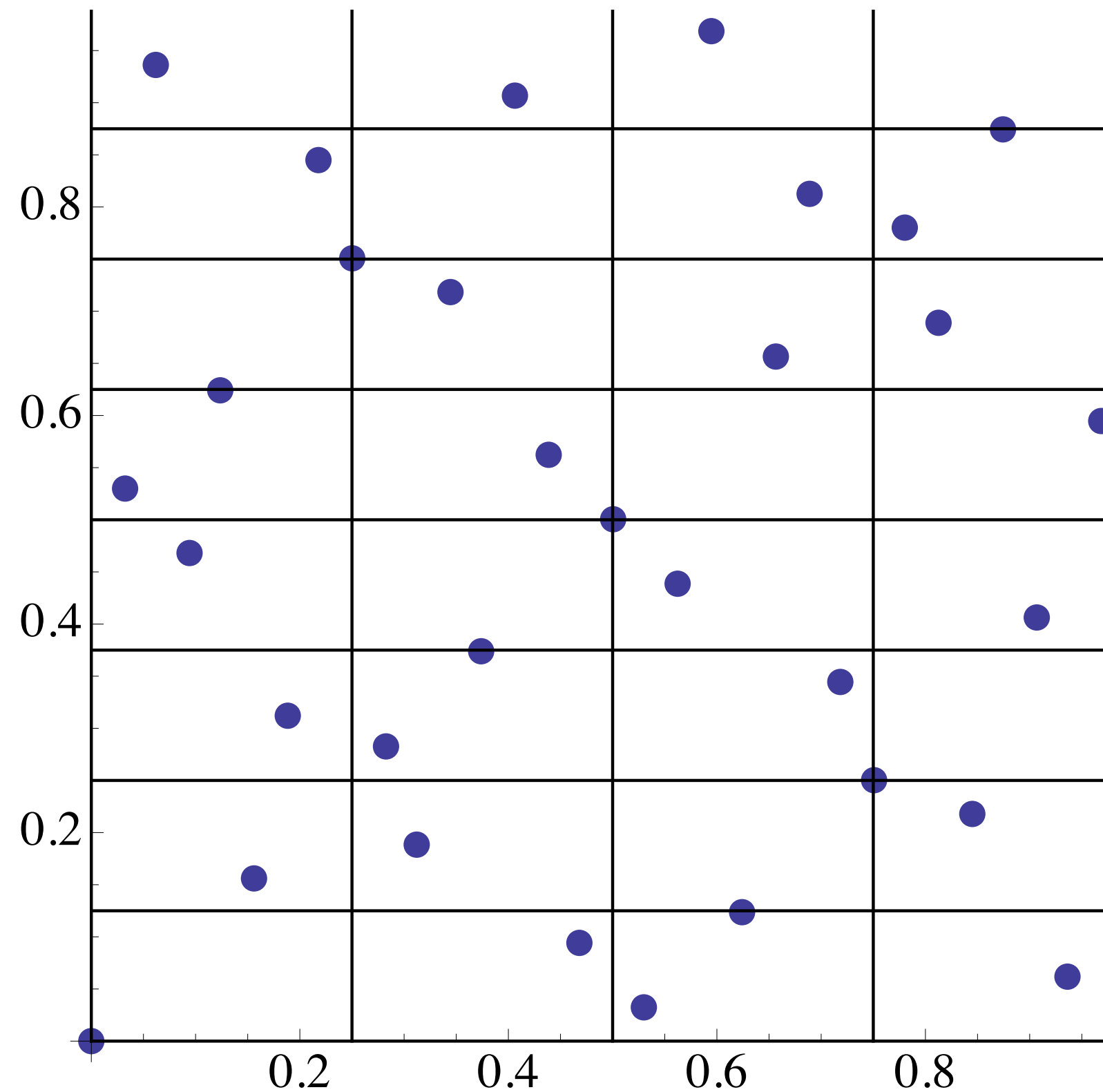
Elementary Intervals (1x32)



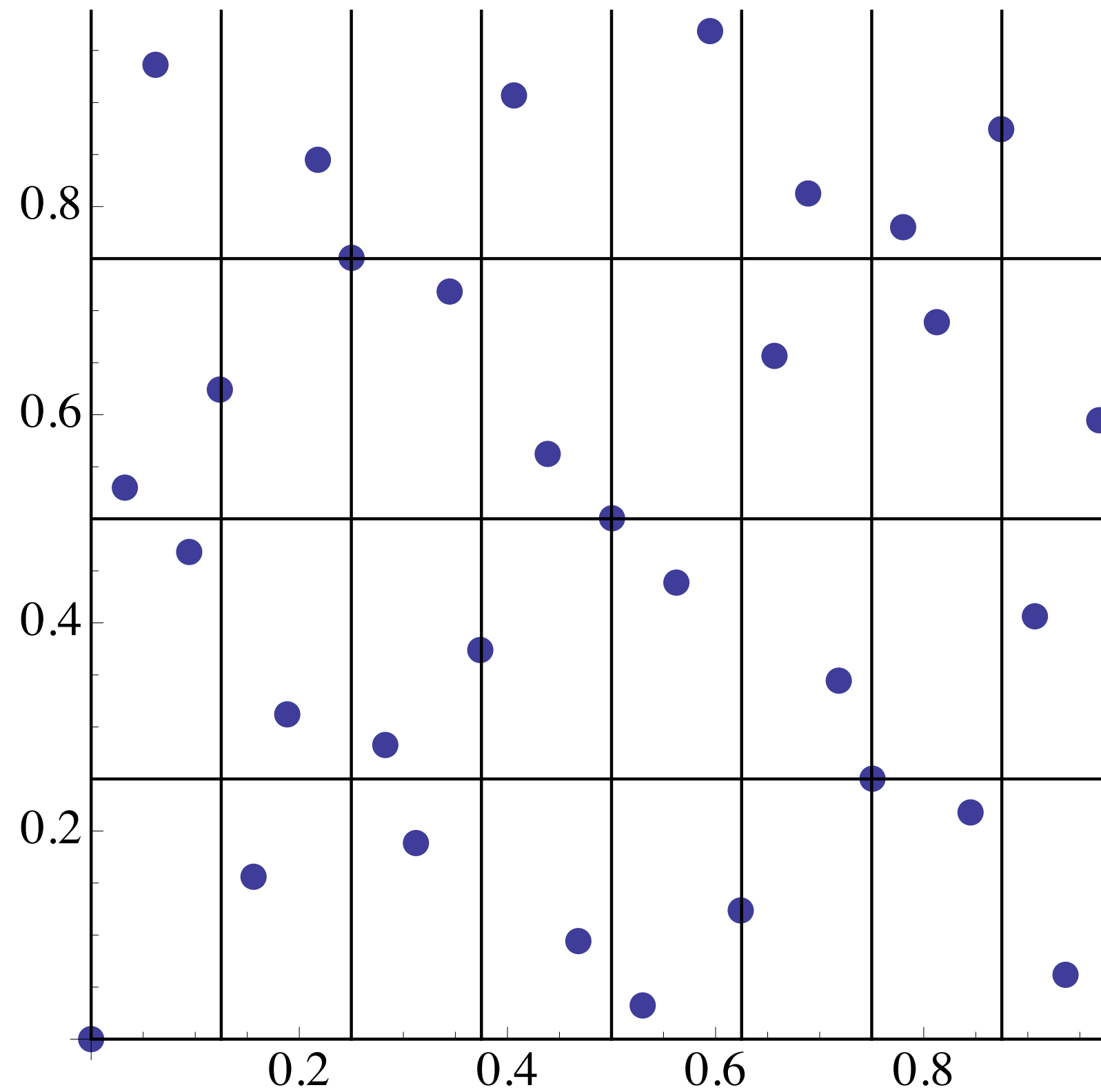
Elementary Intervals (2x16)



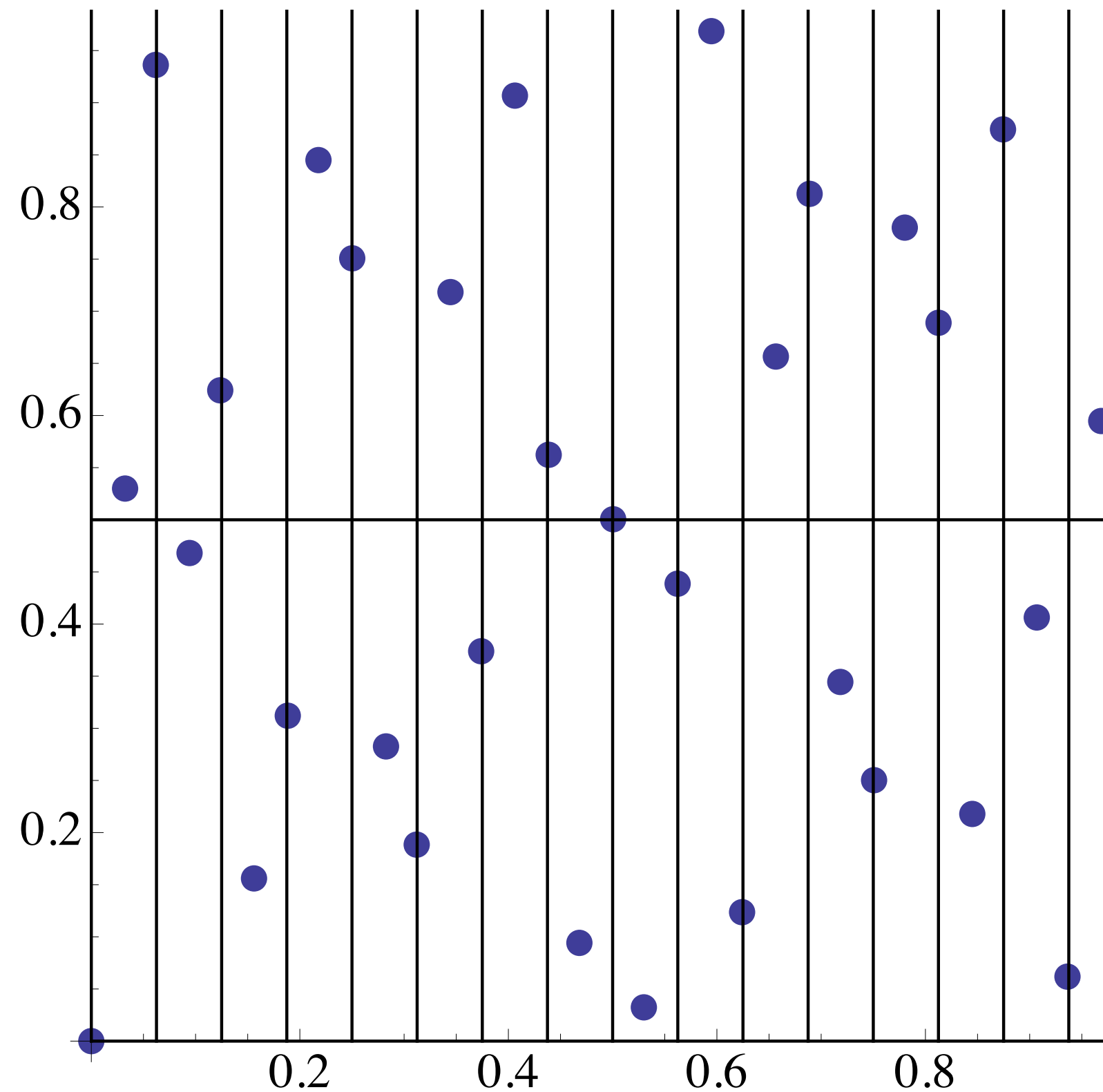
Elementary Intervals (4x8)



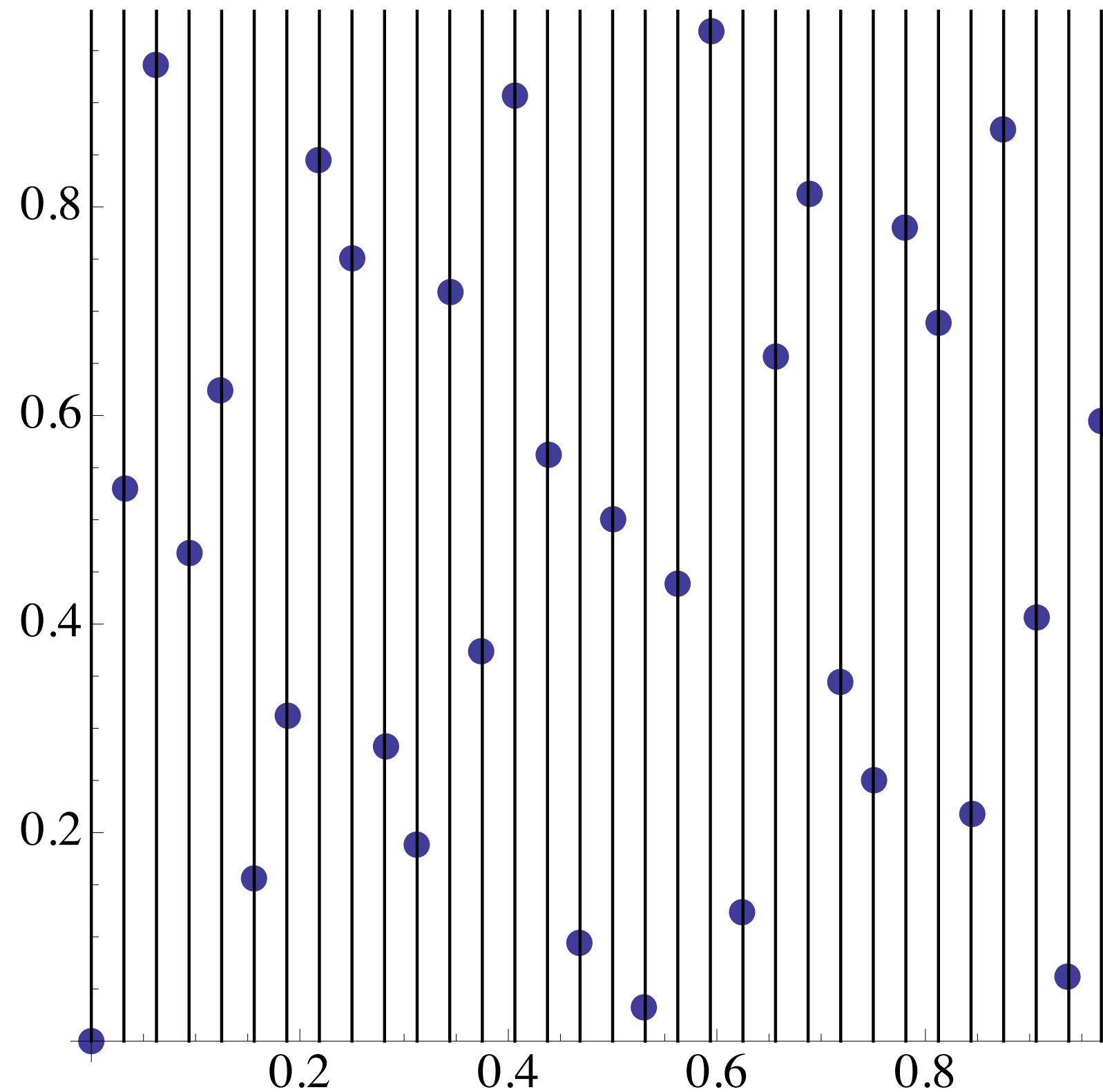
Elementary Intervals (8x4)



Elementary Intervals (16x2)



Elementary Intervals (32x1)





Independent Random Samples, $n=16$
MSE 1x



Stratified Samples, $n=16$
MSE $1/2.41x$



Sobol' Samples, $n=16$
MSE 1/3.38x



Independent Random Samples, $n=16$
MSE 1x

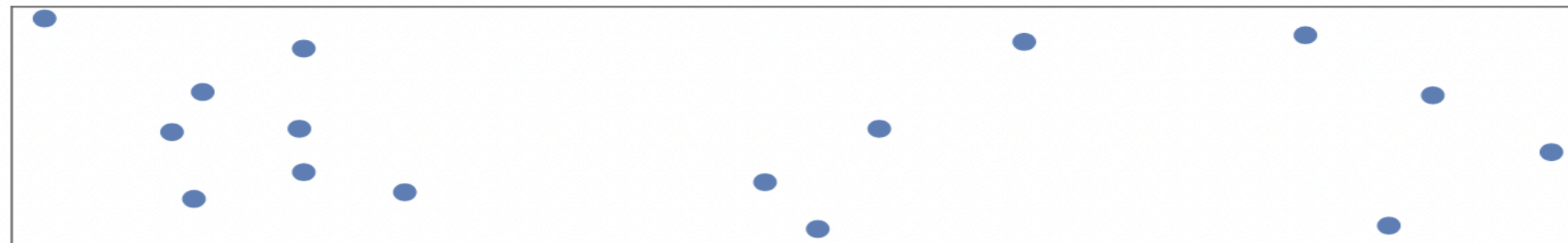


Stratified Samples, $n=16$
MSE $1/2.12x$

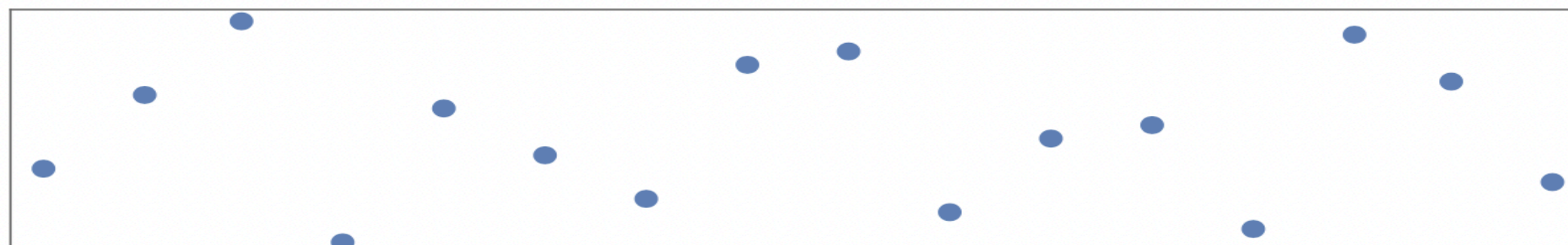


Sobol' Samples, $n=16$
MSE 1/3.95x

Warping Samples to a Quad Light

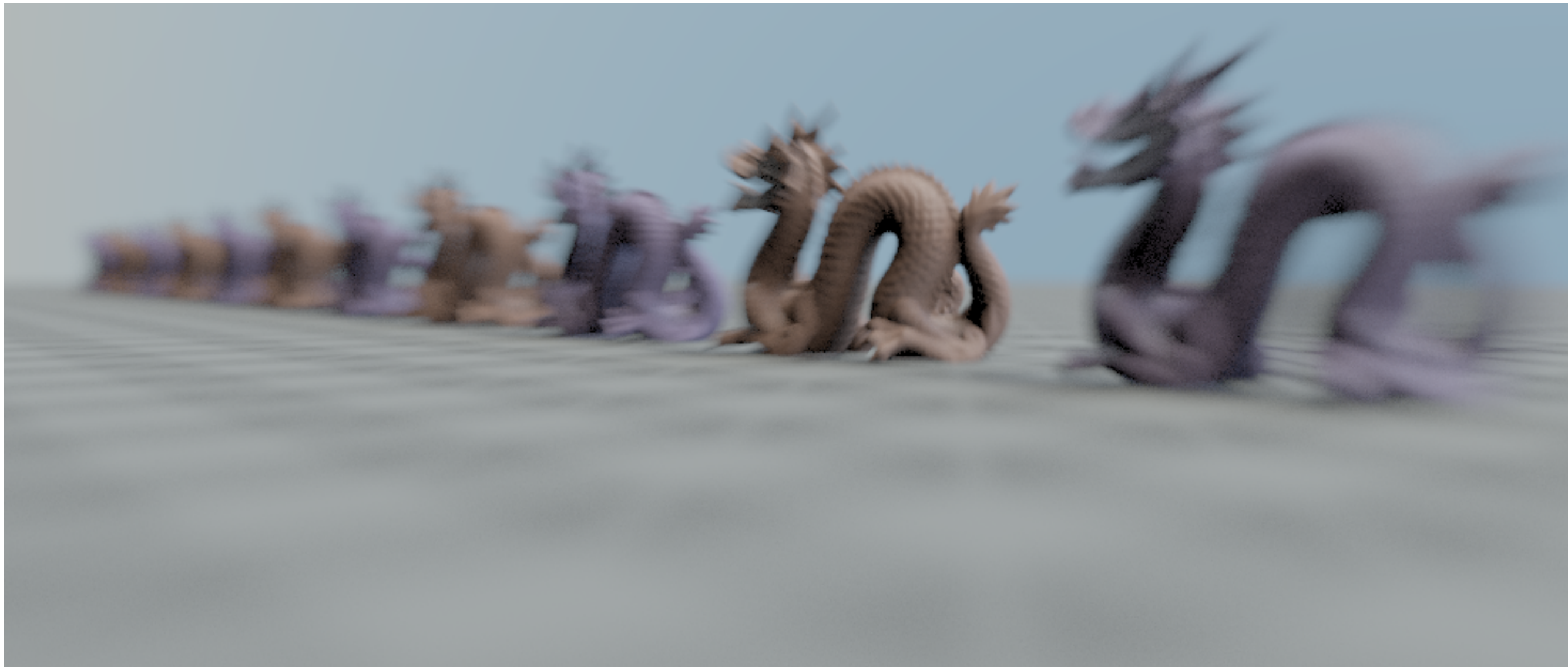


4x4 Stratified



16 Sobol'

Sampling Motion Blur + Defocus



Independent
MSE 1x



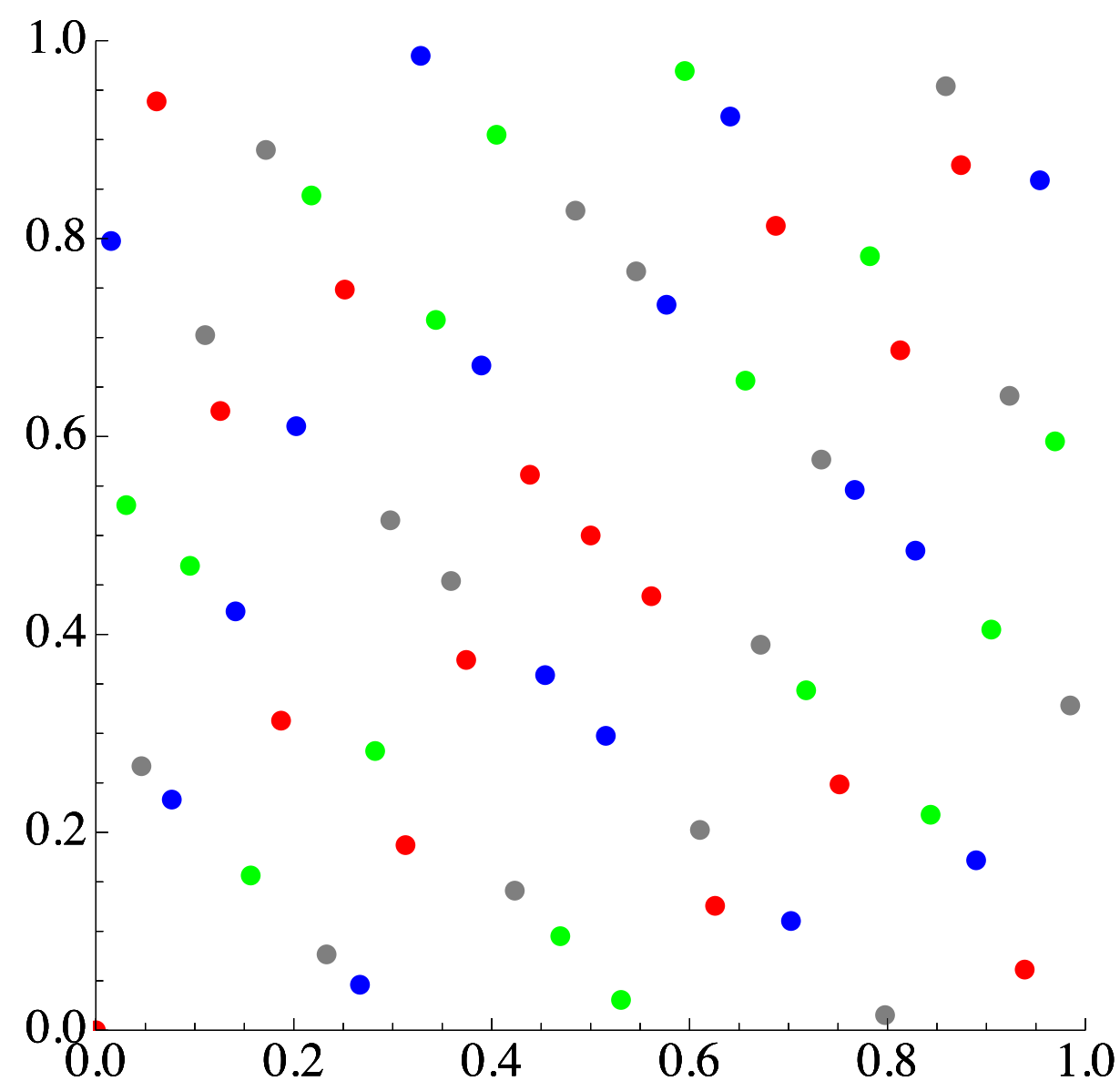
Halton
MSE 1/1.13x



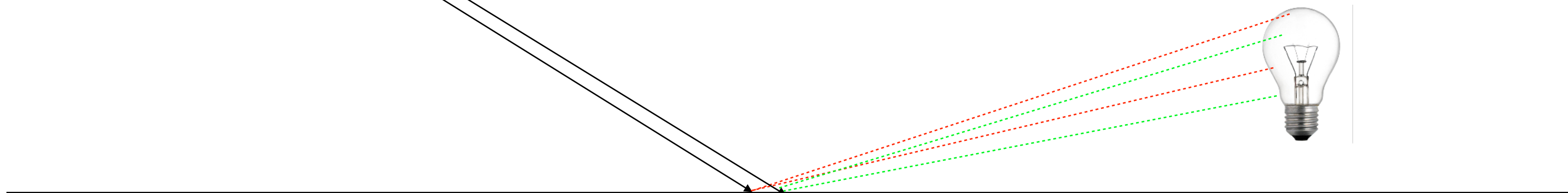
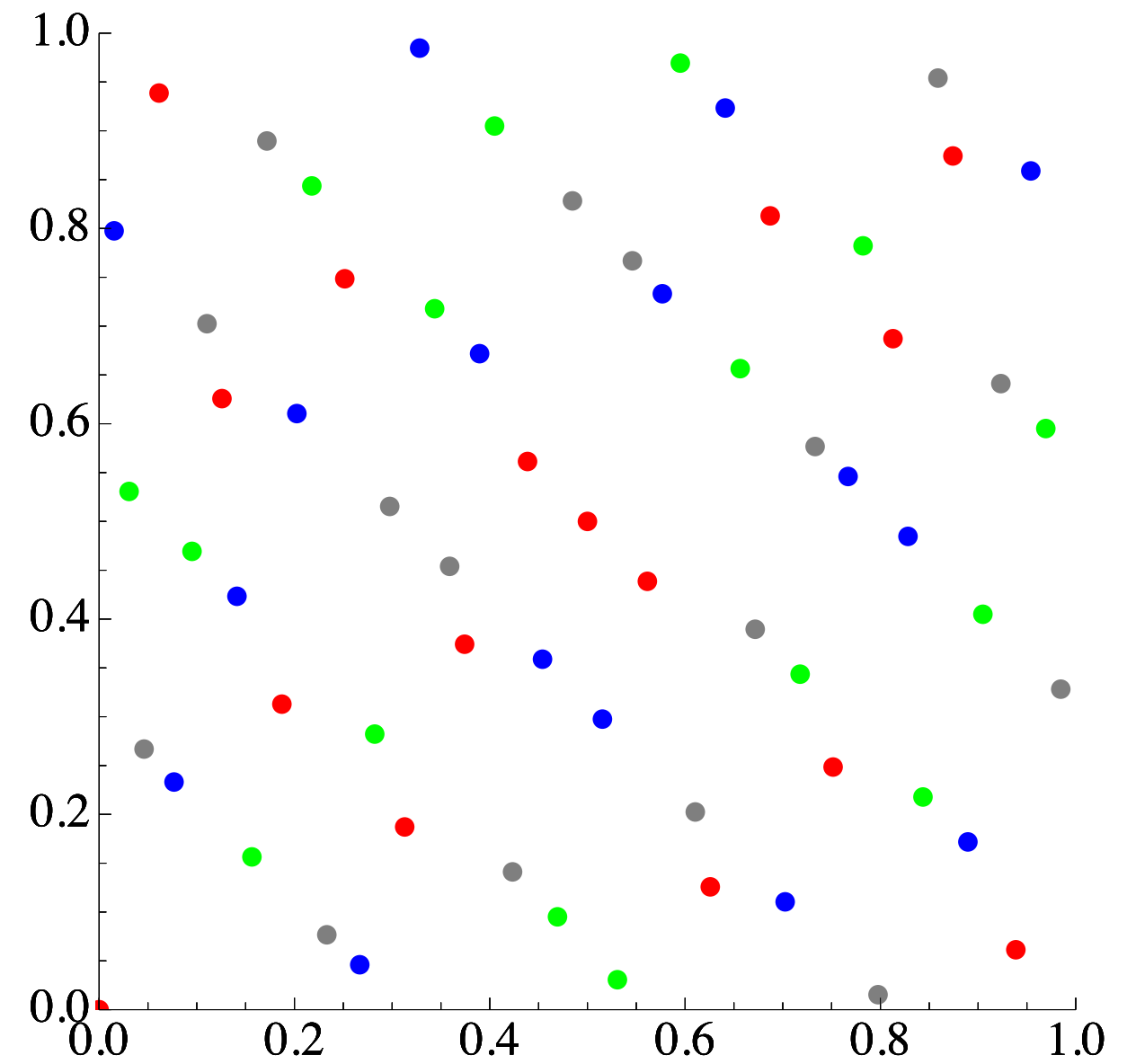
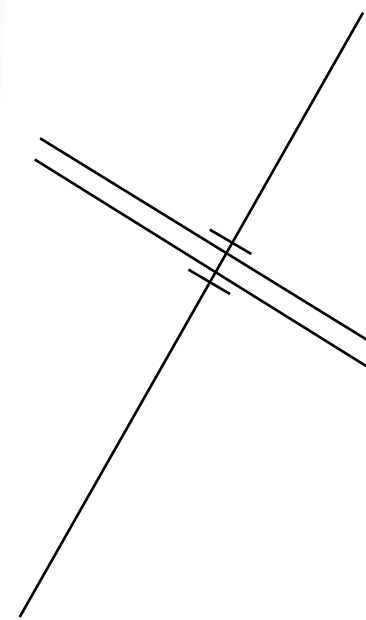
Sobol'
MSE 1/1.80x

(0,2)-sequences

In addition to satisfying general stratification properties, power-of-two length subsequences are well-distributed with respect to each other



Pixel * Light Sampling



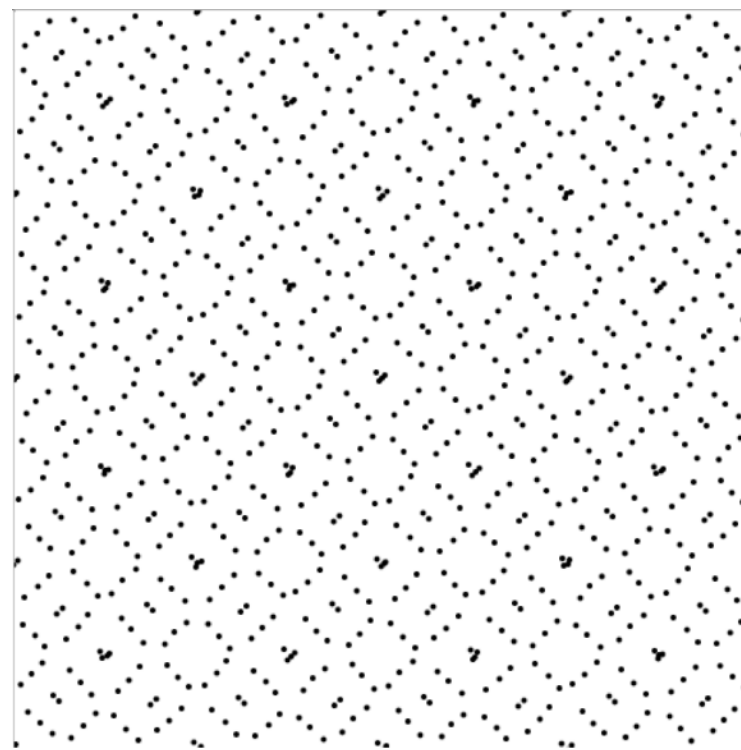
Spectral Analysis of Sampling

Measuring Point Set Quality

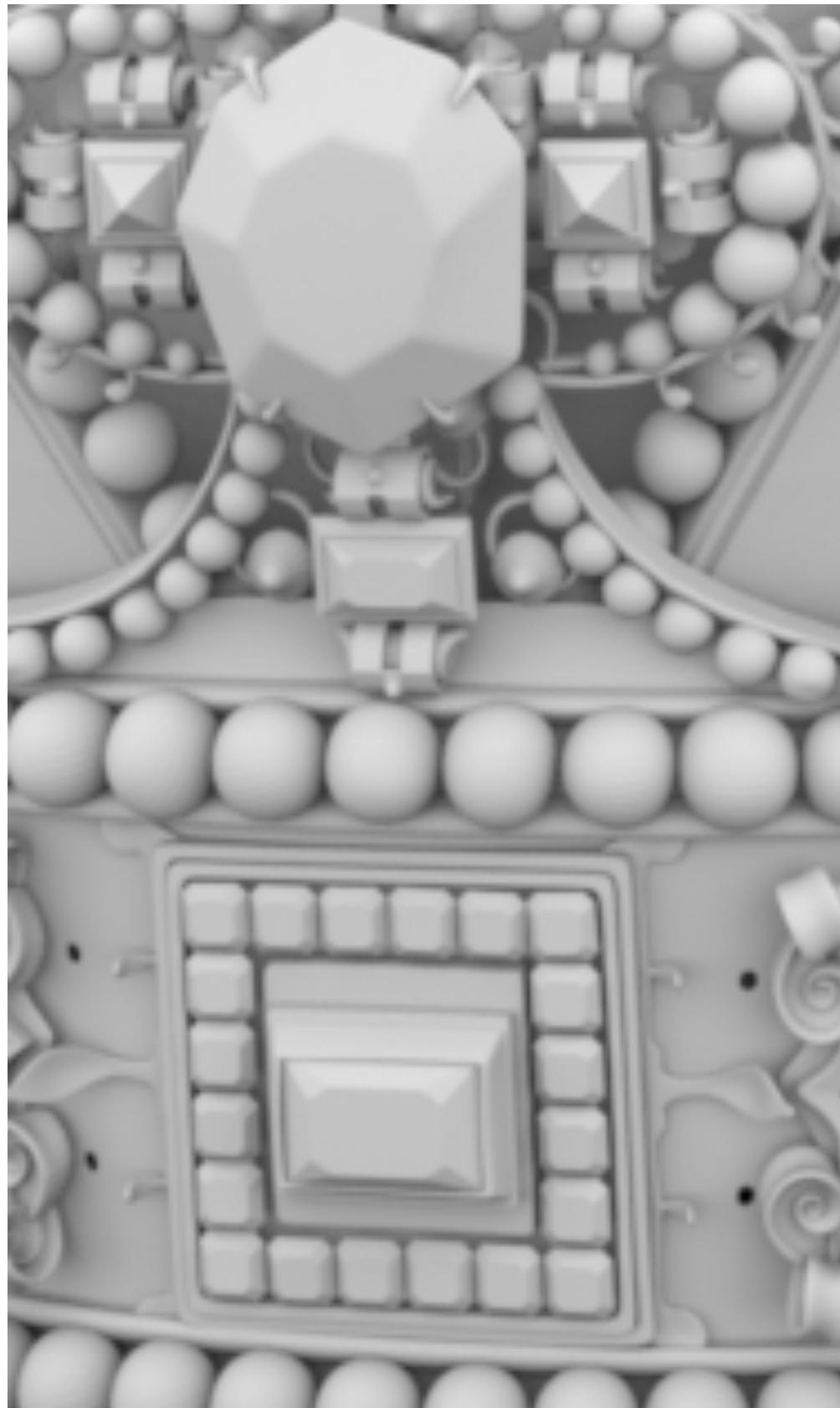
Some problems with discrepancy:

- **Anisotropic: rotating the points changes discrepancy**
- **Not shift-invariant: similarly for translation**
- **Doesn't account for human perception**

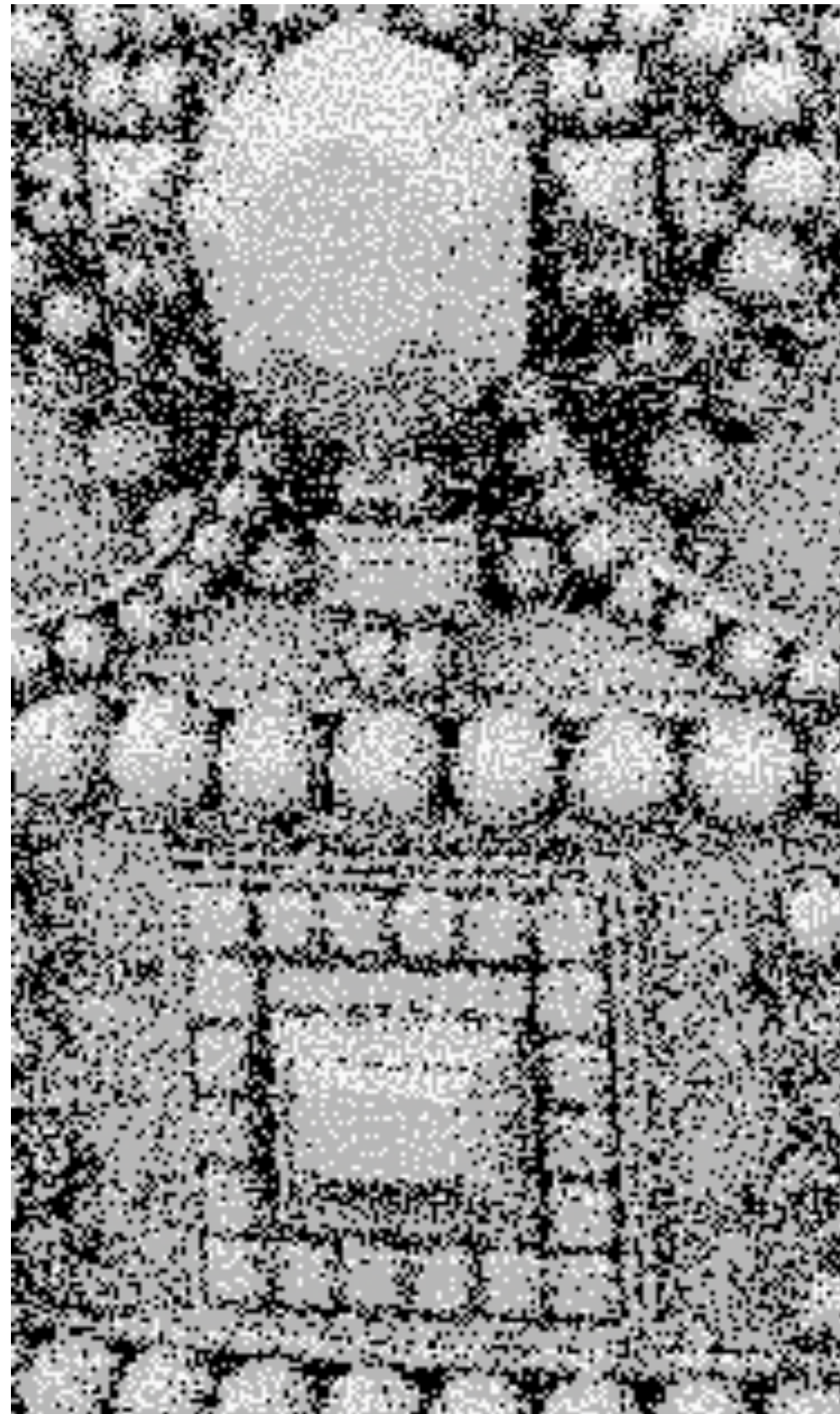
In general, can have low discrepancy yet still have points clumped together:



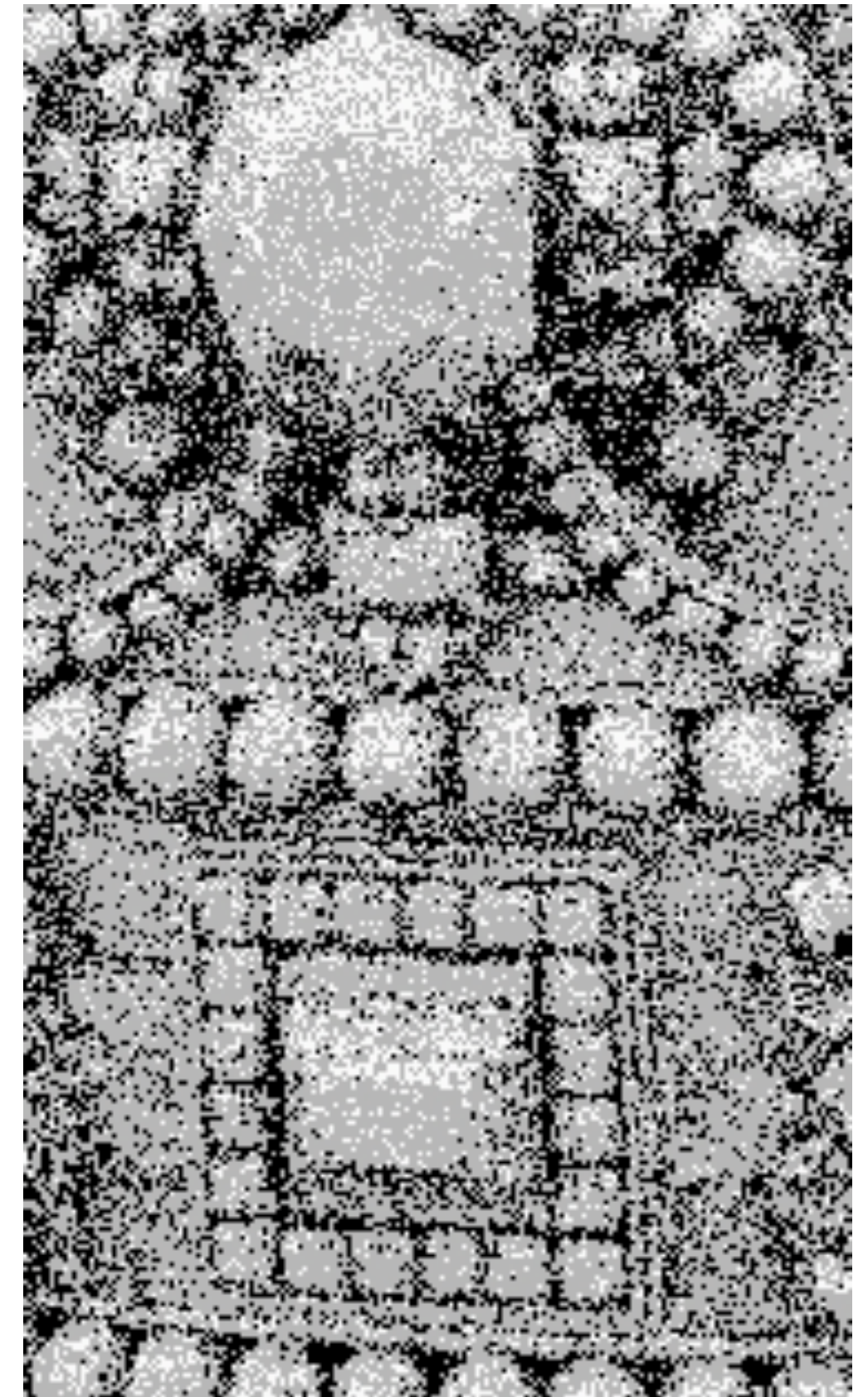
Ambient Occlusion: $\int_{\Omega} V(\omega) \cos \theta \, d\omega$



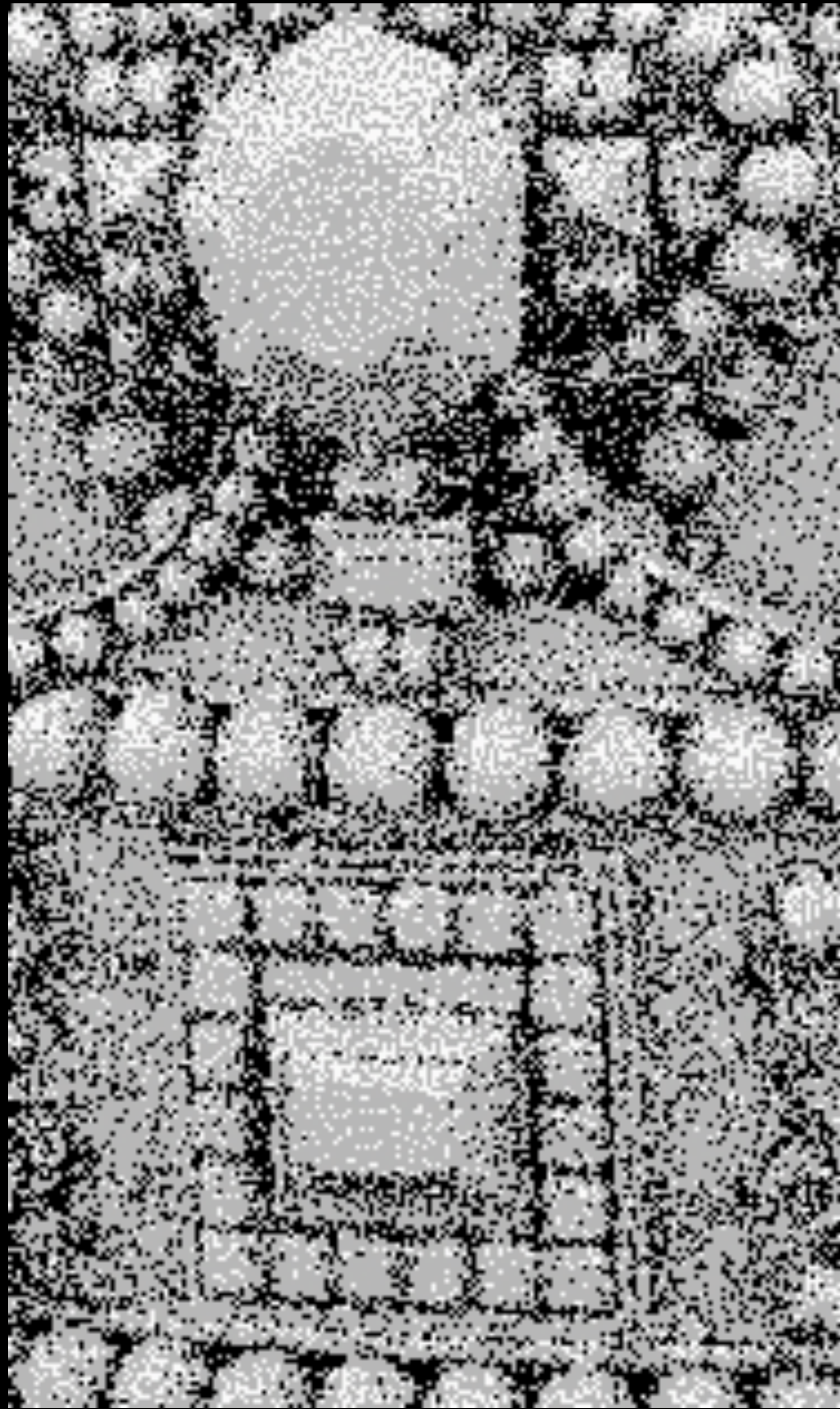
Reference

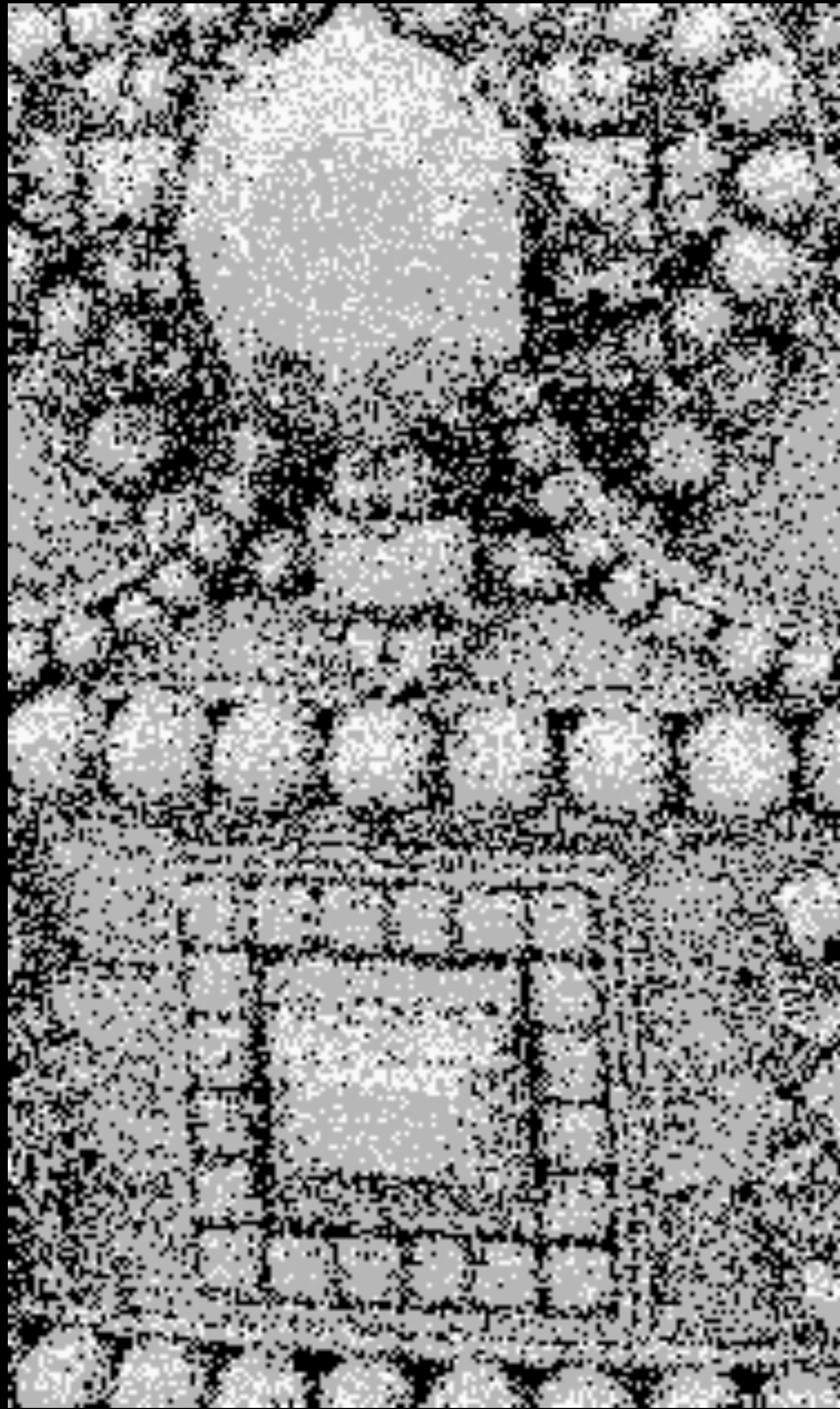


Random A



Random B

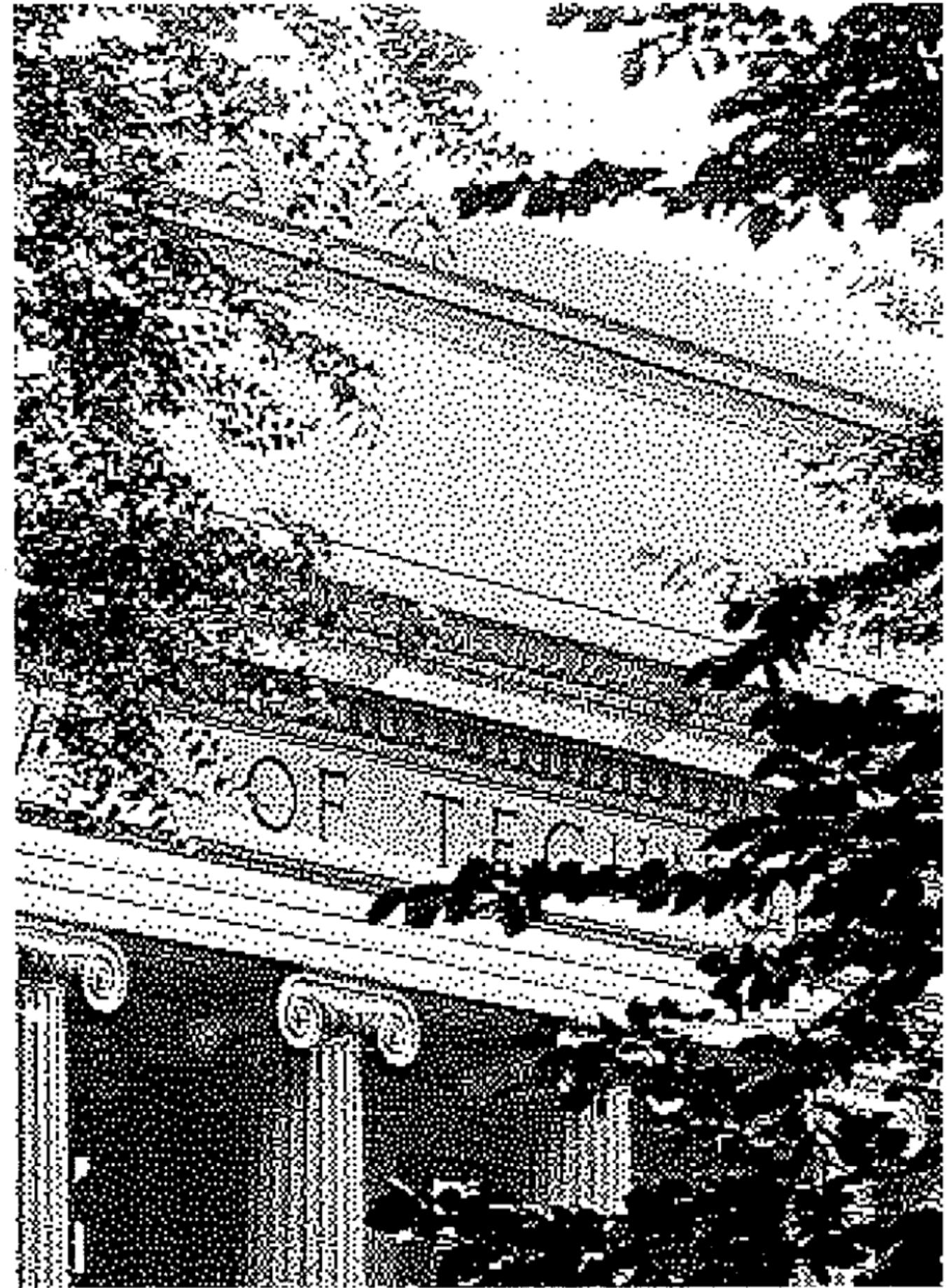




Blue Noise Dithering (Ulichney)



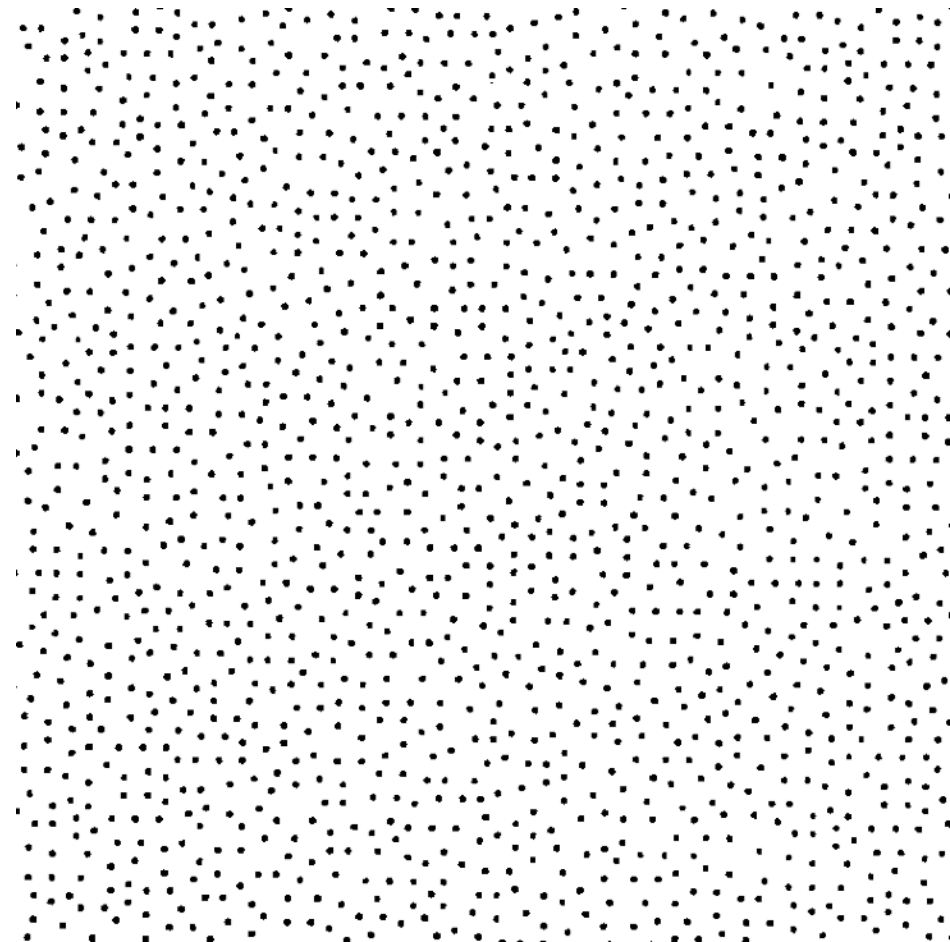
White



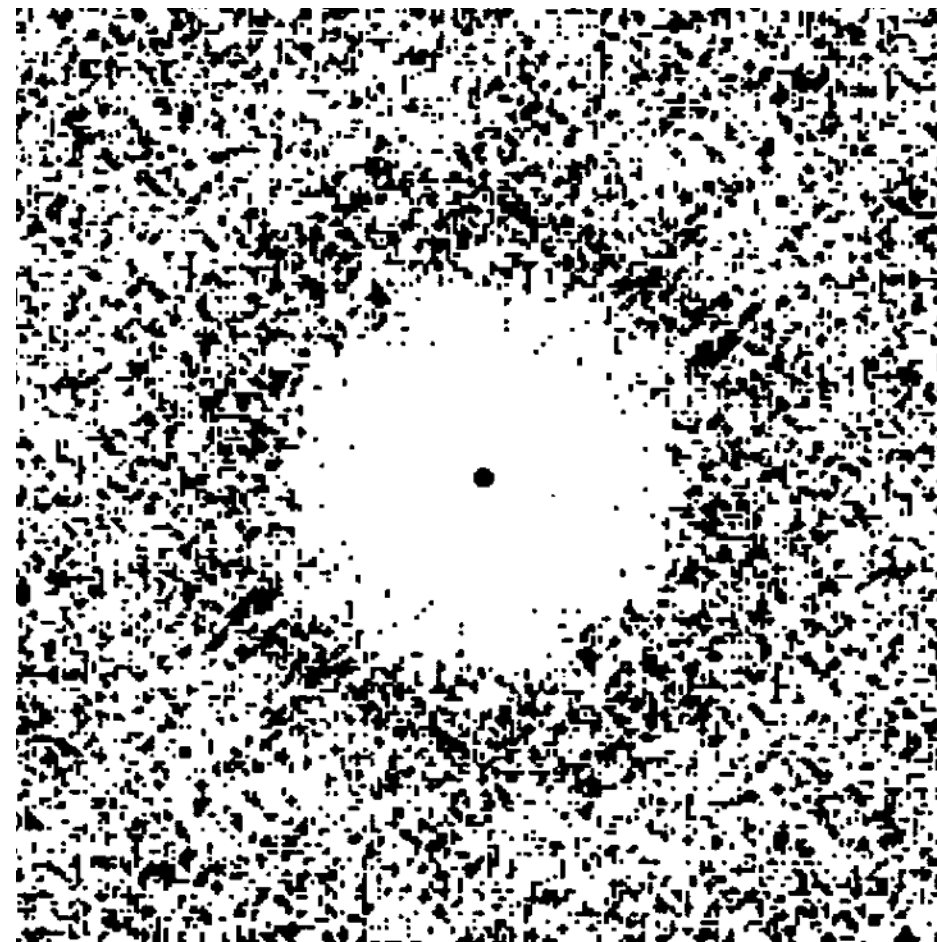
Blue

Power Spectrum of Samples

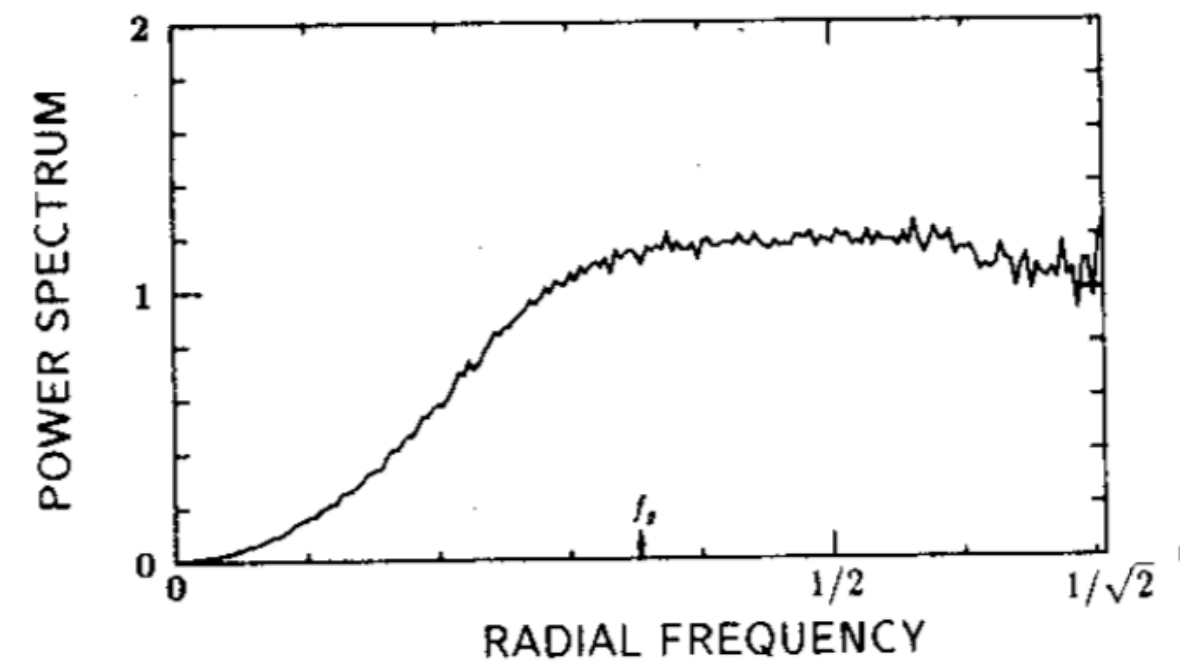
Samples



Fourier Transform



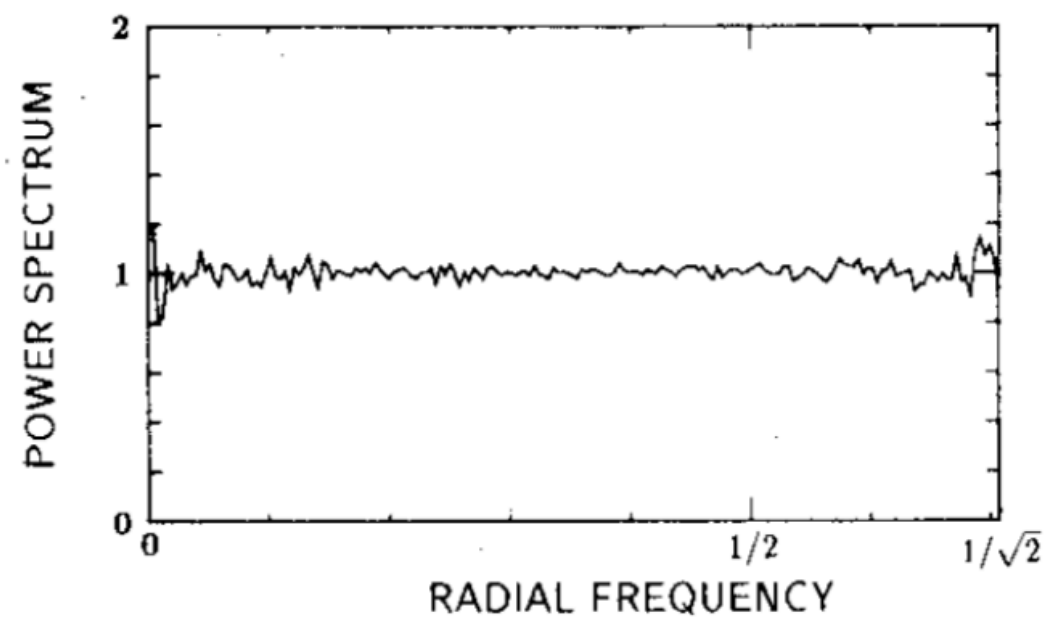
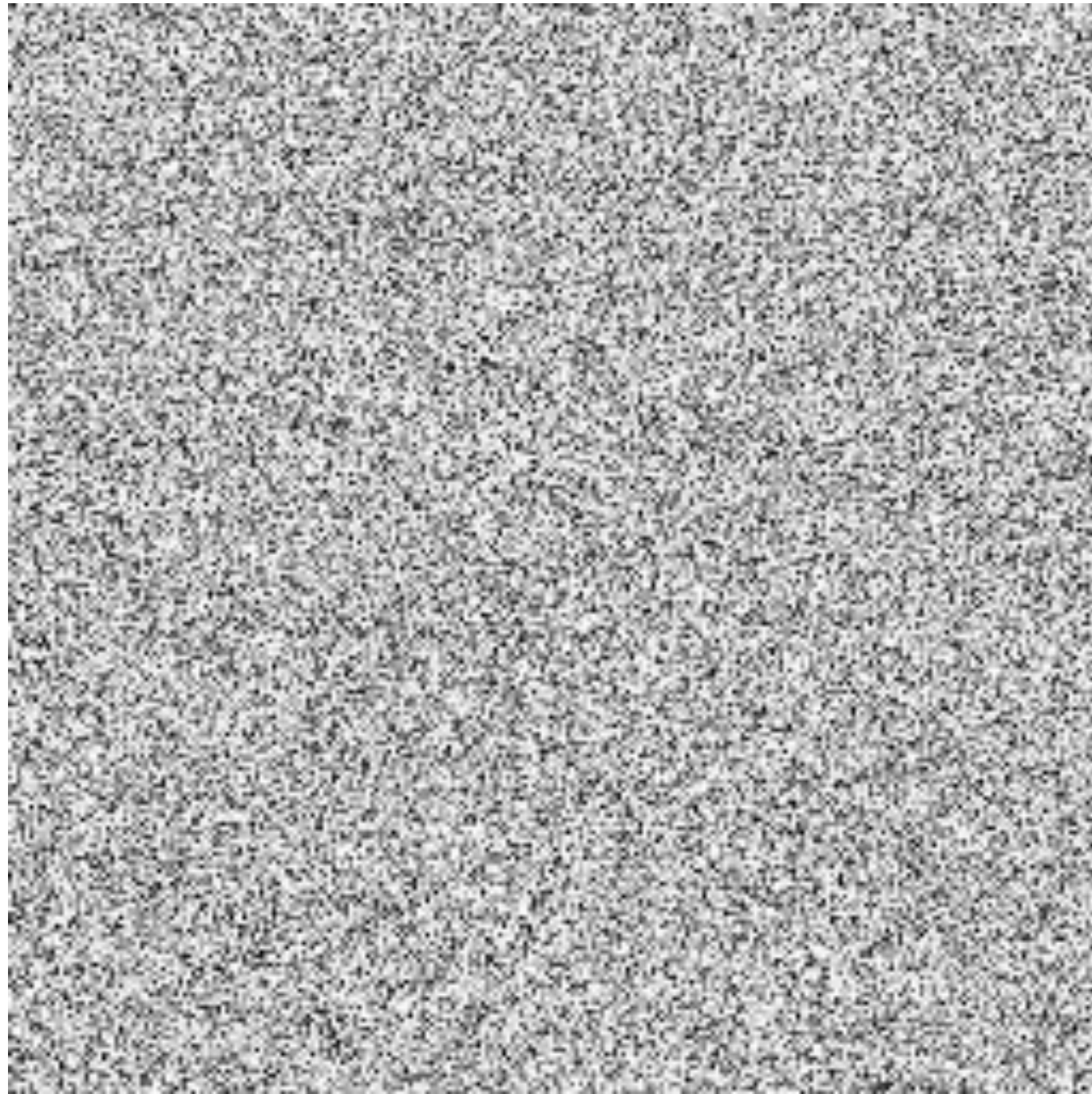
Radial Power Spectrum



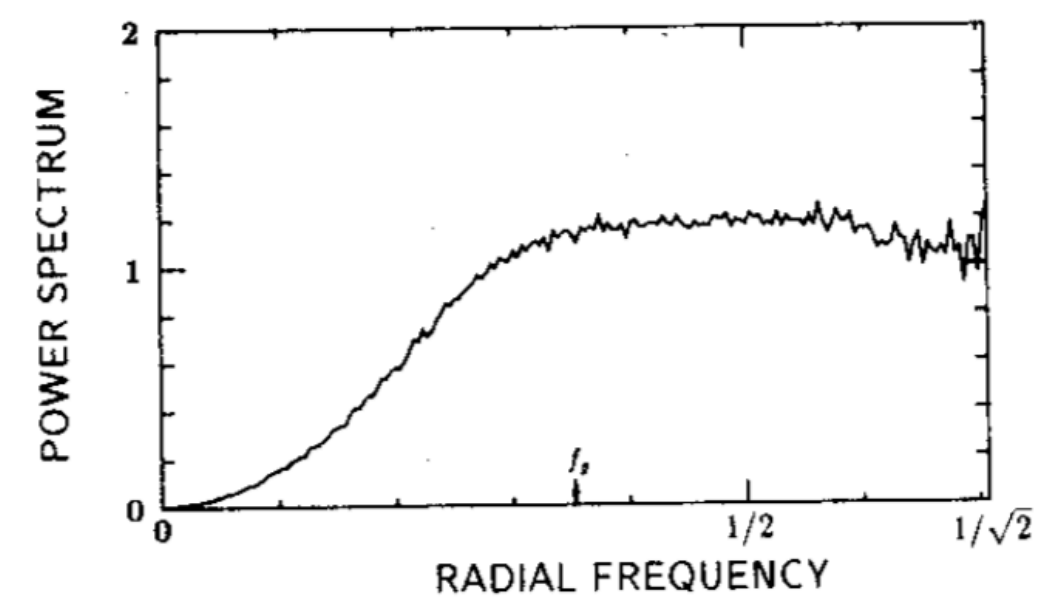
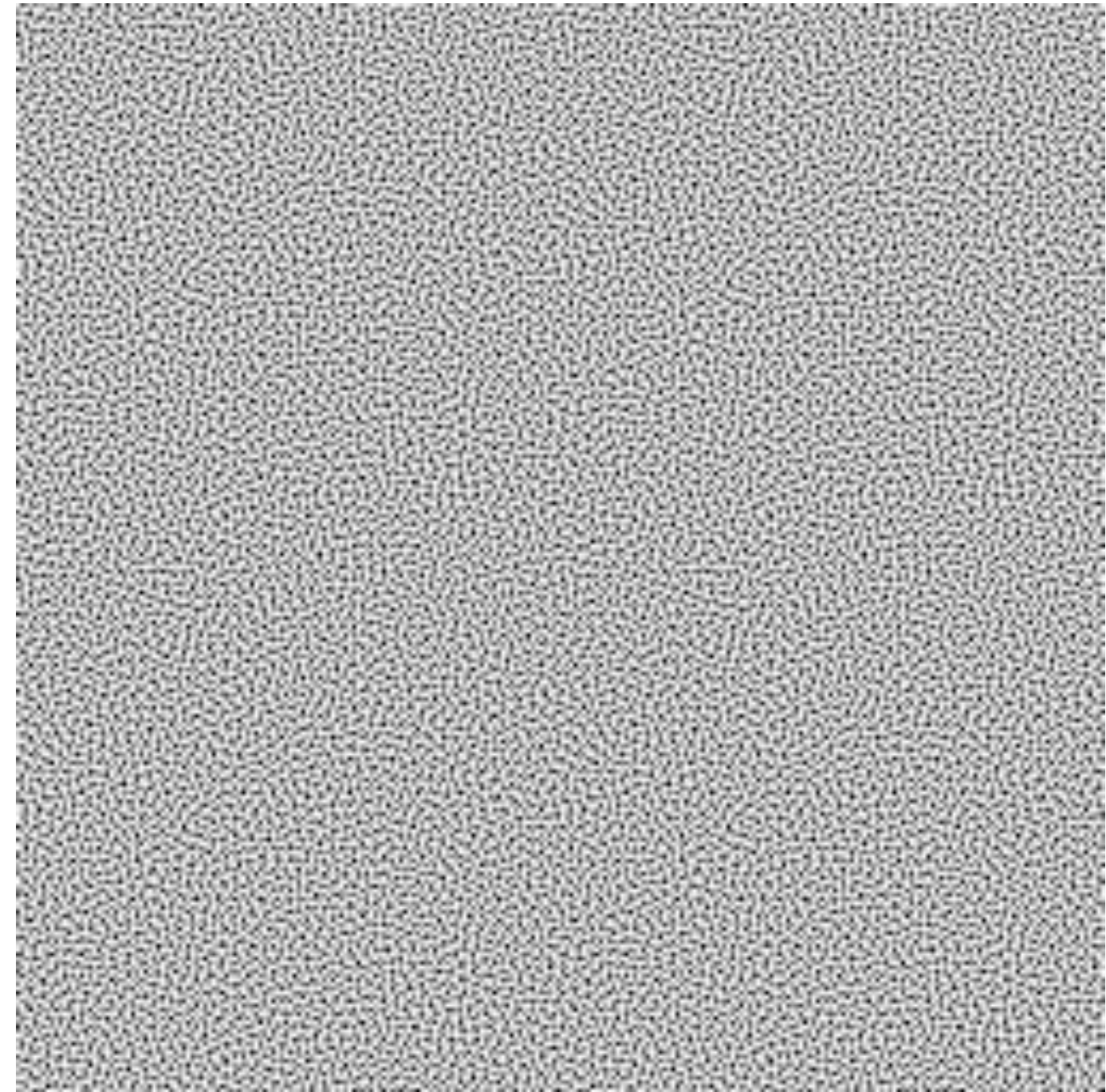
$$\begin{aligned} P_f(\omega) &= F(\omega) \overline{F(\omega)} \\ &= F(\omega)^2 \end{aligned} \quad f \text{ is even}$$

Colors of Noise

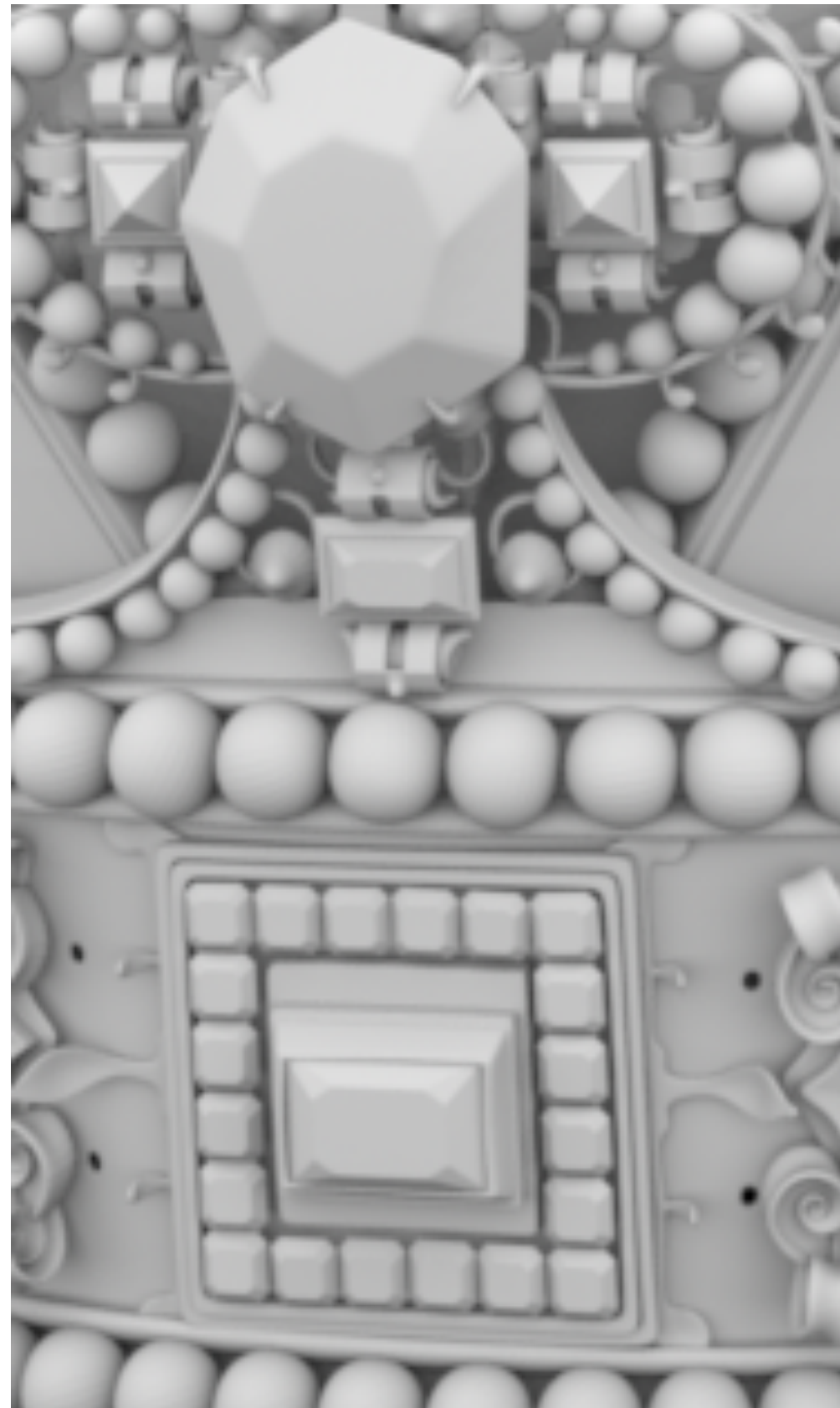
White



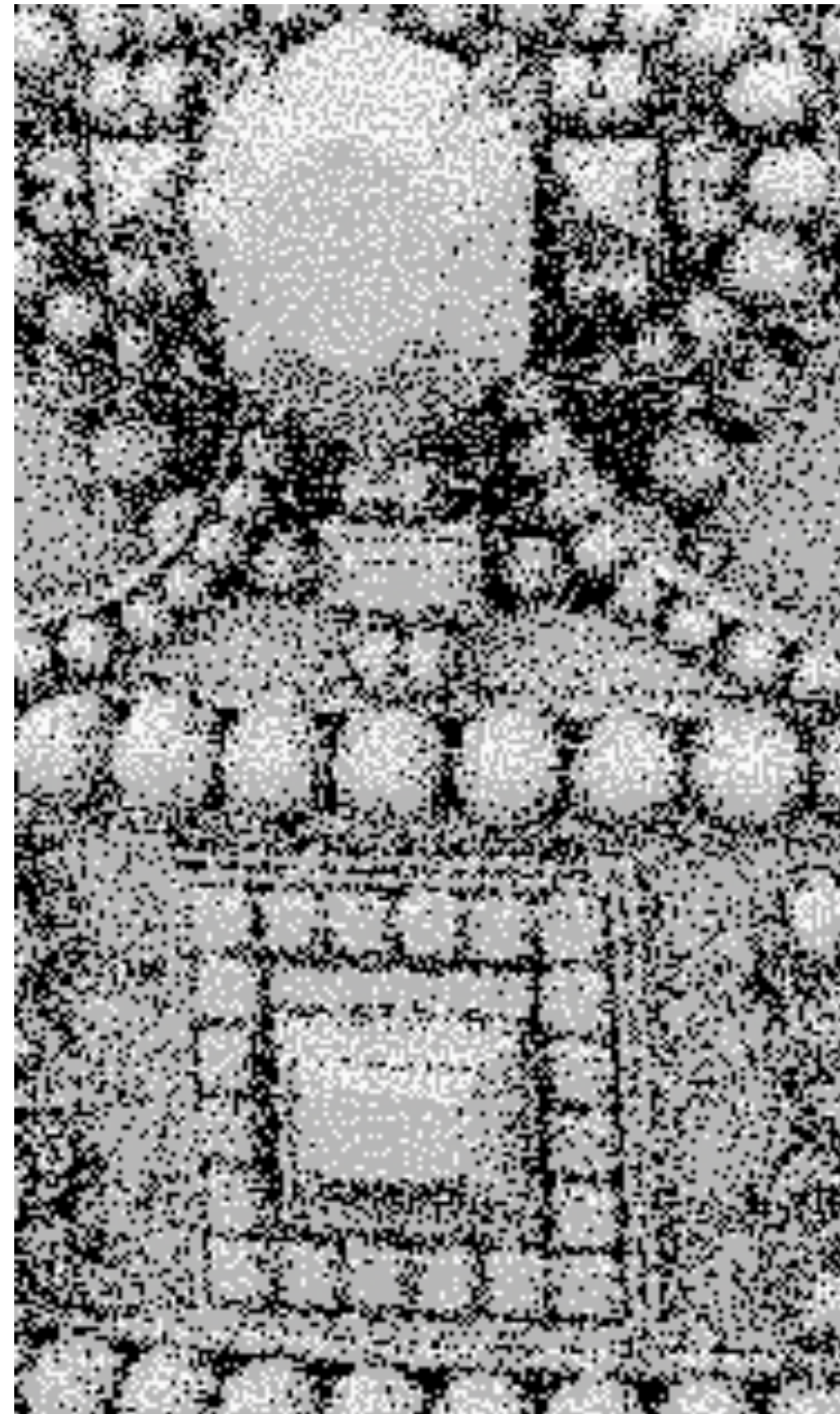
Blue



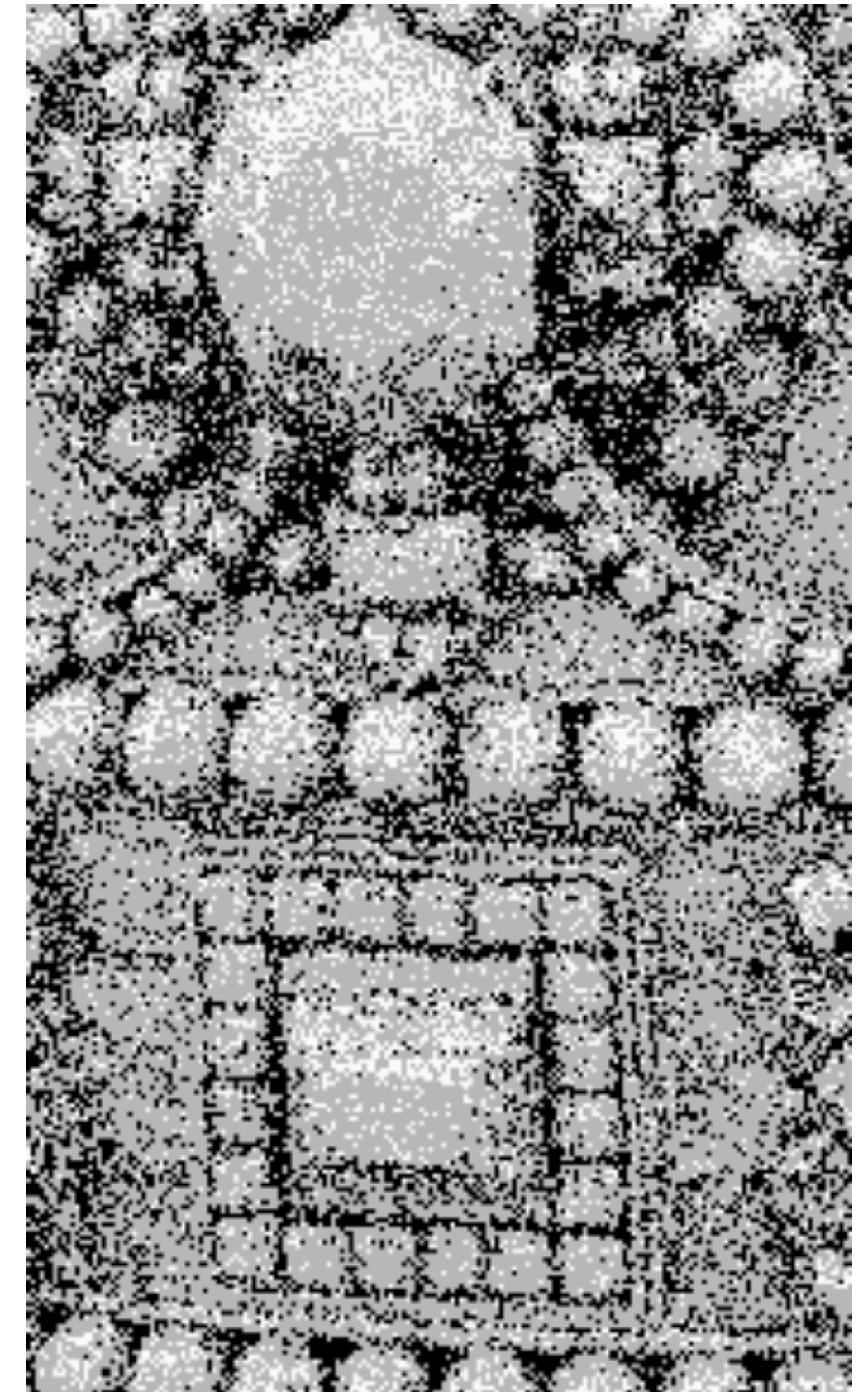
Ambient Occlusion, Revisited $\int_{\Omega} V(\omega) \cos \theta d\omega$



Reference

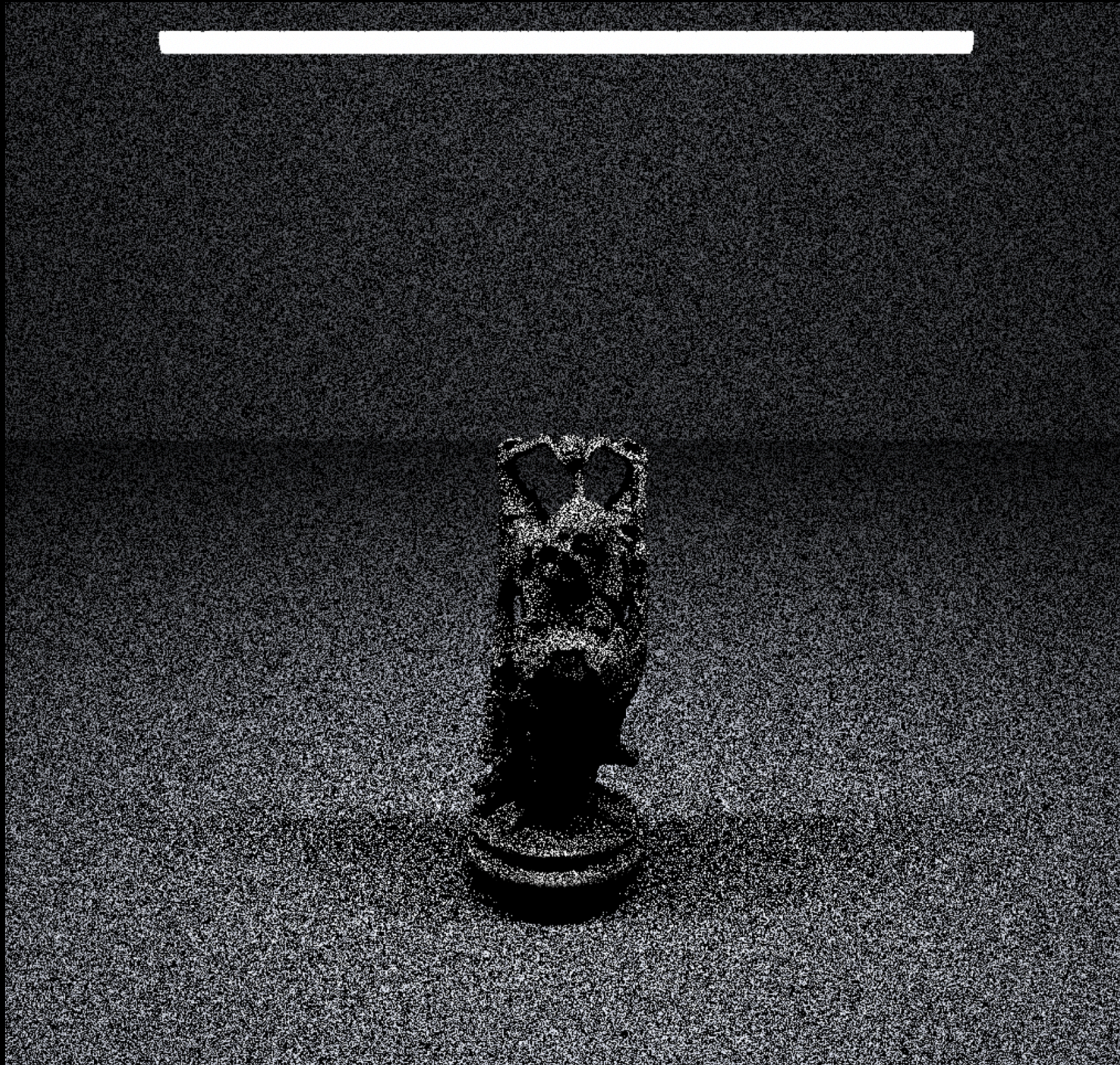


Blue Noise



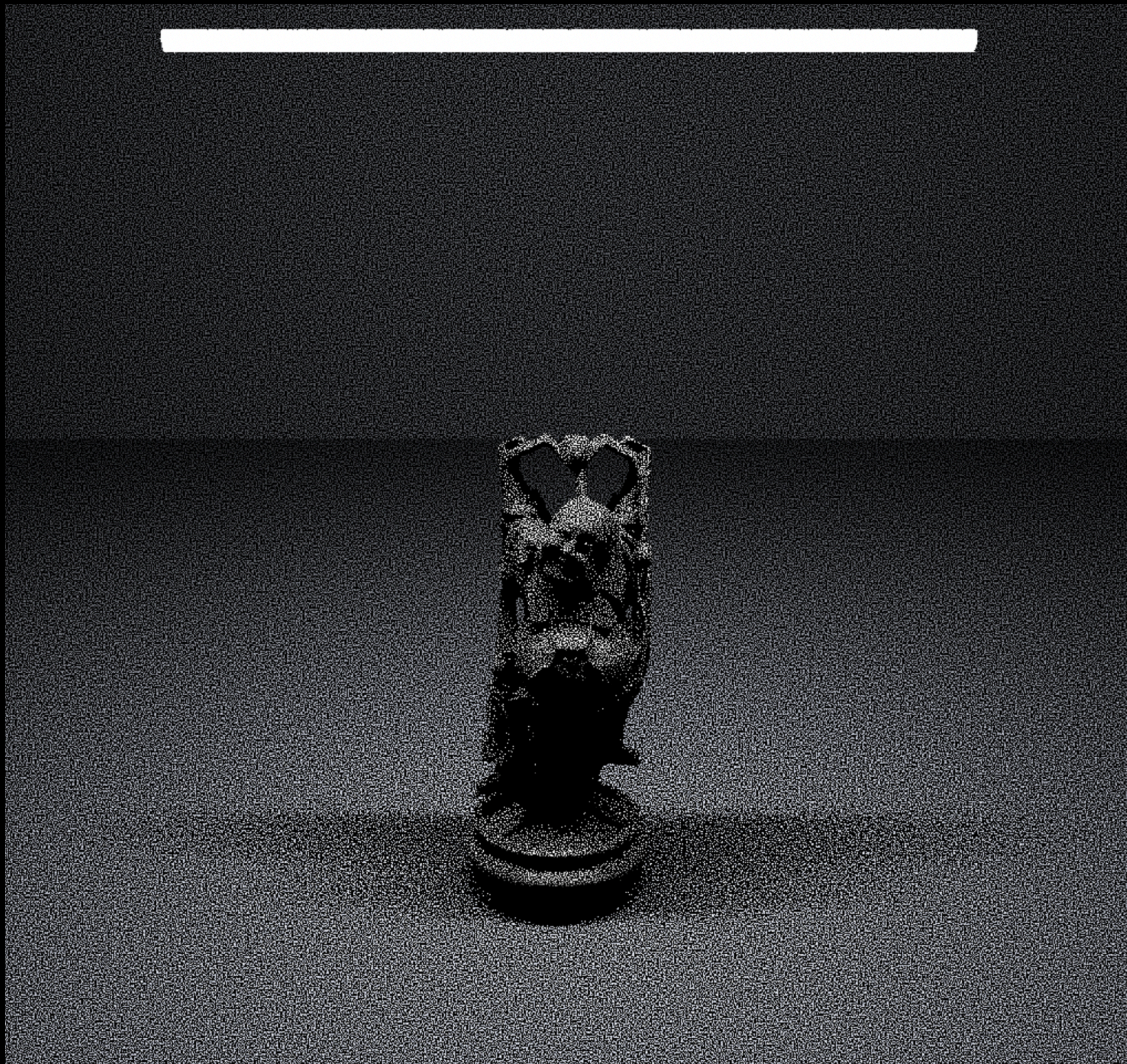
White Noise

(0,2)-sequence



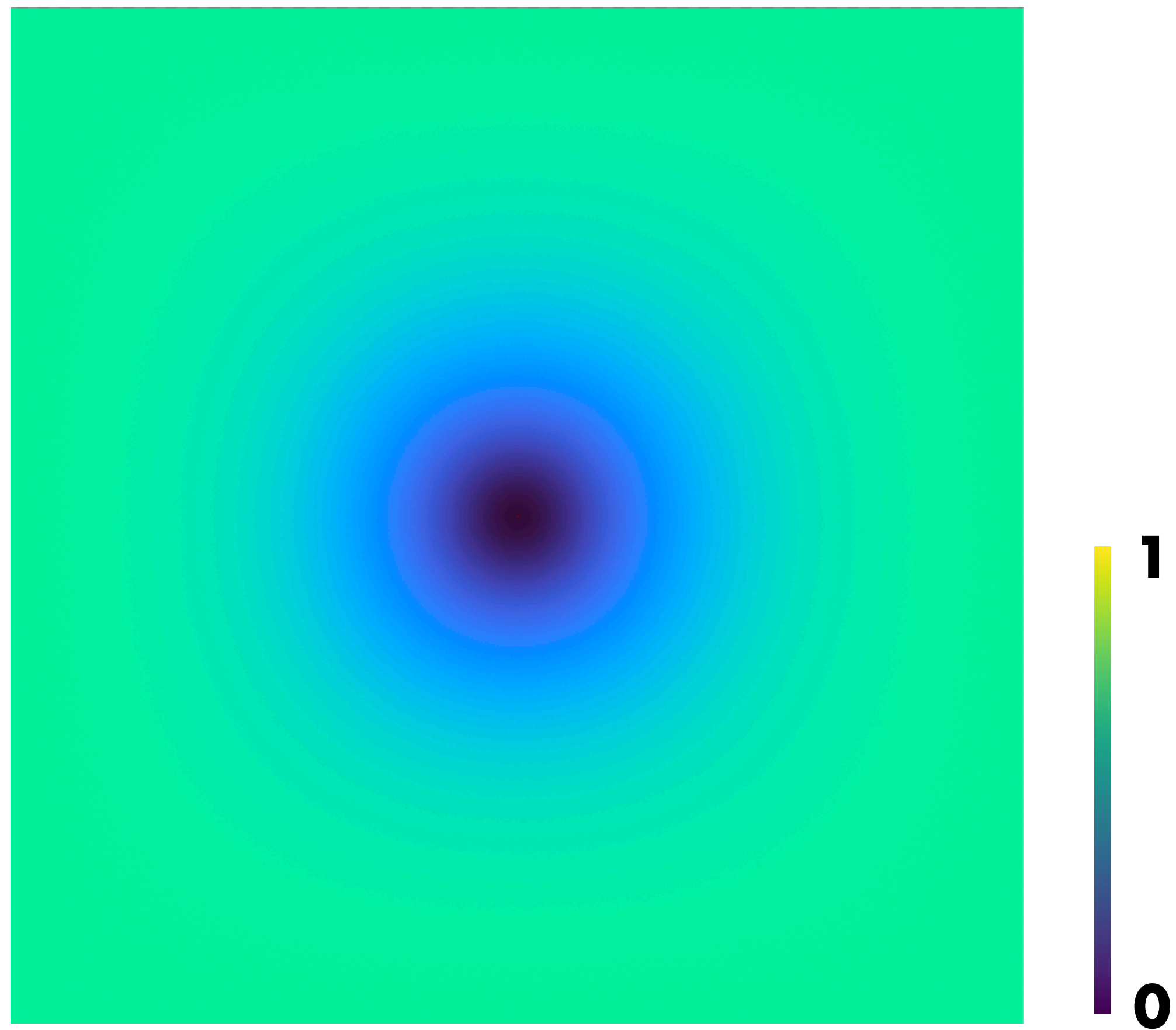
[Ahmed and Wonka 2020]

Blue Noise Sobol'

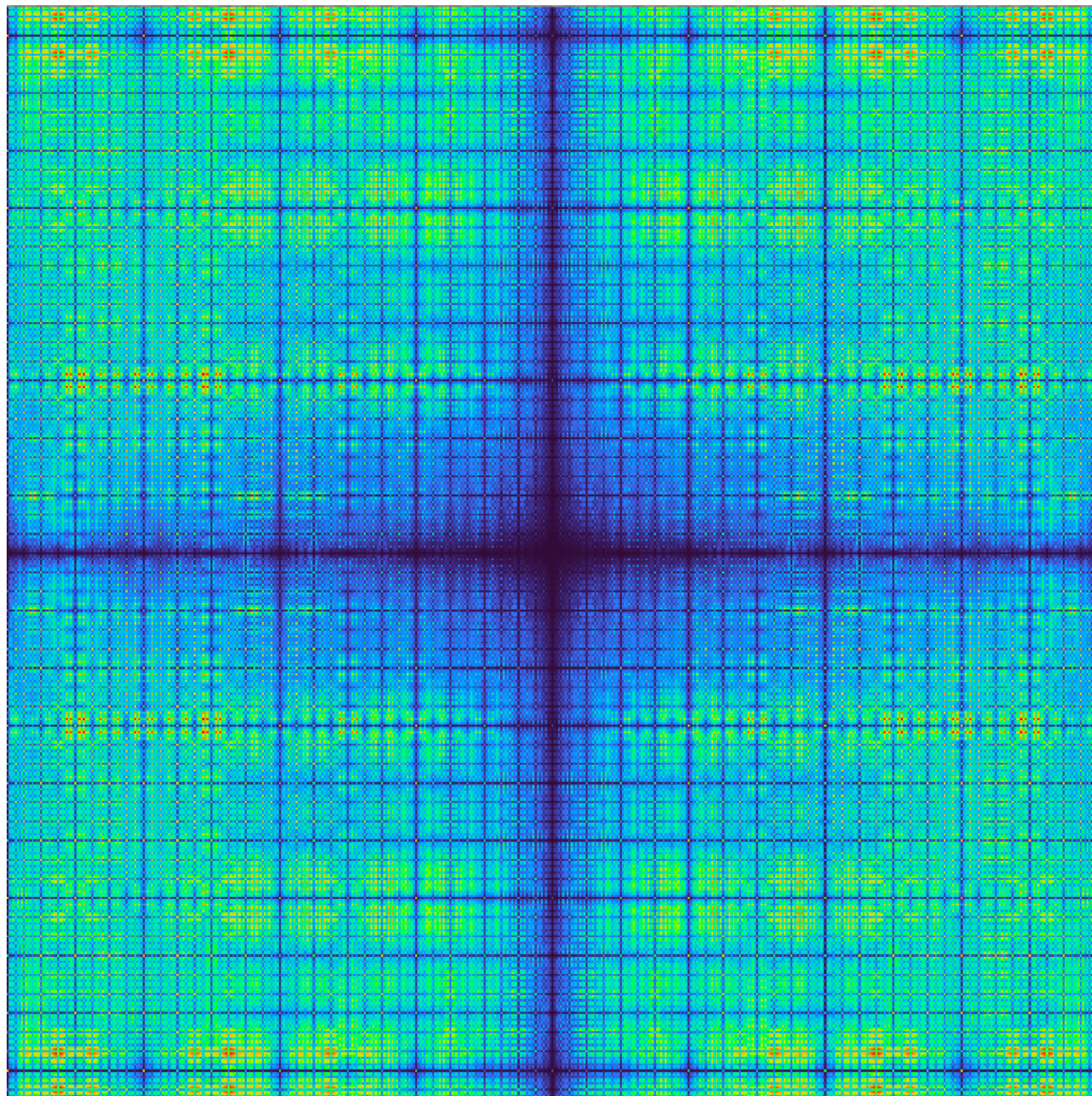


[Ahmed and Wonka 2020]

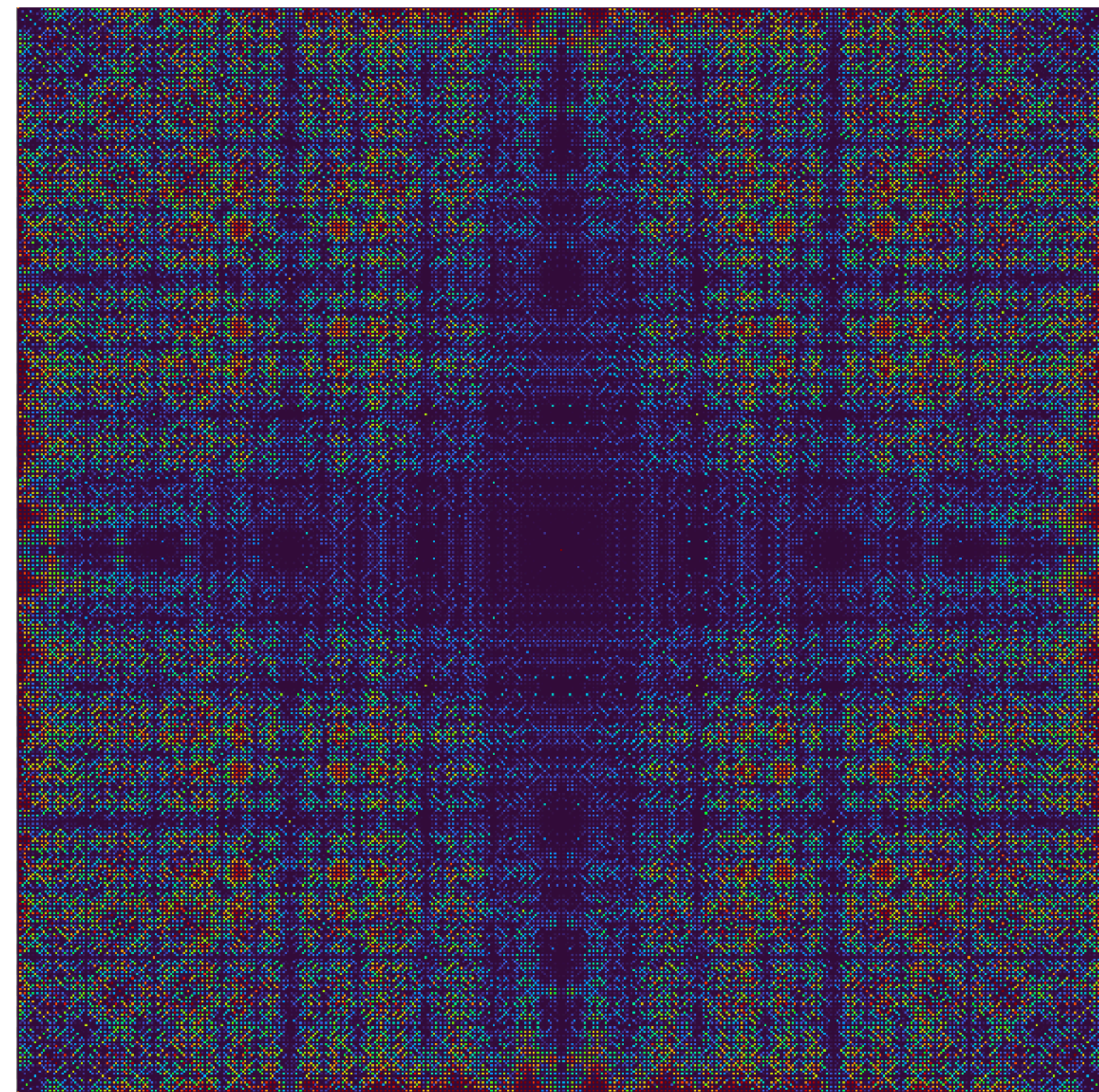
Jittered Power Spectrum



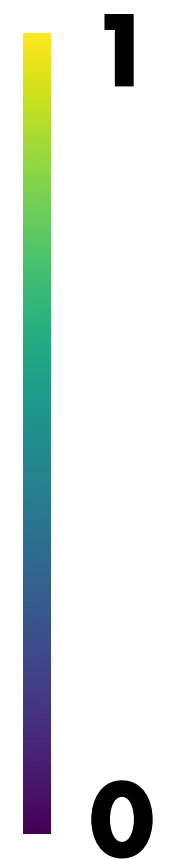
Low Discrepancy Power Spectra



Halton

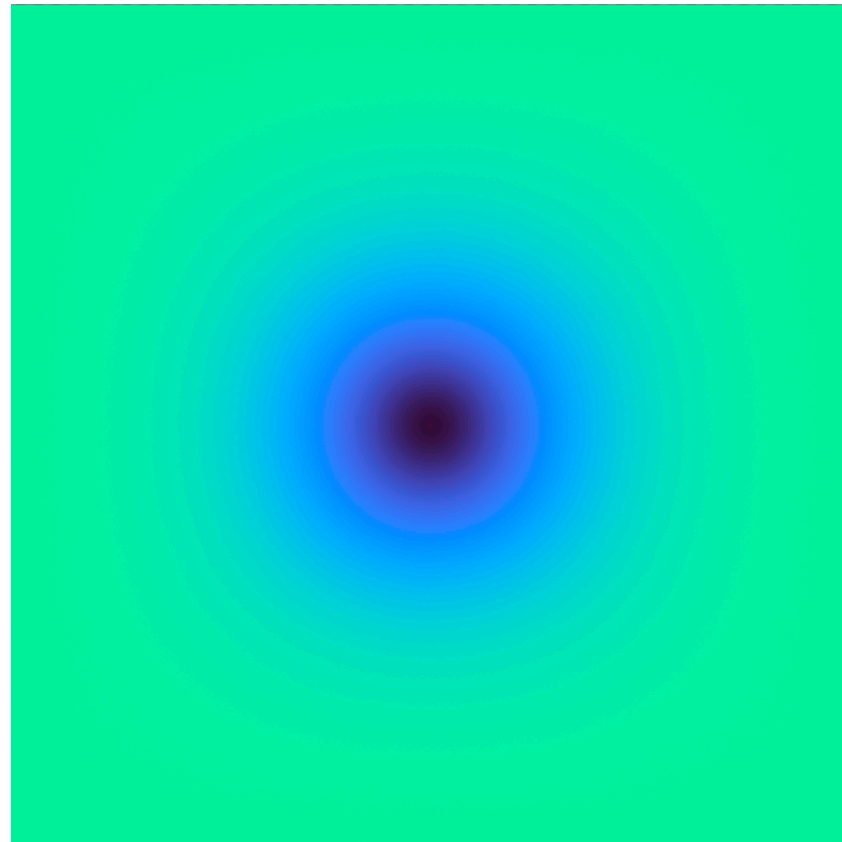


Sobol'

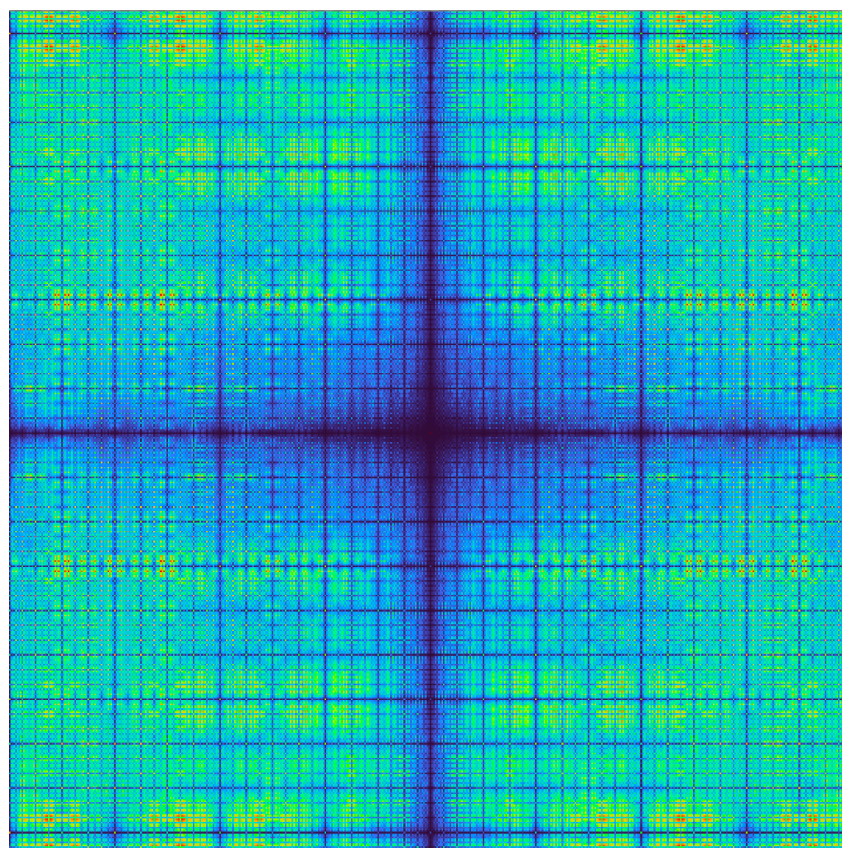


Power Spectra and Aliasing

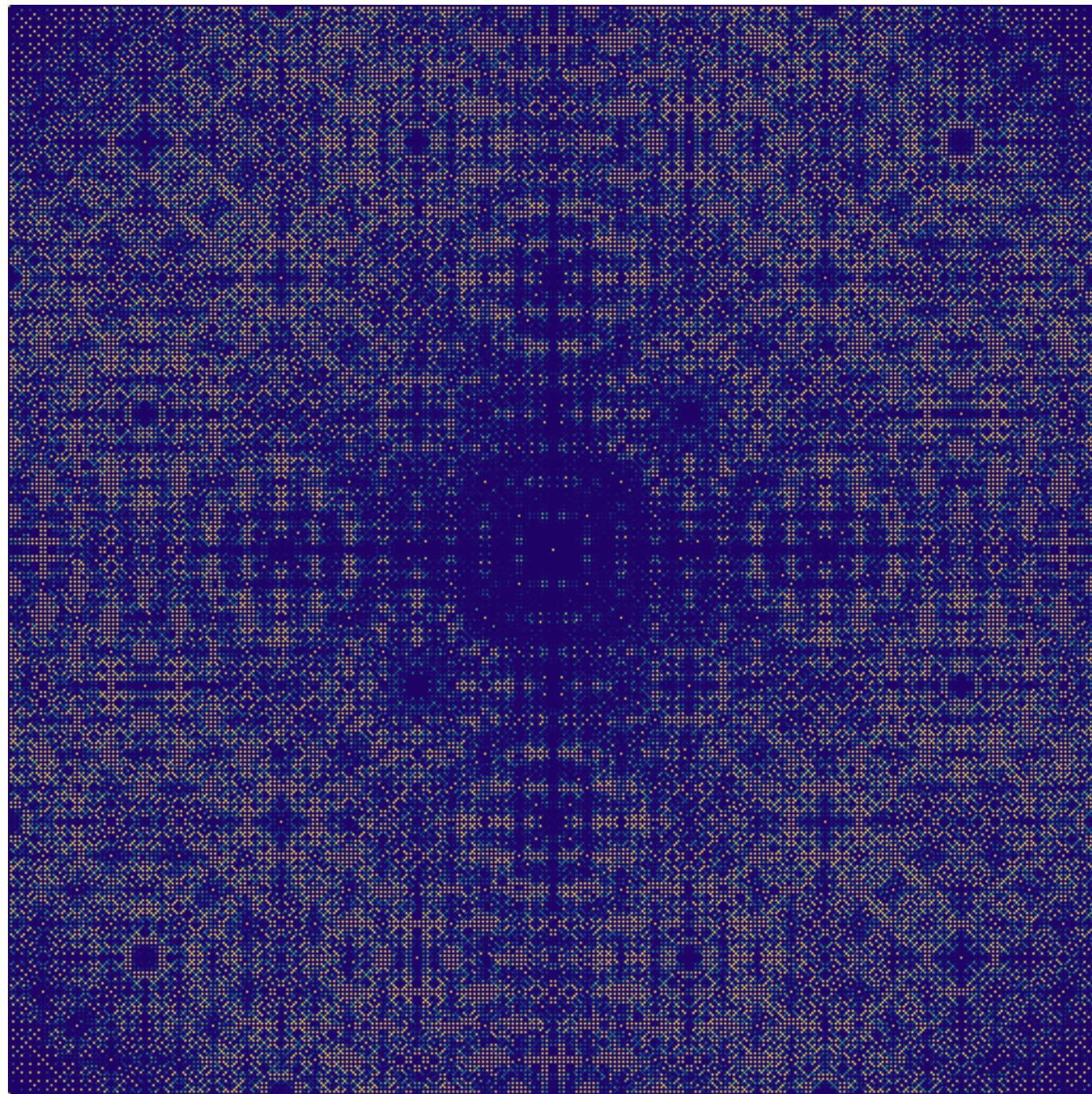
Jittered



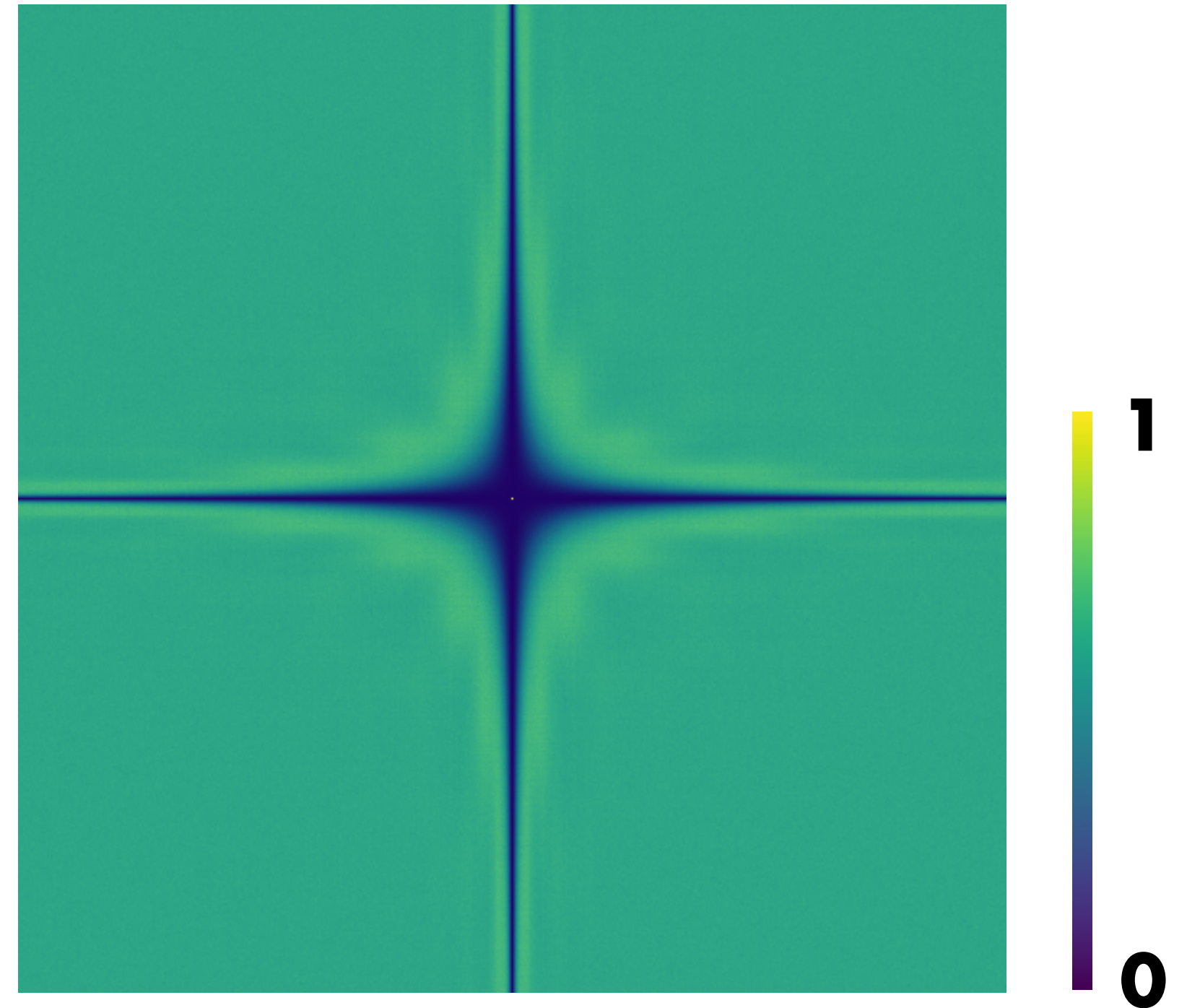
Halton



Randomized LD Power Spectra



Sobol'



Owen Scrambled Sobol'

Integration Error when Sampling

Integral

$$I(f) = \int f(x) dx$$

Sampled integral

Given samples x_i , define $s(x) = \frac{1}{N} \sum_{i=1}^N \delta(x - x_i)$

$$I_s(f) = \int f(x) s(x) dx = \frac{1}{N} \sum f(x_i)$$

Integration Error when Sampling

Sampled integral

$$I_s(f) = \int f(x)s(x) dx = \frac{1}{N} \sum f(x_i)$$

Since:

$$f(x)s(x) \leftrightarrow F(\omega) \oplus S(\omega)$$

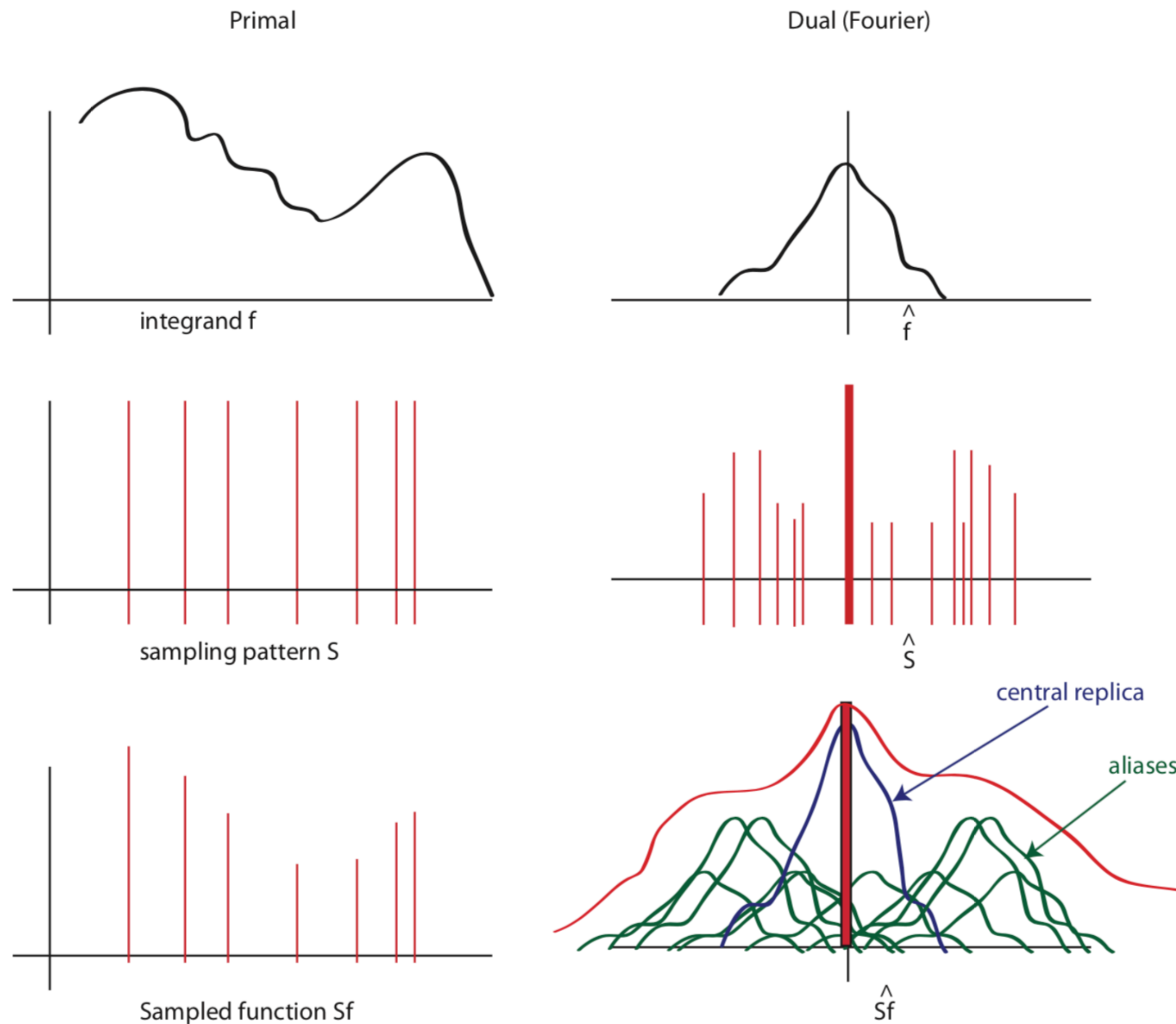
$$I(f) = \int f(x) dx = F(0)$$

$$I_s(f) = \int f(x)s(x) dx = F(\omega) \oplus S(\omega)|_{\omega=0}$$

Error

$$\Delta = F(0) - F(\omega) \oplus S(\omega)|_{\omega=0}$$

Integration in the Frequency Domain



[Durand 2011]

Variance Analysis

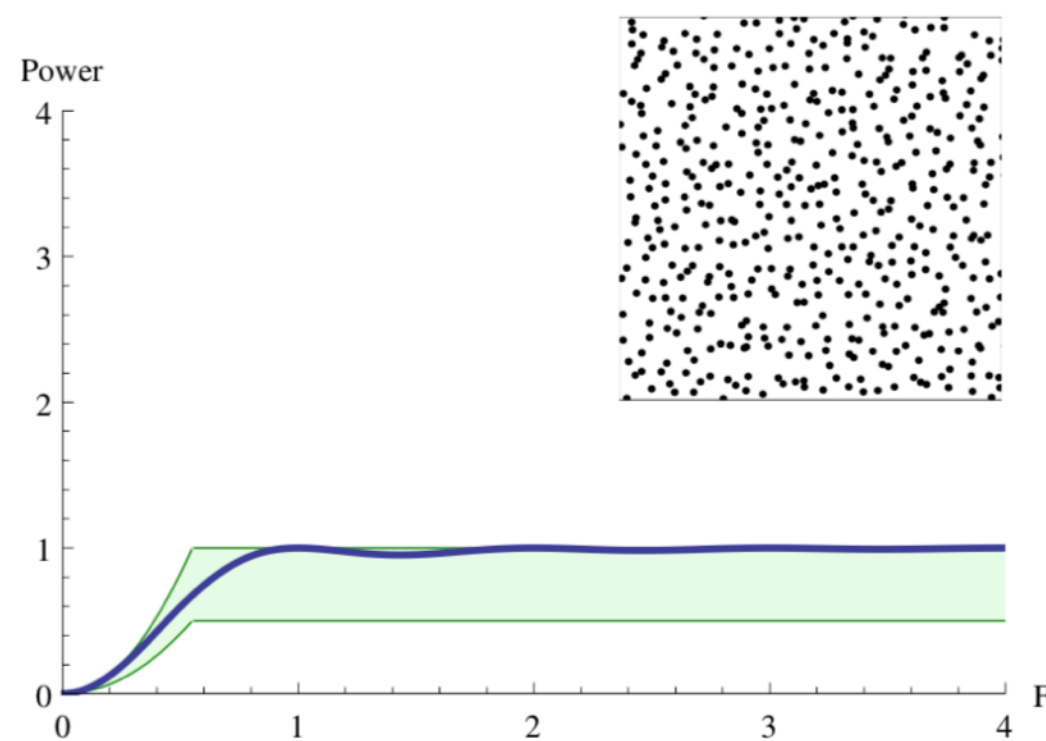
Recall error is: $\Delta = F(0) - F(\omega) \oplus S(\omega)|_{\omega=0}$

Can show that: $V \propto \int_{\Omega - \{0\}} P_F(\omega) P_S(\omega) d\omega$

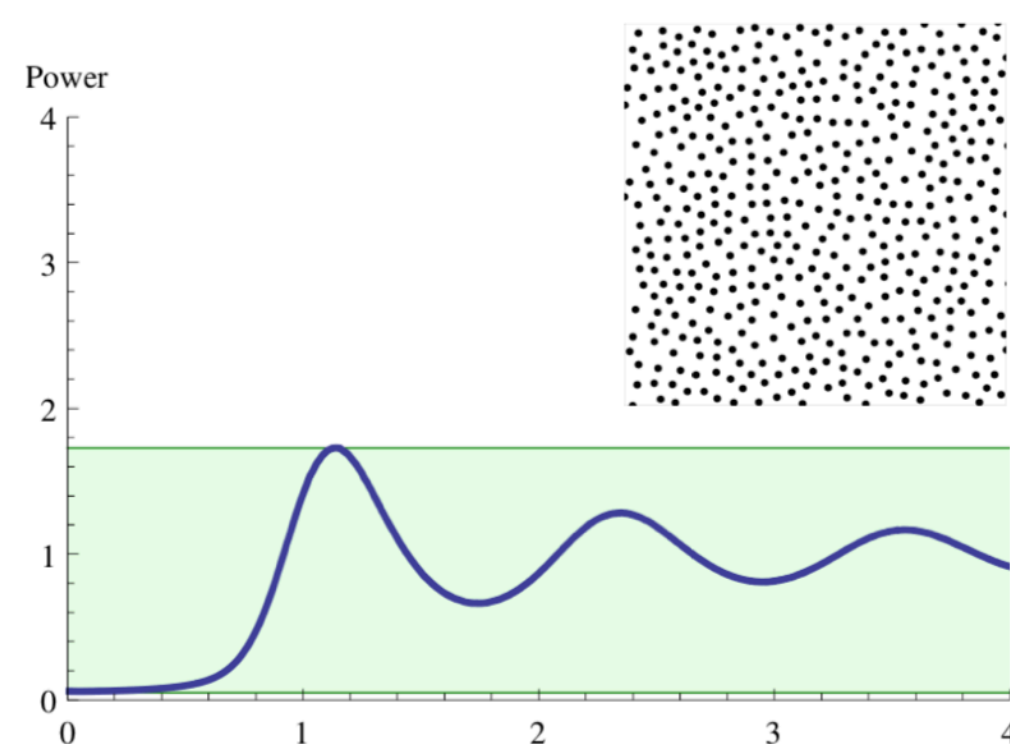
where $P_*(\omega)$ is the power spectrum of the signal

Power Spectra and Variance

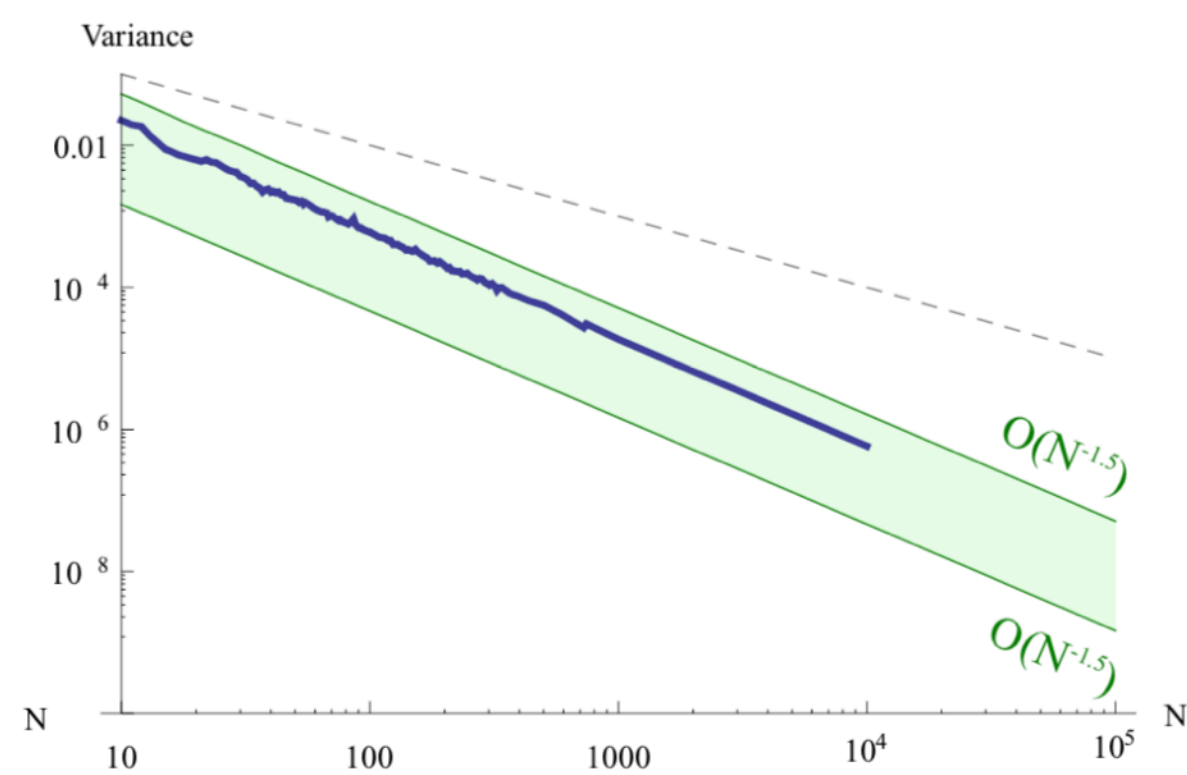
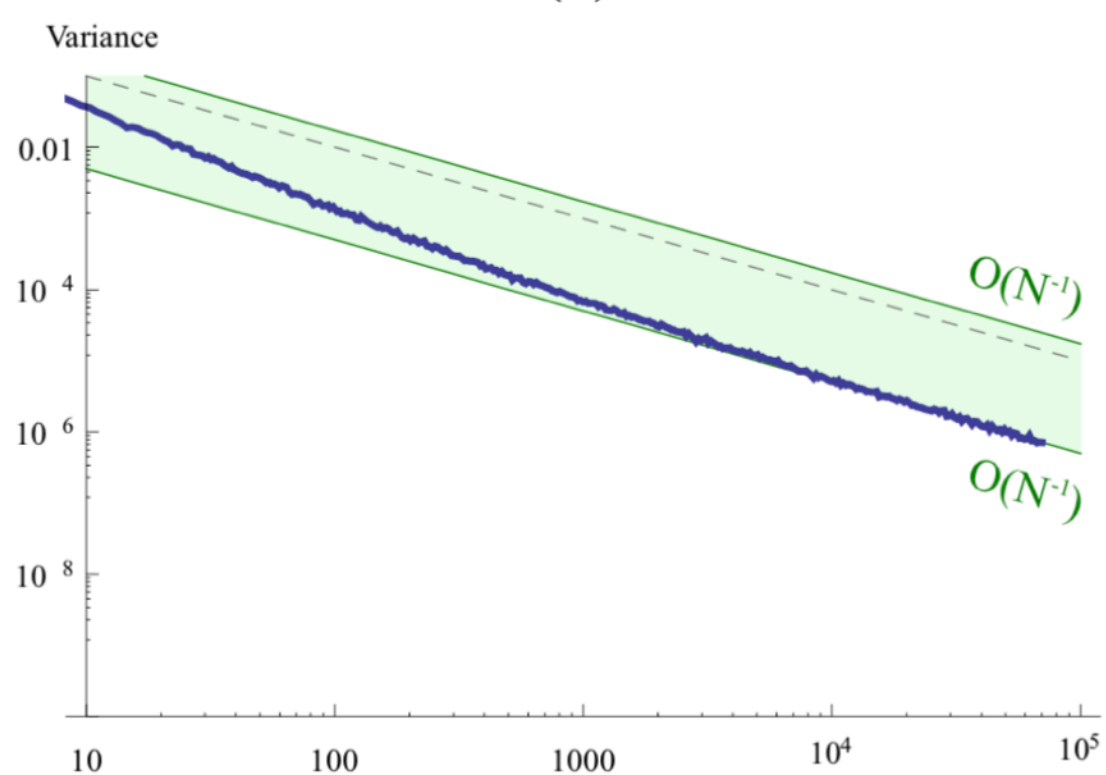
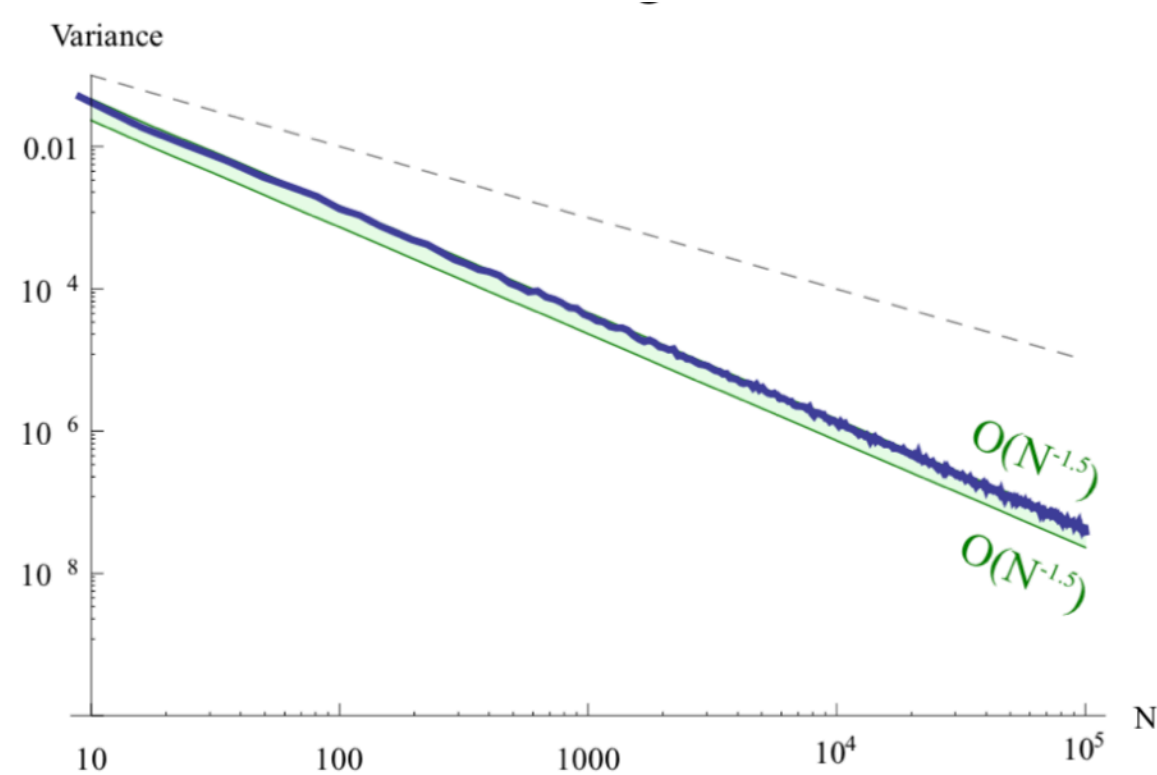
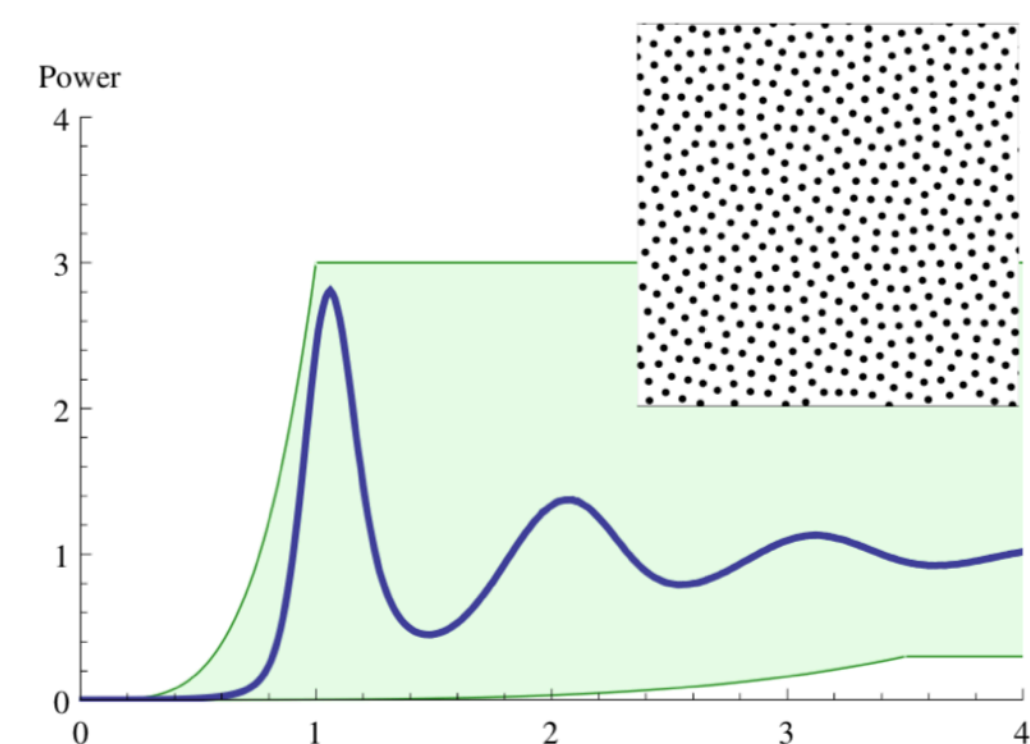
Jittered



Poisson Disk



Blue Noise



[Pilleboue et al. 2015]